

# ECE 137 B: Notes Set 12

## Negative Feedback & Bandwidth: Root Locus

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# Feedback with ideal op-amp

Assume ideal op-amp ( $Z_{in} = \infty \Omega$ ,  $Z_{out} = 0 \Omega$ ,  $CMRR = \infty$ )

$$V_{out} = A_D (V^+ - V^-)$$

$A_D = A_{OL} =$  open-loop gain

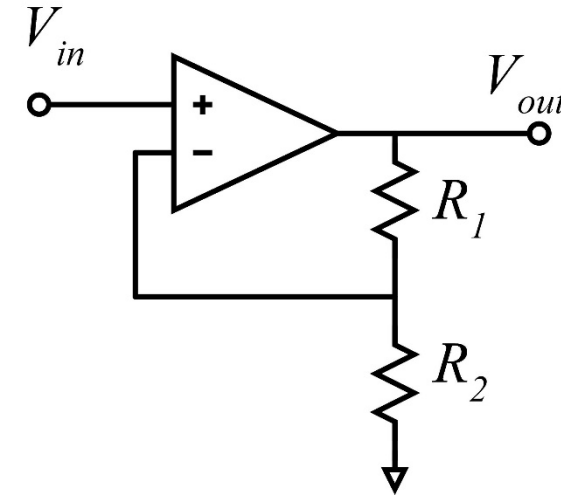
$$V^+ = V_{in}$$

$$V^- = \beta V_{out}$$

$$\beta = \text{feedback factor} = \frac{R_2}{R_1 + R_2}$$

$$T = \text{loop transmission} = A_D \beta = A_{OL} \beta$$

$$A_{CL} = \frac{V_{out}}{V_{in}} = \text{closed-loop gain} = \frac{1}{\beta} \frac{T}{1+T}$$



# Feedback: systems representation

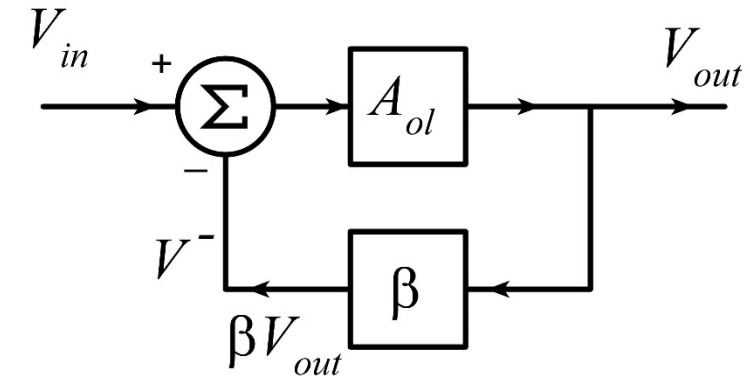
$$V_{out} = A_{OL}(V_{in} - V^-) = A_{OL}(V_{in} - \beta V_{out})$$

$$A_{CL}(s) = \frac{V_{out}(s)}{V_{in}(s)} = \text{closed-loop gain} = \frac{A_{OL}(s)}{1 + A_{OL}(s)\beta(s)} = \frac{1}{\beta(s)} \frac{T(s)}{1 + T(s)}$$

$\beta(s)$  = feedback factor

$A_{OL}(s)$  = open-loop gain

$T(s)$  = loop transmission =  $A_{OL}(s)\beta(s)$



This might represent a physical system with feedback:

Car anti-lock braking

Electronic feedback control of aircraft roll, pitch, or yaw

Car anti-rollover protection

Fuel/air mixture control in car engine

Robotics: position of mechanical arm

....

It is widely stated that feedback increases bandwidth...

once we ensure that the feedback is stable, this may or may not be true.

# Effect of Feedback on Single-Pole System (1)

$$A_{CL}(s) = \frac{V_{out}(s)}{V_{in}(s)} = \frac{A_{OL}(s)}{1 + A_{OL}(s)\beta(s)} = \frac{1}{\beta(s)} \frac{T(s)}{1 + T(s)}$$

Consider an open-loop gain with a single real-axis pole:

$$A_{OL}(s) = \frac{A_{OL,DC}}{1 + s / \omega_{OL}}$$

$A_{OL,DC}$  = open-loop gain at DC

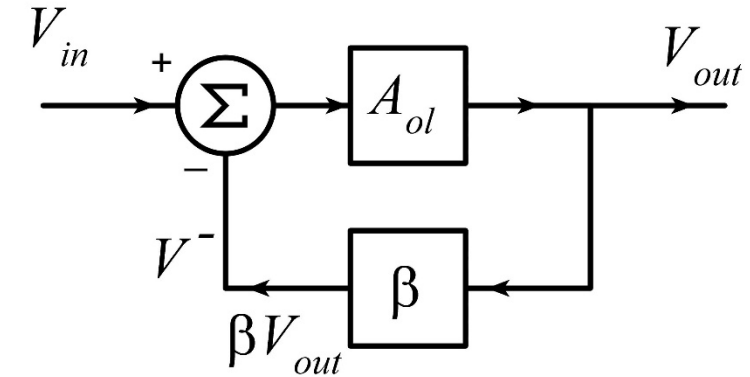
$\omega_{OL}$  = one real-axis pole in the open-loop gain

Consider a frequency-independent feedback factor :

$$\beta(s) = \beta_0 ;$$

$$\rightarrow T(s) = A_{OL}(s)\beta(s) = \frac{A_{OL,DC}\beta_0}{1 + s / \omega_{OL}} = \frac{T_0}{1 + s / \omega_{OL}} = \frac{N_T(s)}{D_T(s)}$$

$$T_0 = A_{OL,DC}\beta_0 = \text{loop transmission at DC}$$



# Effect of Feedback on Single-Pole System (2)

$$\begin{aligned}
 A_{CL}(s) &= \frac{V_{out}(s)}{V_{in}(s)} = \frac{1}{\beta(s)} \frac{T(s)}{1+T(s)} = \frac{1}{\beta(s)} \frac{N_T / D_T}{1+N_T / D_T} = \frac{1}{\beta(s)} \frac{N_T}{N_T + D_T} \\
 &= \frac{1}{\beta_0} \frac{T_0}{T_0 + (1+s/\omega_{OL})} = \frac{1}{\beta_0} \frac{T_0}{1+T_0 + s/\omega_{OL}} \\
 &= \frac{1}{\beta_0} \frac{T_0}{1+T_0} \frac{1}{1+\left(\frac{s}{(1+T_0)\omega_{OL}}\right)}
 \end{aligned}$$

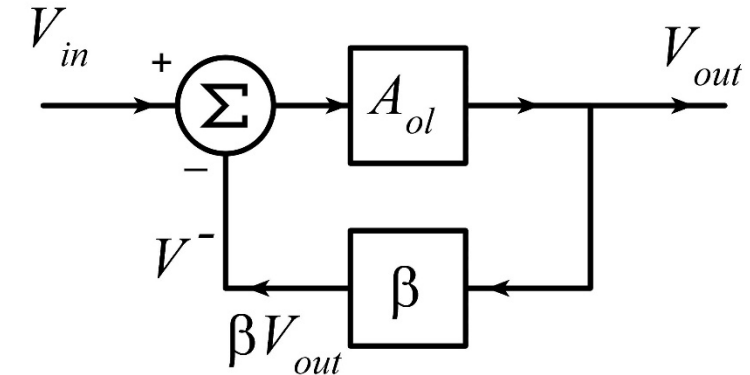
So:

$$A_{CL}(s) = A_{CL,DC} \frac{1}{1+s/\omega_{CL}}$$

where  $A_{CL,DC} = \frac{1}{\beta_0} \frac{T_0}{1+T_0} =$  closed-loop gain at DC

and  $\omega_{CL} = (1+T_0)\omega_{OL} =$  bandwidth (pole frequency) of closed-loop gain.

