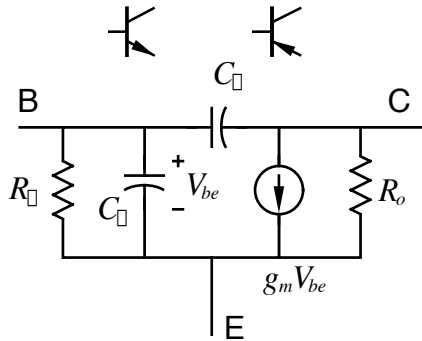


Basics: Transistor small-signal high-frequency models



Hybrid-pi model:

$$C_\mu = C_{bc}$$

$$C_\pi = C_{be} = C_{\pi,depl} + C_{\pi,diff}$$

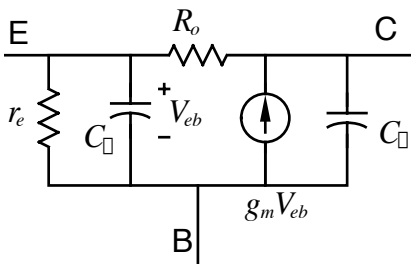
$$C_{\pi,diff} = g_m \tau_f$$

$$f_\tau = g_m / (2\pi(C_\pi + C_\mu))$$

$$g_m = \alpha / r_e = I_c / V_T$$

$$R_\pi = (\beta + 1)r_e$$

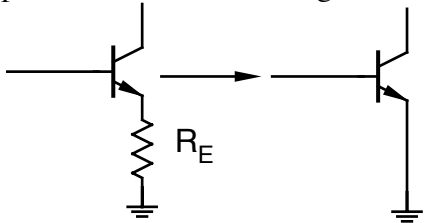
$$R_o = V_A / I_C$$



"T"-model

Makes common-base analysis much easier. Both these models are only approximate, being "good" up to f_t . Additionally $R_{b\oplus}$ is often an important parasitic (not discussed much in 137B).

Simplification of Emitter Degeneration



approximate only; check notes for bounds on validity

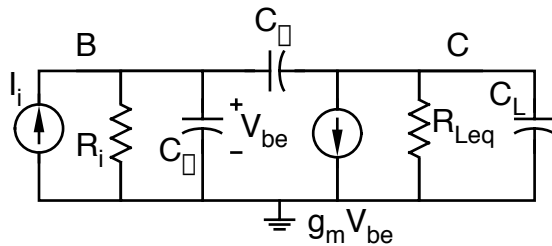
$$\mathcal{G}_\pi = C_\pi (r_e / (r_e + R_E))$$

$$\mathcal{g}_m = g_m (r_e / (r_e + R_E)) = \alpha / (r_e + R_E)$$

$$\mathcal{C}_\mu = C_\mu$$

$$\mathcal{R}_\pi = R_\pi ((r_e + R_E) / r_e) = (\beta + 1)(r_e + R_E)$$

Common-Emitter Stage



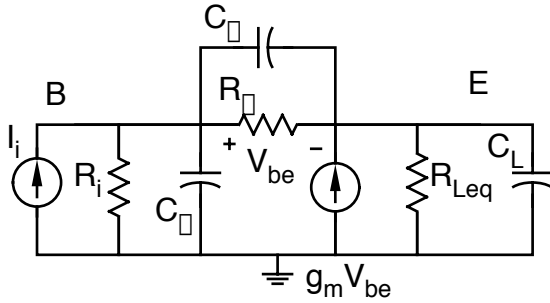
$$V_{out} / V_{gen} = (V_{out} / V_{gen})_{MB} \frac{1 + s\tau_{zero}}{1 + a_1 s + a_2 s^2}$$

$$a_1 = R_i (C_\pi + C_\mu (1 + g_m R_{Leq})) + R_{Leq} (C_\mu + C_L)$$

$$a_2 = R_i R_{Leq} (C_\mu C_L + C_\mu C_\pi + C_\pi C_L)$$

$$\tau_{zero} = -C_\mu / g_m$$

Emitter-Follower Stage



$$V_{out}/V_{gen} = (V_{out}/V_{gen})_{MB} \frac{1 + s\tau_{zero}}{1 + a_1s + a_2s^2}$$

given that $A_{vmb} = (r_e / (r_e + R_{Leq}))$:

$$a_1 = C_\pi (R_\pi \parallel (r_e \parallel R_{Leq} + R_i (1 - A_{vmb})))$$

+ $C_\mu (R_i \parallel \text{transistor input resistance})$

+ $C_L (R_{Leq} \parallel \text{transistor output resistance})$

$$a_2 = (R_i \parallel \text{transistor input resistance}) (R_{Leq} \parallel r_e)$$

$$\times (C_\mu C_\pi + C_\mu C_L + C_L C_\pi)$$

$$\tau_{zero} = g_m / C_\pi$$

General Solutions of Problems: Nodal analysis (Know how to do this!)

