



# WSB: 100-300GHz systems: Architectures and Applications

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Acknowledgement: This work was supported in part by the Semiconductor Research Corporation (SRC) and DARPA.






## Systems

-  **Sundeep Rangan**  
UC Berkeley  
Networks, Applications, MIMO, Power
-  **Upamanyu Madhow**  
UC Santa Barbara  
MIMO algorithms  
Imaging algorithms  
Compressive imaging
-  **Christoph Studer**  
Cornell  
MIMO algorithms  
VLSI MIMO  
digital beamforming
-  **Andreas Molisch**  
USC  
100-300GHz  
propagation  
measurements
-  **Danijela Cabric**  
UCLA  
MIMO algorithms  
(funding via CONIX)

## ICs

-  **Ali Niknejad**  
UC Berkeley  
mm-wave CMOS: hub  
mm-wave arrays  
mm-wave MIMO
-  **James Buckwalter**  
UC Santa Barbara  
efficient PAs  
III-V arrays
-  **Kenneth O**  
UT Dallas  
140-300GHz  
SiGe ICs
-  **Muhannad Bakir**  
Georgia Tech  
high-frequency  
packaging
-  **Gabriel Rebeiz**  
UC San Diego  
mm-wave CMOS: handset  
mm-wave arrays
-  **Alyosha Molnar**  
Cornell  
N-path mixers  
MIMO ADCs
-  **Elad Alon**  
UC Berkeley  
design automation  
equalizers
-  **Tim Fisher**  
UCLA  
advanced  
packaging  
materials
-  **Andrew Kummel**  
UCSD  
advanced  
packaging  
materials

## Transistors

-  **Umesh Mishra**  
UC Santa Barbara  
N-polar GaN HEMTs  
for 140, 210GHz
-  **Huili (Grace) Xing**  
Cornell  
AlN/GaN HEMTs  
for 140, 210GHz
-  **Susanne Stemmer**  
UC Santa Barbara  
transistors in  
novel materials
-  **Debdeep Jena**  
Cornell  
GaN HEMTs  
on Si
-  **Srabanti Chowdhury**  
UC Davis  
Diamond cooling  
for GaN

Massive MIMO demo.  **Borivoje Nikolic**  
UC Berkeley  
VLSI design automation  
VLSI MIMO processors

Compressive imaging  **Amin Arbabian**  
Stanford  
140GHz radar chipsets  
and arrays

140/210/280GHz arrays for demos.  **Mark Rodwell**  
UC Santa Barbara  
THz HBTs for PAs  
THz HEMTs for LNAs

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# 100-300GHz Wireless

Wireless networks: exploding demand.

Immediate industry response: 5G.

~10~40 GHz ("5G")

~40~100GHz ("5.5G ?")

increased spectrum, extensive beamforming

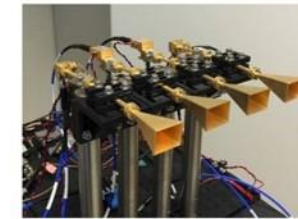
Next generation (6G ??): above 100GHz.. (?)

greatly increased spectrum, massive spatial multiplexing

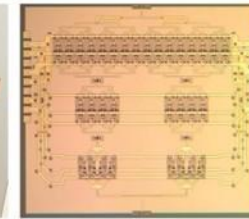
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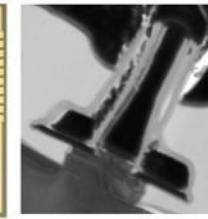
— Systems —