The bandwidth has increased by the factor  $(1+T_0)$



# Effect of Feedback on Single-Pole System (3)

Example:

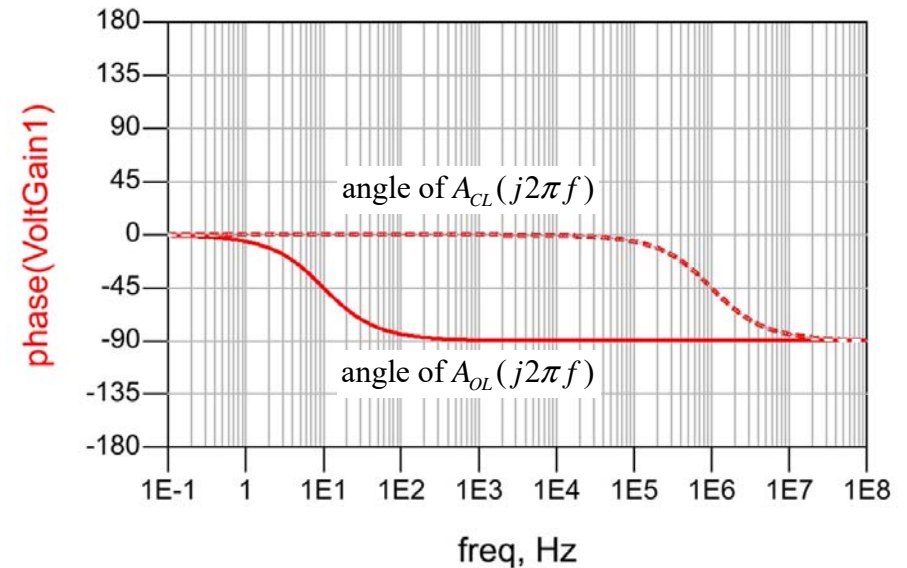
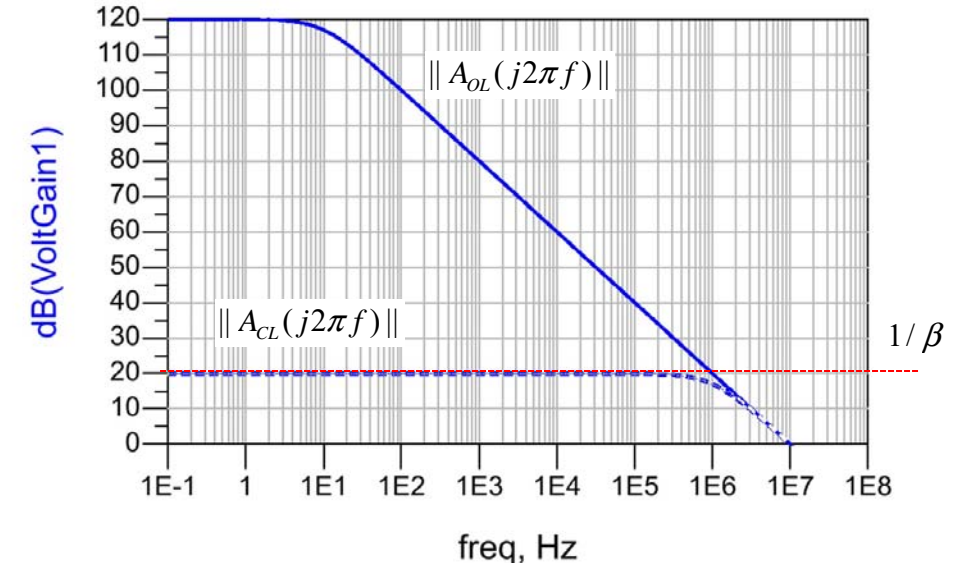
$$A_{OL}(j2\pi f) = \frac{A_{OL,DC}}{1 + jf / f_{OL}} = \frac{10^6}{1 + jf / 10 \text{ Hz}}$$

$$\beta(j2\pi f) = \beta_0 = 1/10$$

$$A_{CL}(j2\pi f) = \frac{1}{\beta(j2\pi f)} \frac{T(j2\pi f)}{1 + T(j2\pi f)} = A_{CL,DC} \frac{1}{1 + jf / f_{CL}}$$

$$\text{where } A_{CL,DC} = \frac{1}{\beta_0} \frac{T_0}{1 + T_0} = 10 \frac{10^6 / 10}{1 + 10^6 / 10} \approx 10$$

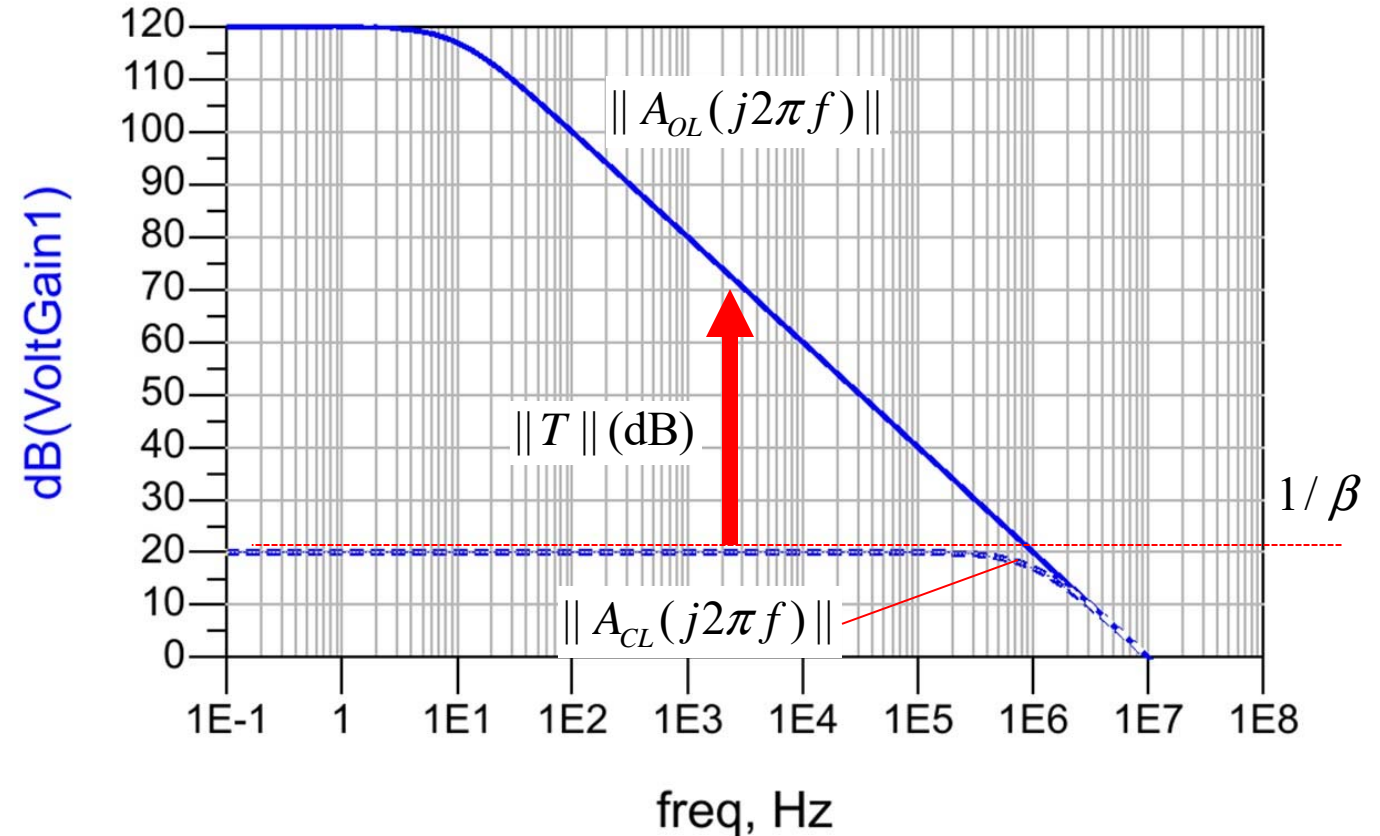
$$\text{and } f_{CL} = (1 + T_0) f_{OL} = (1 + 10^5)(10 \text{ Hz}) = 1 \text{ MHz.}$$