<p>1) Write the nodal equations (sum of the currents=0) at each circuit node, and put the resulting equations in matrix form (the Y's being various combinations of gm's, 1/R's, and sC's):</p> $\begin{bmatrix} Y_{11} & Y_{12} & Y_{13} & Y_{14} \\ Y_{21} & Y_{22} & Y_{23} & Y_{24} \\ Y_{31} & Y_{32} & Y_{33} & Y_{34} \\ Y_{41} & Y_{42} & Y_{43} & Y_{44} \end{bmatrix} \begin{bmatrix} V_1 = V_{in} \\ V_2 \\ V_3 \\ V_4 = V_{out} \end{bmatrix} = \begin{bmatrix} I_{in} \\ 0 \\ 0 \\ 0 \end{bmatrix}$	<p>2) Use Cramer's rule to solve:</p> $\begin{bmatrix} Y_{11} & Y_{12} & Y_{13} & I_{in} \\ Y_{21} & Y_{22} & Y_{23} & 0 \\ Y_{31} & Y_{32} & Y_{33} & 0 \\ Y_{41} & Y_{42} & Y_{43} & 0 \end{bmatrix} = V_{out}$
<p>3) This comes out as:</p> $\frac{V_{out}}{I_{in}} = ks^m \frac{c_0 + c_1s + c_2s^2 + \dots}{d_0 + d_1s + d_2s^2 + \dots}$ <p>,which is divided through to get:</p> <p>(if present, m is the number of zeros, minus the number of poles, in the transfer function)</p>	<p>4)</p> $\frac{V_{out}}{I_{in}} = \left(\frac{V_{out}}{I_{in}} \right)_{mb} s^m \frac{1 + b_1s + b_2s^2 + \dots}{1 + a_1s + a_2s^2 + \dots}$ <p>...and the poles and zeroes are found by factoring the numerator and denominator. The separated-pole approximation, if applicable, makes this factoring easy.</p>
<p>5) To find the <i>impulse response</i>, do a partial-fraction expansion and then take the inverse LaPlace transform.</p>	<p>6) To find the sinusoidal frequency response, set $s = j\omega$.</p>

General Solutions of Problems: Method of Time Constants

<p>Circuit without Capacitors or Inductors</p>	<p>Circuit without Capacitors or Inductors</p>
<p>Circuit without Capacitors or Inductors</p>	$\frac{V_{out}}{V_{gen}} = \left(\frac{V_{out}}{V_{gen}} \right)_{mb} \frac{1 + b_1 s + b_2 s^2 + \dots}{1 + a_1 s + a_2 s^2 + \dots}$ $a_1 = R_{11}^0 C_1 + R_{22}^0 C_2 + R_{33}^0 C_3 + R_{44}^0 C_4$ $a_2 = R_{11}^0 R_{22}^1 C_1 C_2 + R_{11}^0 R_{33}^1 C_1 C_3 + R_{11}^0 R_{44}^1 C_1 C_4 + R_{22}^0 R_{33}^2 C_2 C_3 + R_{22}^0 R_{44}^2 C_2 C_4 + R_{33}^0 R_{44}^3 C_3 C_4$ <p>notethat $R_{xx}^0 R_{yy}^x = R_{xx}^y R_{yy}^0$</p>
	$R_{xx}^0 = R_{\pi} \parallel \left(r_e \parallel R_{Leq} + R_i (1 - A_{vmb}) \right)$ $A_{vmb} = \left(R_{Leq} / (r_e + R_{Leq}) \right)$
	$R_i = R_x \parallel R_{\pi}$ $R_{yy}^0 = R_i (1 + g_m R_{Leq}) + R_{Leq}$