— ICs —

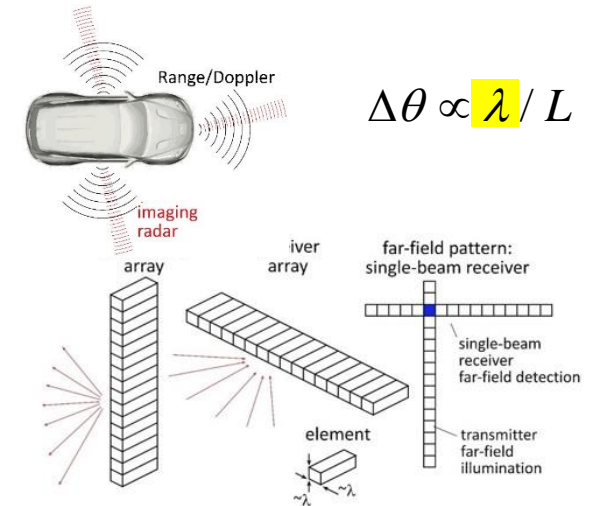
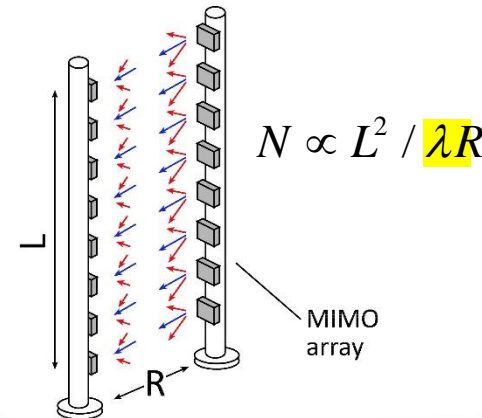
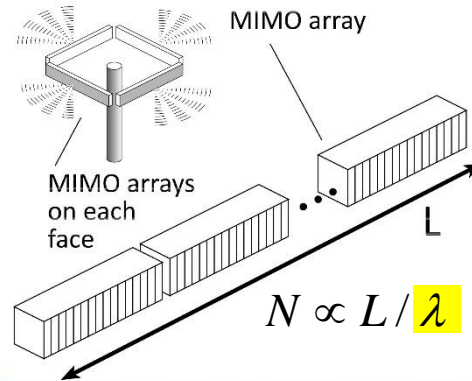
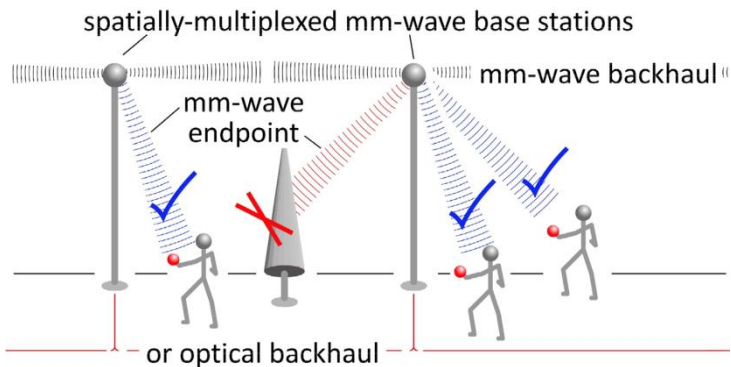


— Devices —



100-300GHz carriers, massive spatial multiplexing

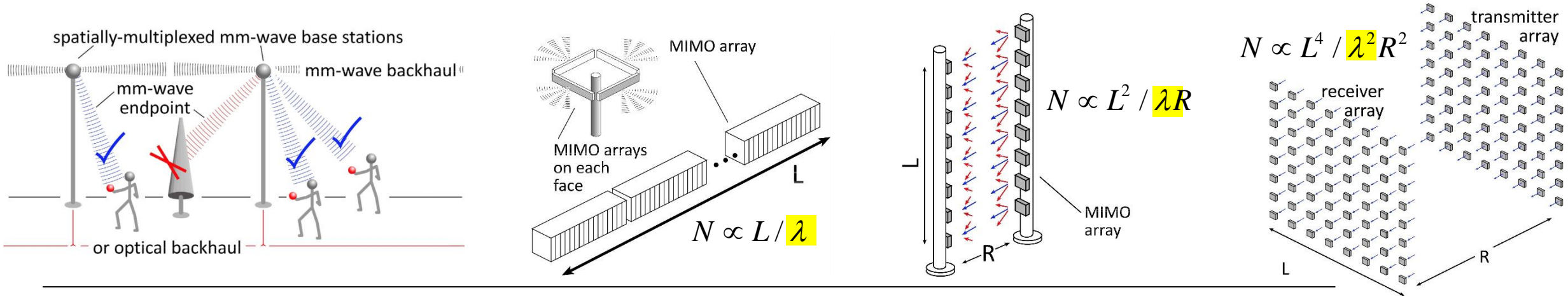
→ Terabit hubs and backhaul links, high-resolution imaging radar