# Effect of Feedback on Single-Pole System (4)

$$T = A_{OL}\beta \rightarrow T(\text{dB}) = A_{OL}(\text{dB}) - (1/\beta)(\text{dB})$$

$$A_{CL} = \frac{1}{\beta} \frac{T}{1+T} = \frac{A_{OL}}{1+T} = \begin{cases} \frac{1}{\beta} & \|T\| \gg 1 \\ \frac{1}{\beta} \frac{e^{j\theta_r}}{1+e^{j\theta_r}} & \|T\| = 1 \\ A_{OL} & \|T\| \ll 1 \end{cases}$$



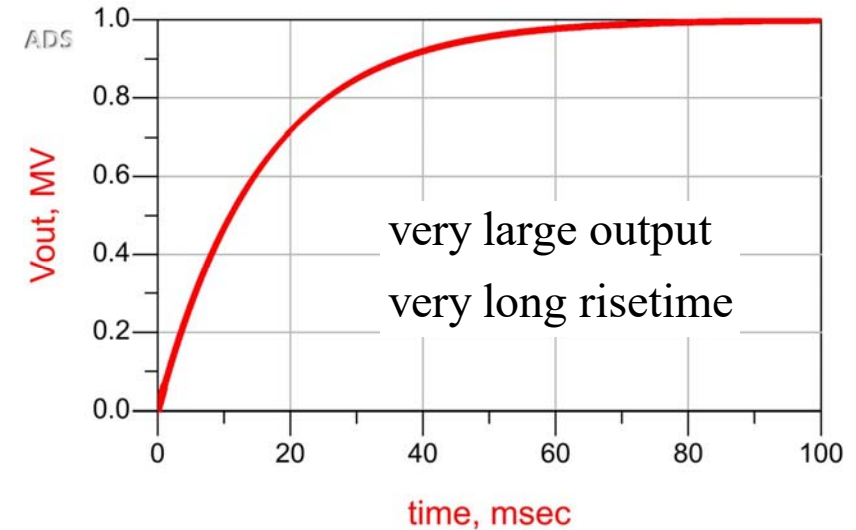
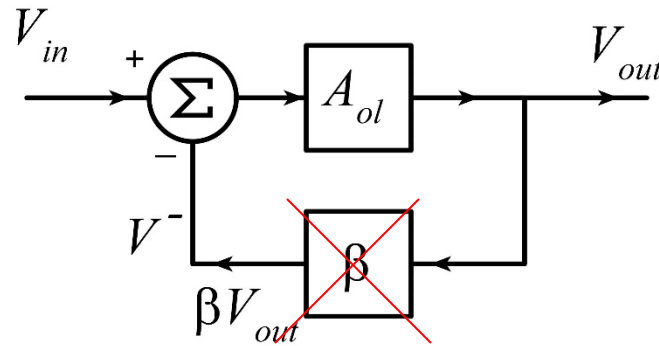
Example:

$$A_{OL}(j2\pi f) = \frac{A_{OL,DC}}{1 + jf/f_{OL}} = \frac{10^6}{1 + jf/10 \text{ Hz}}$$

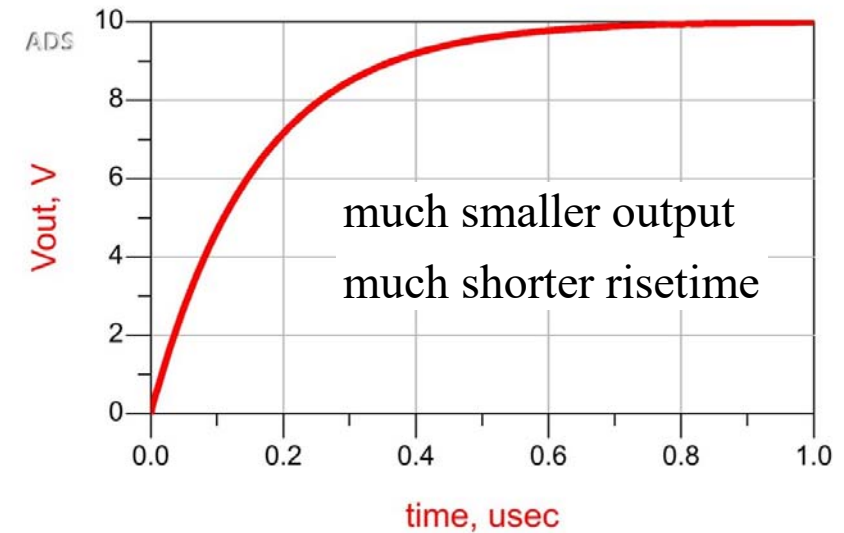
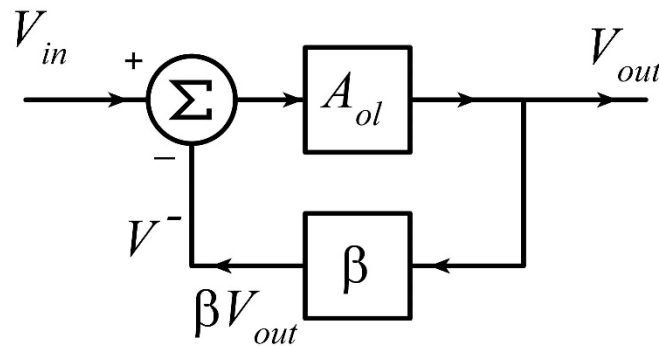
$$\beta(j2\pi f) = \beta_0 = 1/10$$

# Effect of Feedback on Single-Pole System (5)

Step response, no feedback



Step response, with feedback



Example:

$$A_{OL}(j2\pi f) = \frac{A_{OL,DC}}{1 + jf / f_{OL}} = \frac{10^6}{1 + jf / 10 \text{ Hz}}$$

$$\beta(j2\pi f) = \beta_0 = 1/10$$



# Effect of Feedback on Single-Pole System (6)

$$A_{OL}(s) = \frac{A_{OL,DC}}{1 + s / \omega_{OL}}$$

$$\beta(s) = \beta_0$$

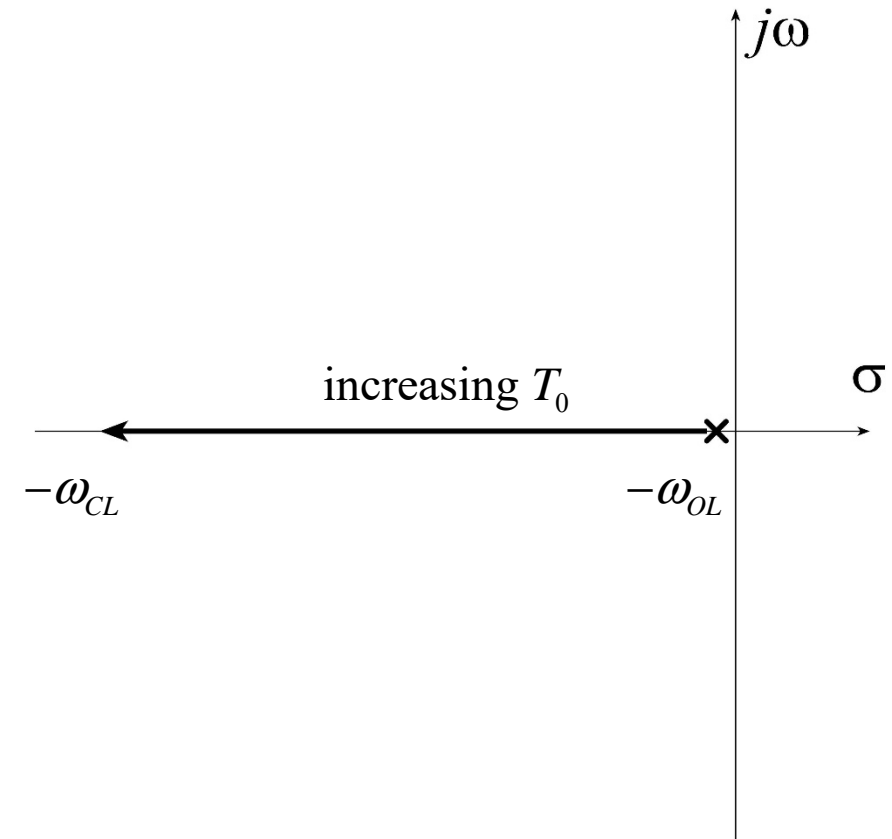
$$T(s) = A_{OL}(s)\beta_0$$

$$T(s) = \frac{A_{OL,DC}\beta_0}{1 + s / \omega_{OL}}$$

$$A_{CL}(s) = \frac{1}{\beta(s)} \frac{T(s)}{1 + T(s)}$$

As we apply increasing amounts of negative feedback, the closed-loop plot moves to the left

This plot, showing the movement of the pole positions with feedback, is called a **root locus plot**.



# Effect of Feedback on Two-Pole System (1)

$$A_{OL}(s) = \frac{A_{OL,DC}}{(1 + s/\omega_{OL1})(1 + s/\omega_{OL2})} = \frac{A_{OL,DC}}{1 + s(1/\omega_{OL1} + 1/\omega_{OL2}) + s^2/\omega_{OL1}\omega_{OL2}}$$

$$= \frac{A_{OL,DC}}{1 + a_1s + a_2s^2} \text{ where } a_1 = (1/\omega_{OL1} + 1/\omega_{OL2}) \text{ and } a_2 = 1/\omega_{OL1}\omega_{OL2}$$

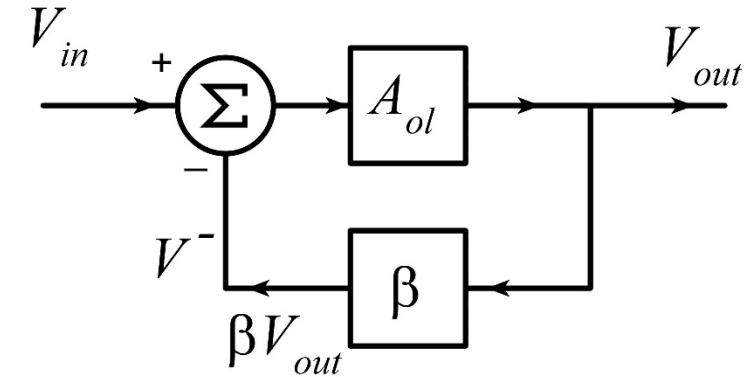
$$\beta(s) = \beta_0$$

$$T(s) = A_{OL}(s)\beta(s) = \frac{A_{OL,DC}\beta_0}{1 + a_1s + a_2s^2} = \frac{T_0}{1 + a_1s + a_2s^2} = \frac{N_T(s)}{D_T(s)}$$

$$A_{CL}(s) = \frac{1}{\beta(s)} \frac{T(s)}{1 + T(s)} = \frac{1}{\beta(s)} \frac{N_T/D_T}{1 + N_T/D_T} = \frac{1}{\beta(s)} \frac{N_T}{N_T + D_T}$$

$$A_{CL}(s) = \frac{1}{\beta_0} \frac{T_0}{T_0 + 1 + a_1s + a_2s^2} = \frac{1}{\beta_0} \frac{T_0}{T_0 + 1} \frac{1}{1 + \left(\frac{a_1}{1 + T_0}\right)s + \left(\frac{a_2}{1 + T_0}\right)s^2}$$

$$A_{CL}(s) = A_{CL,DC} \frac{1}{1 + s(2\zeta/\omega_n) + s^2/\omega_n^2}$$



# Effect of Feedback on Two-Pole System (2)

$$\frac{1}{\omega_n^2} = \left( \frac{a_2}{1+T_0} \right) \text{ so } \omega_n = \sqrt{\frac{1+T_0}{a_2}}$$

$$\frac{2\zeta}{\omega_n} = \left( \frac{a_1}{1+T_0} \right) \text{ so } \zeta = \left( \frac{a_1}{1+T_0} \right) \frac{\omega_n}{2} = \frac{1}{2} \left( \frac{a_1}{1+T_0} \right) \sqrt{\frac{1+T_0}{a_2}} = \frac{1}{2} \frac{a_1}{\sqrt{a_2}} \sqrt{\frac{1}{1+T_0}}$$

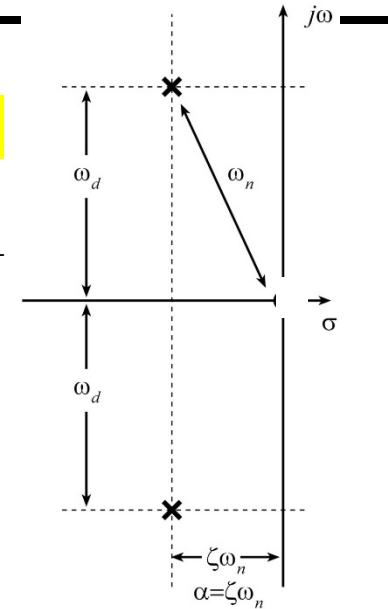
$$\omega_n = \sqrt{\frac{1+T_0}{a_2}} \text{ more feedback } \rightarrow \text{ higher natural resonant frequency}$$

$$\zeta = \frac{1}{2} \frac{a_1}{\sqrt{a_2}} \sqrt{\frac{1}{1+T_0}} \text{ more feedback } \rightarrow \text{ lower damping}$$

$$\zeta < 1$$

$$s_{p1,2} = -\zeta\omega_n \pm j\omega_d$$

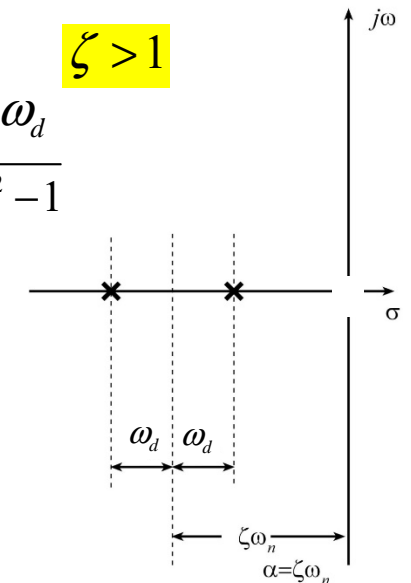
$$\text{where } \omega_d = \omega_n \cdot \sqrt{1-\zeta^2}$$



$$\zeta > 1$$

$$s_{p1,2} = -\zeta\omega_n \pm \omega_d$$

$$\text{where } \omega_d = \omega_n \cdot \sqrt{\zeta^2 - 1}$$



# Effect of Feedback on Two-Pole System (3)

$$\omega_n = \sqrt{\frac{1+T_0}{a_2}} = \sqrt{1+T_0} \cdot \sqrt{\omega_{OL1}\omega_{OL2}}$$