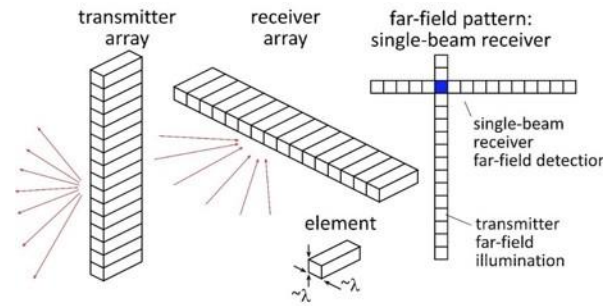
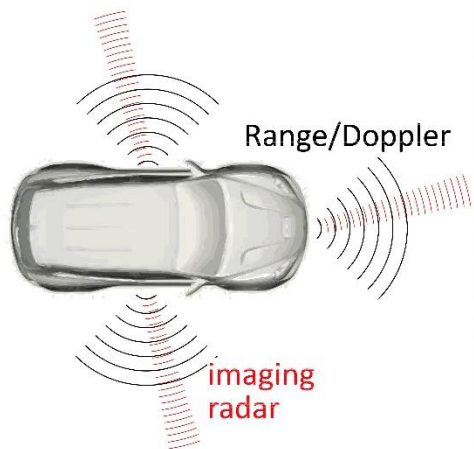
$$\Delta\theta \propto \lambda / L$$

# Benefits of Short Wavelengths

**Communications:** Massive spatial multiplexing, massive # of parallel channels. **Also, more spectrum!**



**Imaging:** very fine angular resolution



$$\Delta\theta \propto \lambda / L$$

**But:**

High losses in foul or humid weather.  
High  $\lambda^2/R^2$  path losses.  
ICs: poorer PAs & LNAs.  
Beams easily blocked.

**100-340GHz wireless:  
terabit capacity,  
short range,  
highly intermittent**

## Wireless above 100 GHz:

### Massive capacities

large available bandwidths

massive spatial multiplexing in base stations and point-point links

### Very short range: few 100 meters

short wavelength, high atmospheric losses. Easily-blocked beams.

### IC Technology

All-CMOS for short ranges below 200 GHz.

SiGe, GaN, or III-V LNAs and PAs for longer-range links. Just like cell phones today

SiGe or III-V frequency extenders for 220GHz and beyond

### The challenges

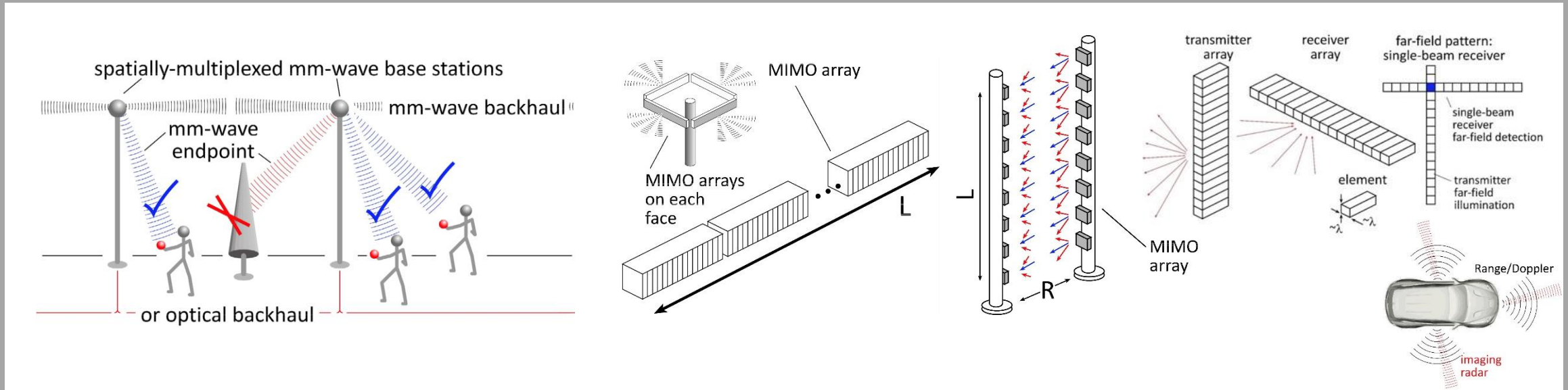
digital beamformer computational complexity

packaging: fitting signal channels in very small areas

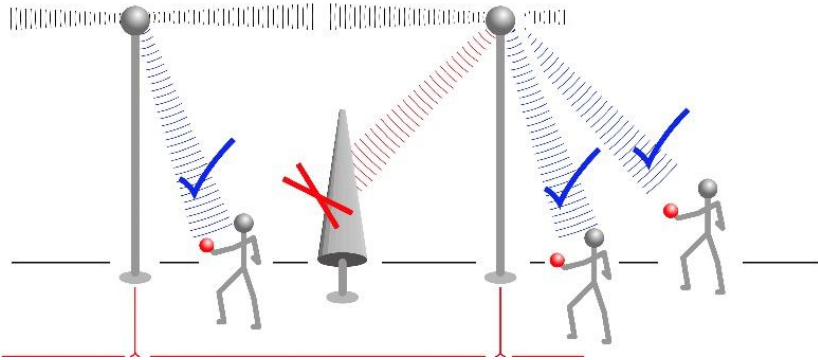
mesh networking to accommodate beam blockage