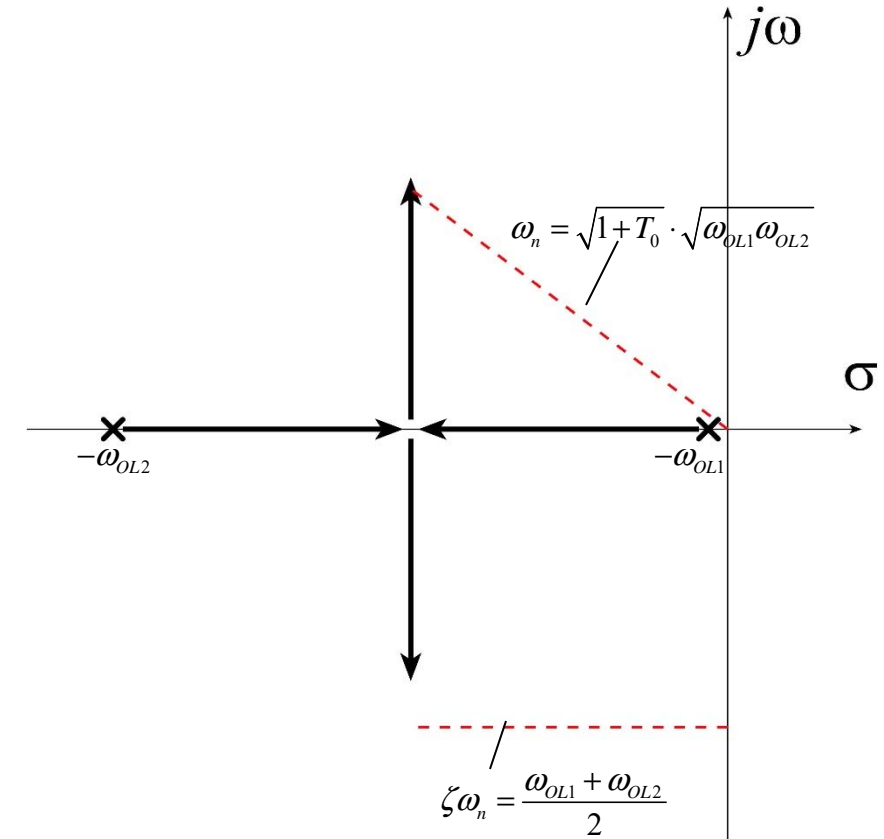
$$\zeta\omega_n = \frac{1}{2} \frac{a_1}{\sqrt{a_2}} \sqrt{\frac{1}{1+T_0}} \sqrt{\frac{1+T_0}{a_2}} = \frac{1}{2} \frac{a_1}{a_2} = \frac{1}{2} \frac{1/\omega_{OL1} + 1/\omega_{OL2}}{1/\omega_{OL1}\omega_{OL2}} = \frac{\omega_{OL1} + \omega_{OL2}}{2}$$

Here again is the root locus

Increasing feedback  $\rightarrow$  poles move towards each other

Eventually, they will meet.

Further increased feedback  $\rightarrow$  complex poles



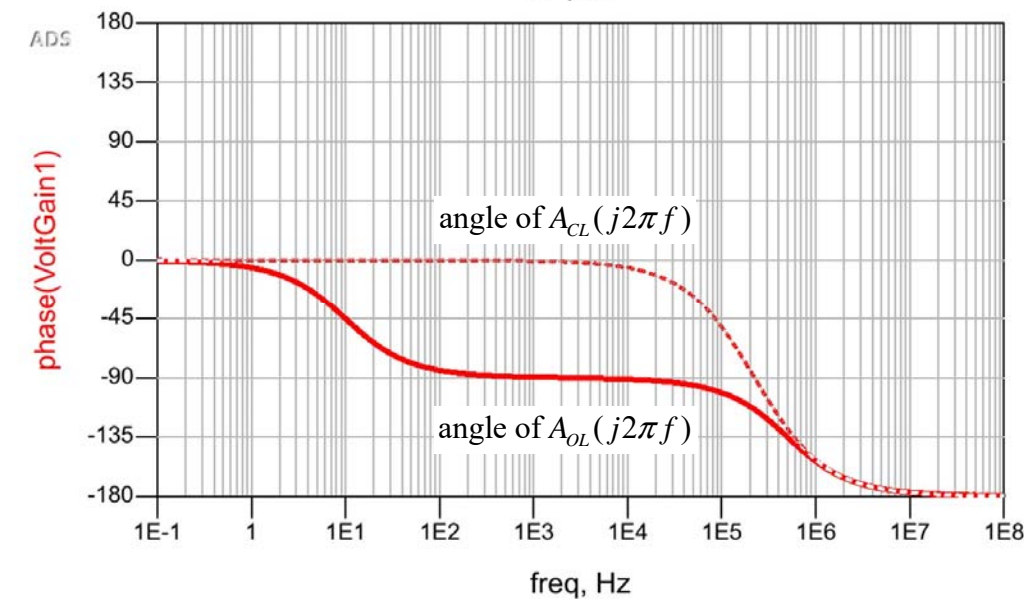
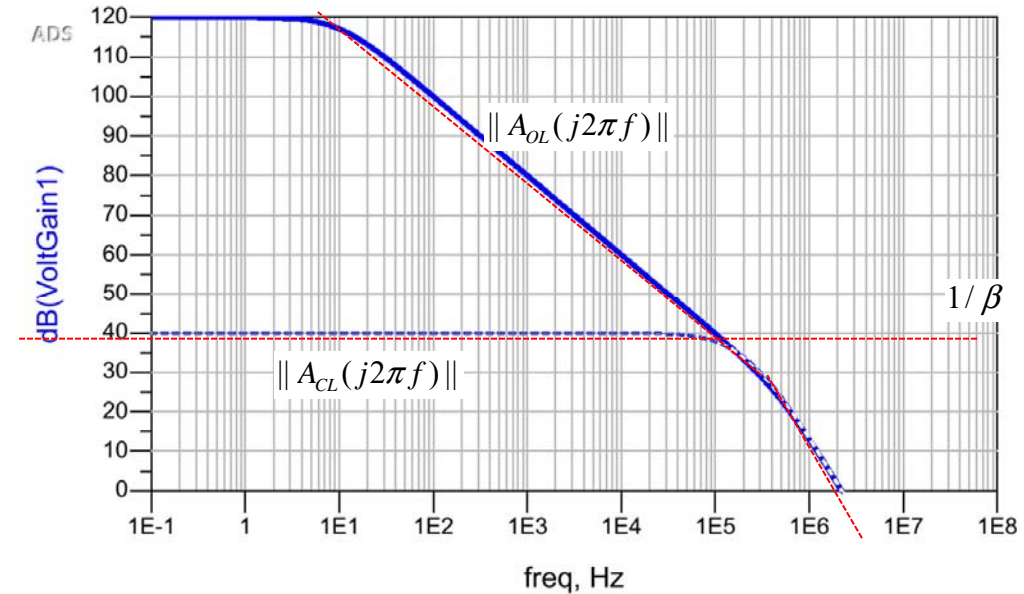
# Effect of Feedback on Two-Pole System (3)

Example:

$$A_{OL}(j2\pi f) = \frac{A_{OL,DC}}{(1 + jf / f_{OL1})(1 + jf / f_{OL2})}$$

$$= \frac{10^6}{(1 + jf / 10 \text{ Hz})(1 + jf / 500 \text{ kHz})}$$

$$\beta(j2\pi f) = \beta_0 = 1/100$$



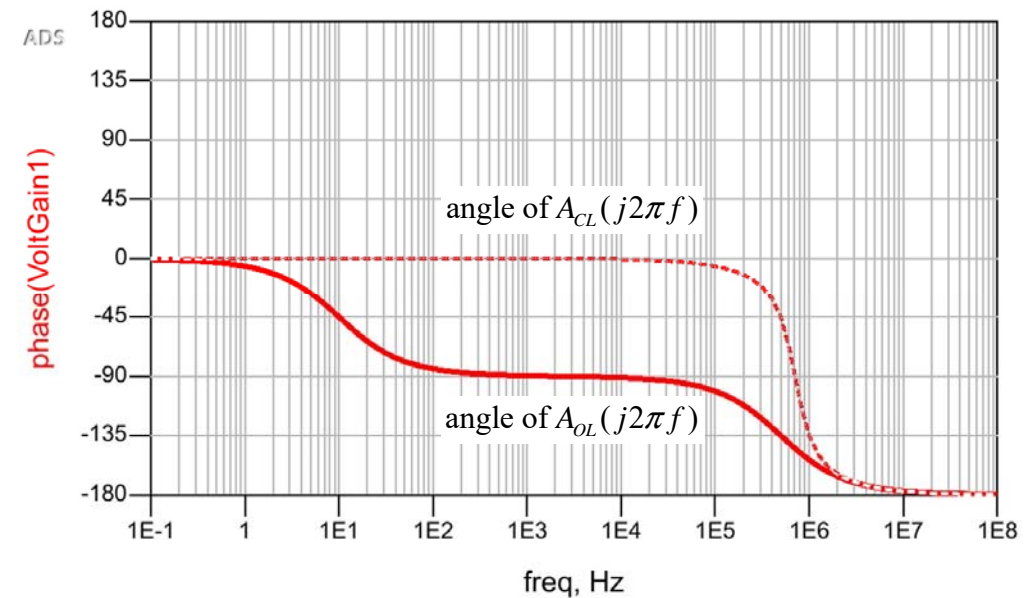
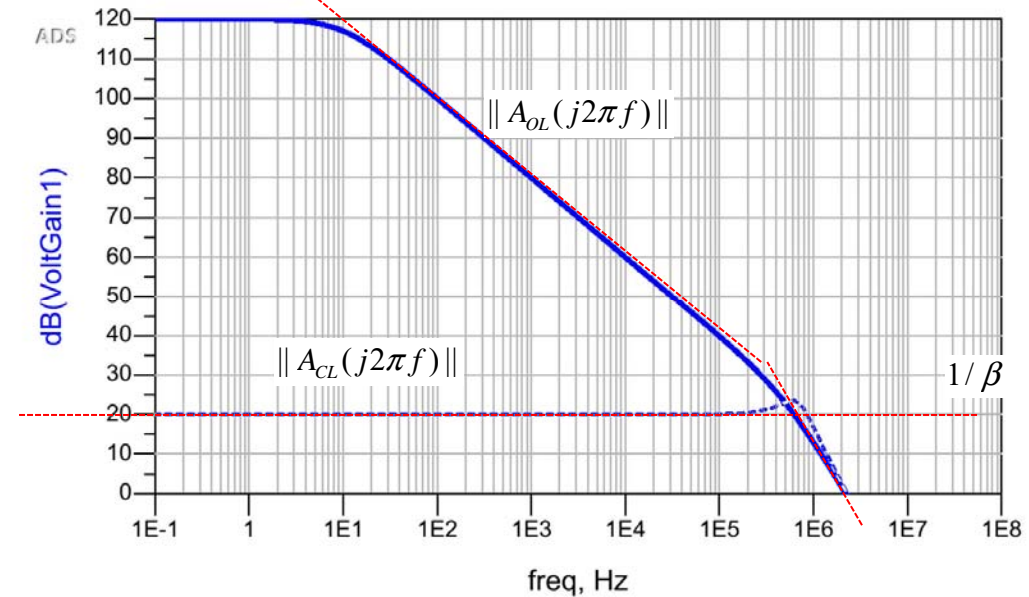
# Effect of Feedback on Two-Pole System (5)

Example:

$$A_{OL}(j2\pi f) = \frac{A_{OL,DC}}{(1 + jf / f_{OL1})(1 + jf / f_{OL2})}$$

$$= \frac{10^6}{(1 + jf / 10 \text{ Hz})(1 + jf / 500 \text{ kHz})}$$

$$\beta(j2\pi f) = \beta_0 = 1/10$$



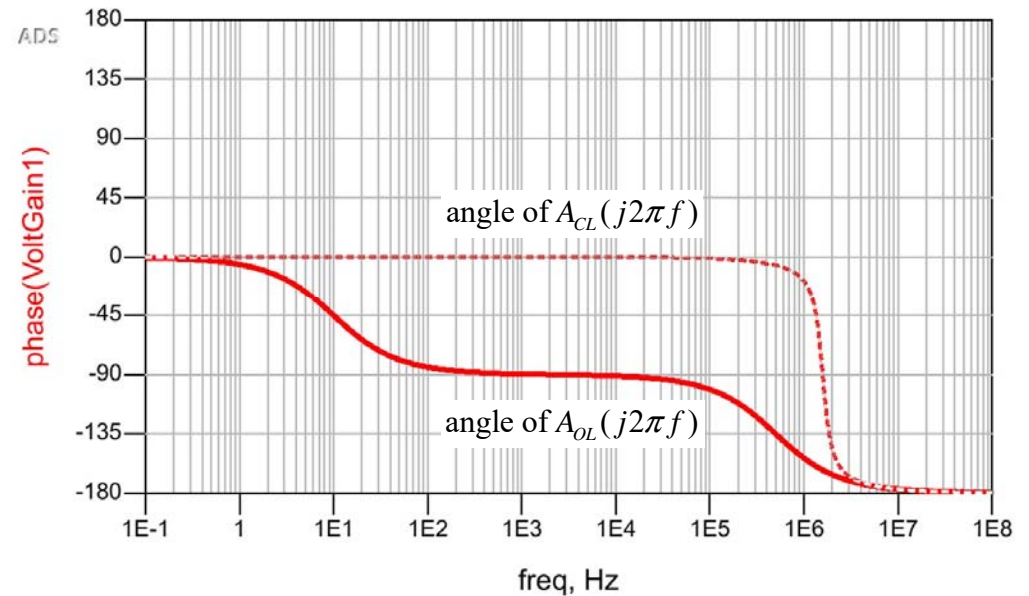
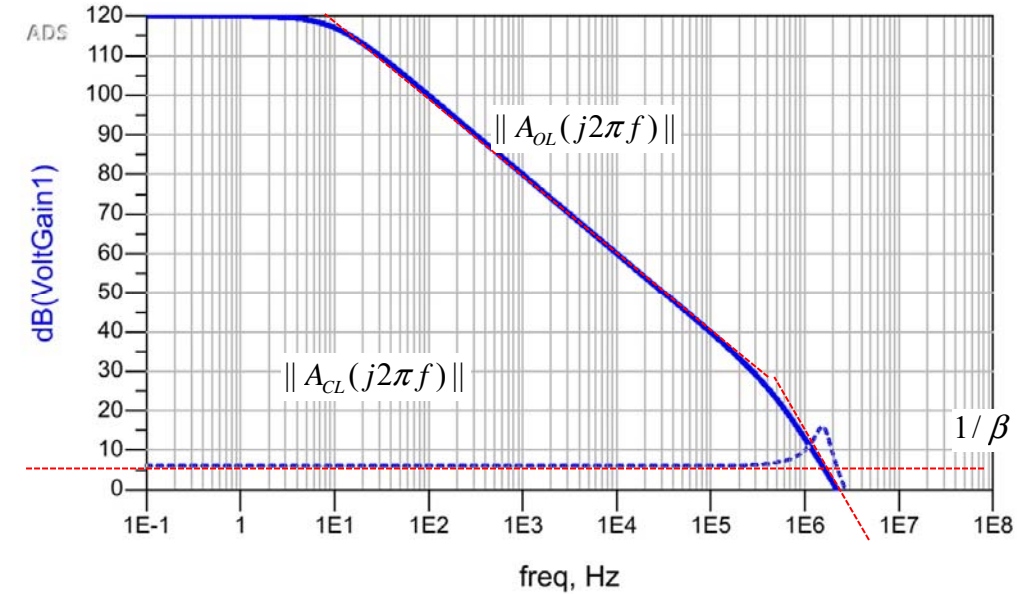
# Effect of Feedback on Two-Pole System (5)

Example:

$$A_{OL}(j2\pi f) = \frac{A_{OL,DC}}{(1 + jf / f_{OL1})(1 + jf / f_{OL2})}$$

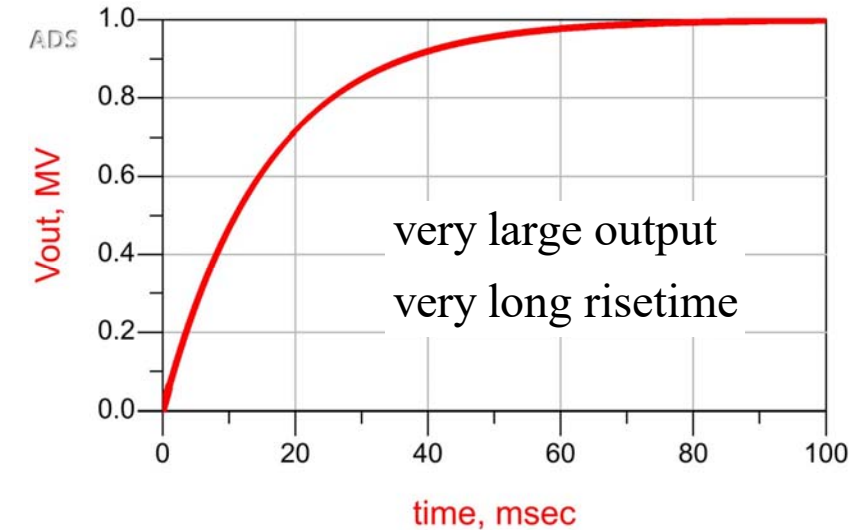
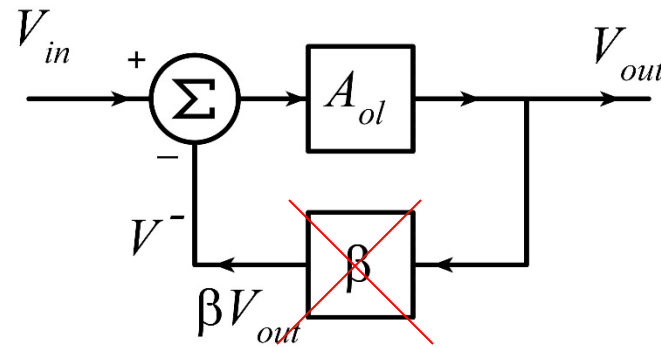
$$= \frac{10^6}{(1 + jf / 10 \text{ Hz})(1 + jf / 500 \text{ kHz})}$$

$$\beta(j2\pi f) = \beta_0 = 1/2$$

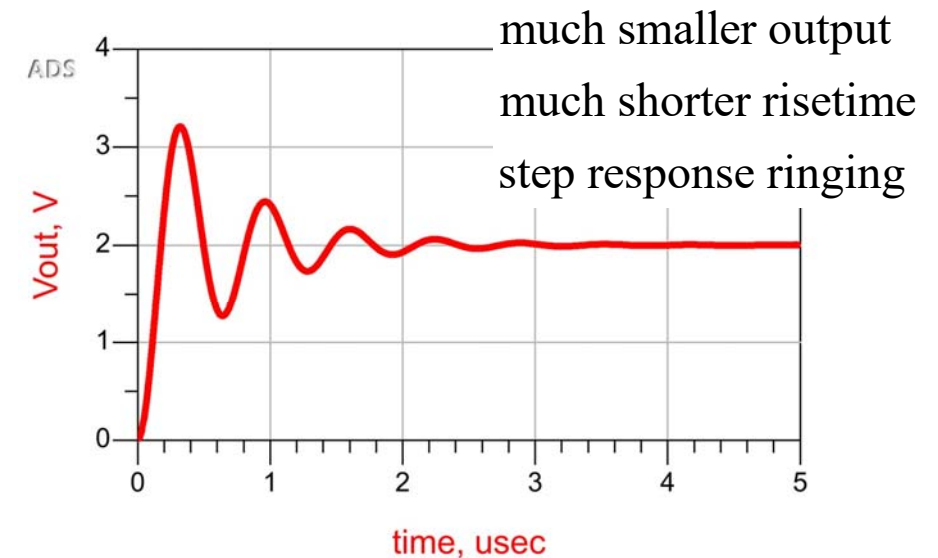
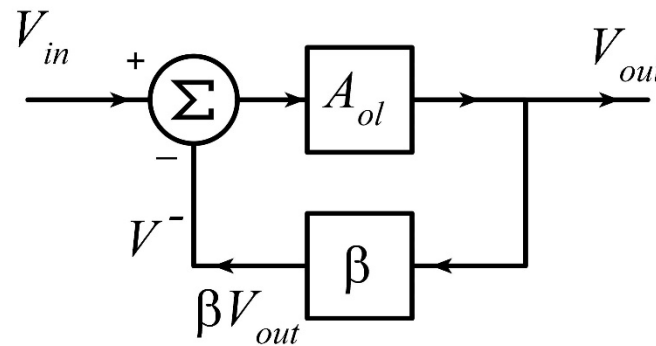


# Effect of Feedback on Two-Pole System (6)

Step response, no feedback



Step response, with feedback



$$A_{OL}(j2\pi f) = \frac{10^6}{(1 + jf / 10 \text{ Hz})(1 + jf / 500 \text{ kHz})}$$

$$\beta(j2\pi f) = 1/2$$



# Effect of Feedback on Three-Pole System (1)

$$A_{OL}(s) = \frac{A_{OL,DC}}{(1 + s / \omega_{OL1})(1 + s / \omega_{OL2})(1 + s / \omega_{OL3})}$$

$$= \frac{A_{OL,DC}}{1 + a_1s + a_2s^2 + a_3s^3} ; \text{ you can work out } a_1, a_2, \text{ and } a_3.$$

$$\beta(s) = \beta_0$$

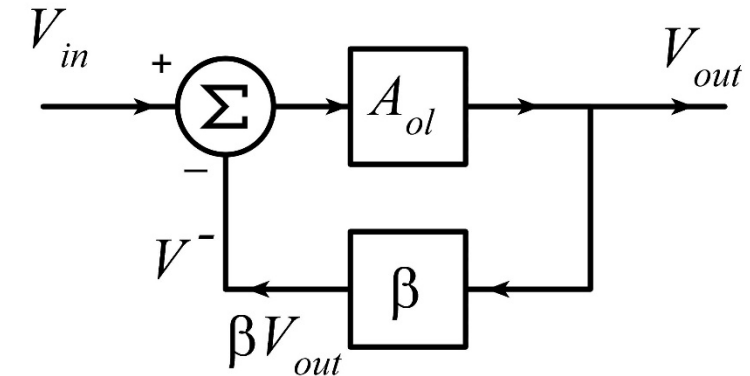
$$T(s) = A_{OL}(s)\beta(s) = \frac{T_0}{1 + a_1s + a_2s^2 + a_3s^3} = \frac{N_T(s)}{D_T(s)} \text{ where } T_0 = A_{OL,DC}\beta_0$$

$$A_{CL}(s) = \frac{1}{\beta(s)} \frac{T(s)}{1 + T(s)} = \frac{1}{\beta(s)} \frac{N_T / D_T}{1 + N_T / D_T} = \frac{1}{\beta(s)} \frac{N_T}{N_T + D_T}$$

$$A_{CL}(s) = \frac{1}{\beta_0} \frac{T_0}{T_0 + 1 + a_1s + a_2s^2 + a_3s^3}$$

$$A_{CL}(s) = \frac{1}{\beta_0} \frac{T_0}{T_0 + 1} \frac{1}{1 + \left(\frac{a_1}{1 + T_0}\right)s + \left(\frac{a_2}{1 + T_0}\right)s^2 + \left(\frac{a_3}{1 + T_0}\right)s^3}$$

Finding the roots (poles) of this cubic equation is hard work, but we can quickly make some key observations.



# Effect of Feedback on Three-Pole System (2)

$$\text{poles: } 1 + \left( \frac{a_1}{1+T_0} \right) s_p + \left( \frac{a_2}{1+T_0} \right) s_p^2 + \left( \frac{a_3}{1+T_0} \right) s_p^3 = 0$$

Suppose that  $(1+T_0)$  is very large.

Then  $|s_p|$  must be very large.

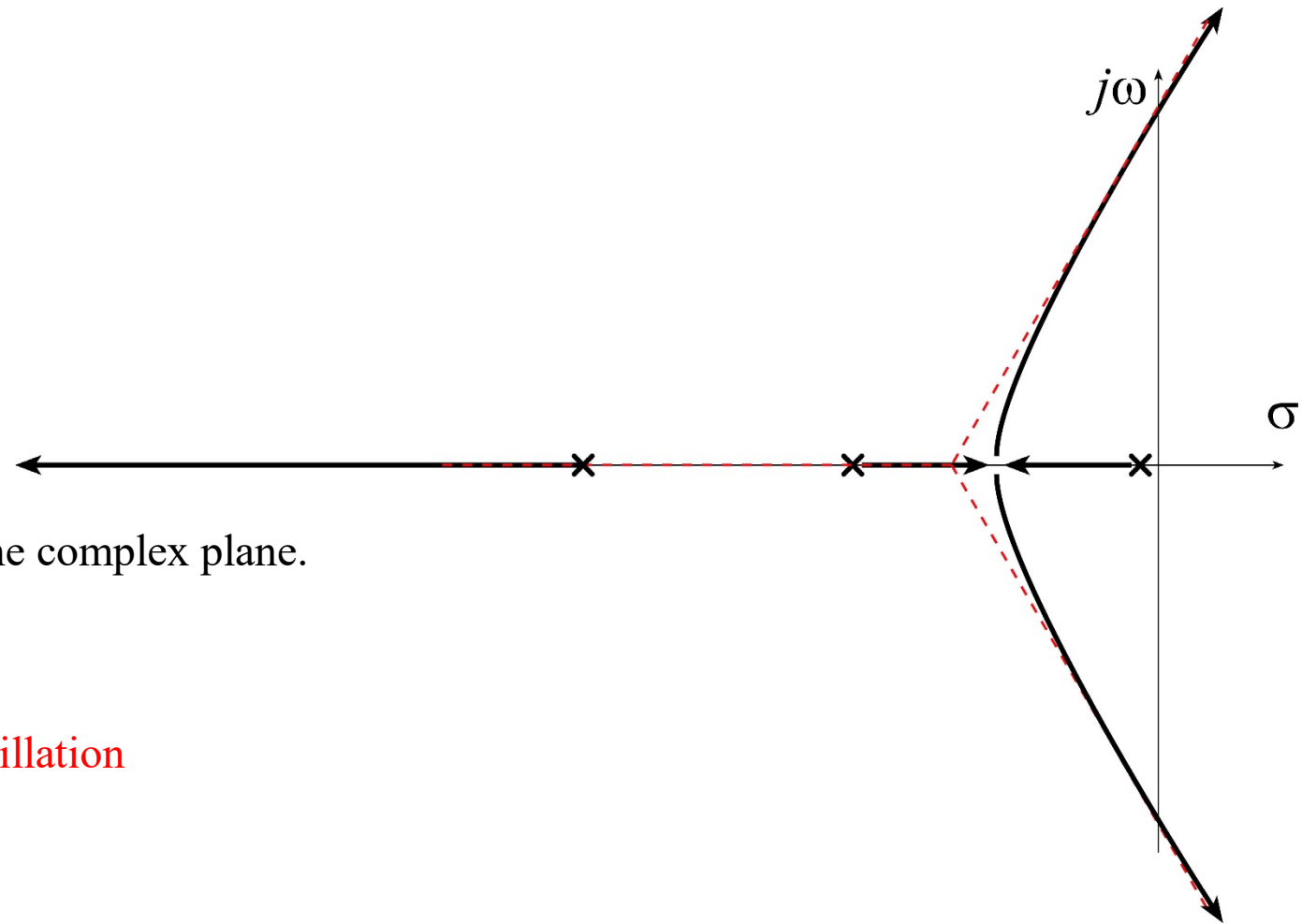
So  $|s_p^3| \gg |s_p^2| \gg |s_p^1|$

$$\text{so } s_p \approx \left( -\frac{1+T_0}{a_3} \right)^{1/3} \text{ if } \|T_0\| \gg 1$$

$(-1)^{1/3}$  has 3 roots, at angles of  $60^\circ$ ,  $-60^\circ$ , and  $180^\circ$  in the complex plane.

Root locus is therefore as sketched.

With large  $T_0$ , poles move into right half plane  $\rightarrow$  oscillation



# Comments

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Feedback is used in transistor circuits.

Feedback is used in far, far more things than transistor circuits.  
In general, the subject is called "control system theory".

There are 10-week undergraduate courses in control systems.

There's enough material for 4-8 Ph.D. -level courses in control systems.

We are just learning the basics.