driving the technologies to low cost

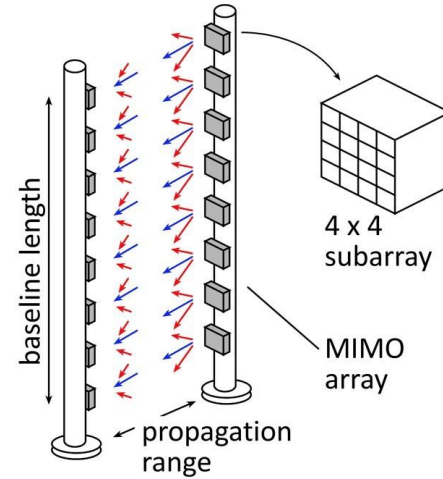
# Applications



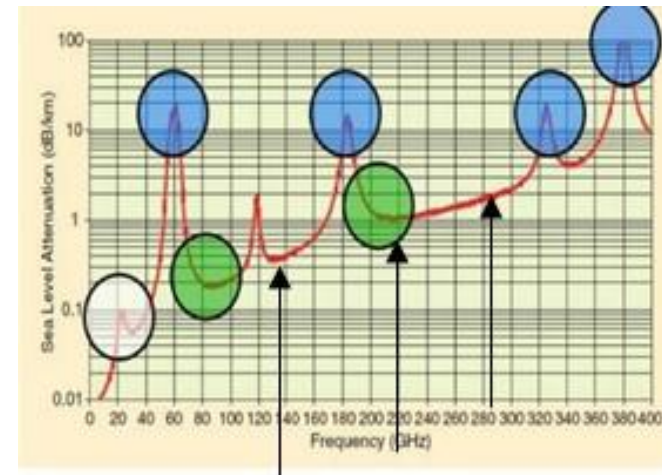
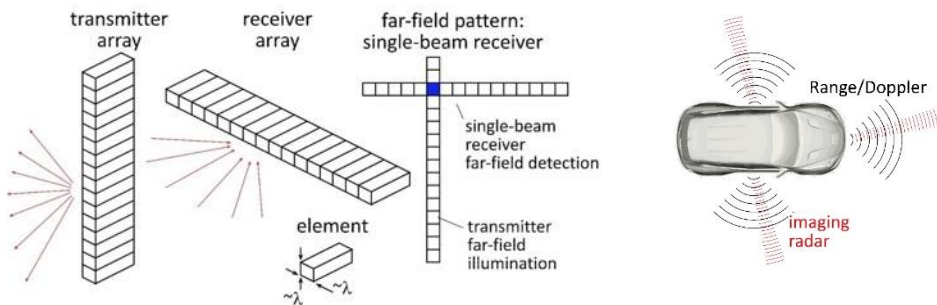
## 140GHz MIMO Hub



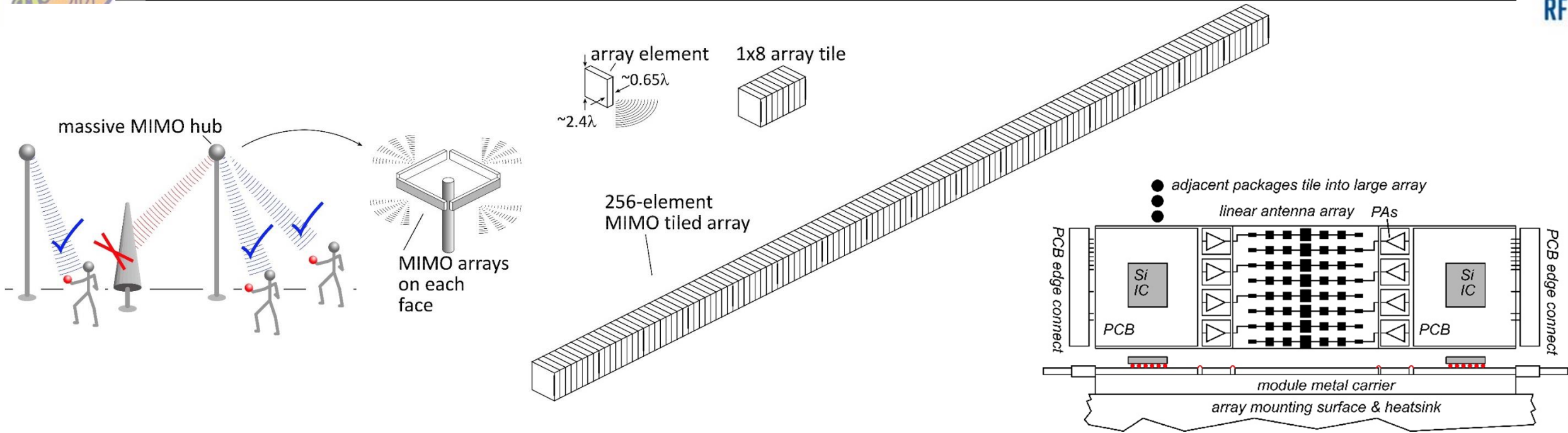
## 210 or 280GHz MIMO Backhaul



## 140 or 210GHz Imaging Radar





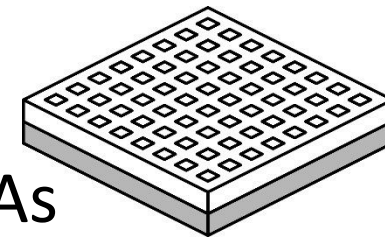


## 0.5-5 Tb/s spatially-multiplexed 140GHz base station

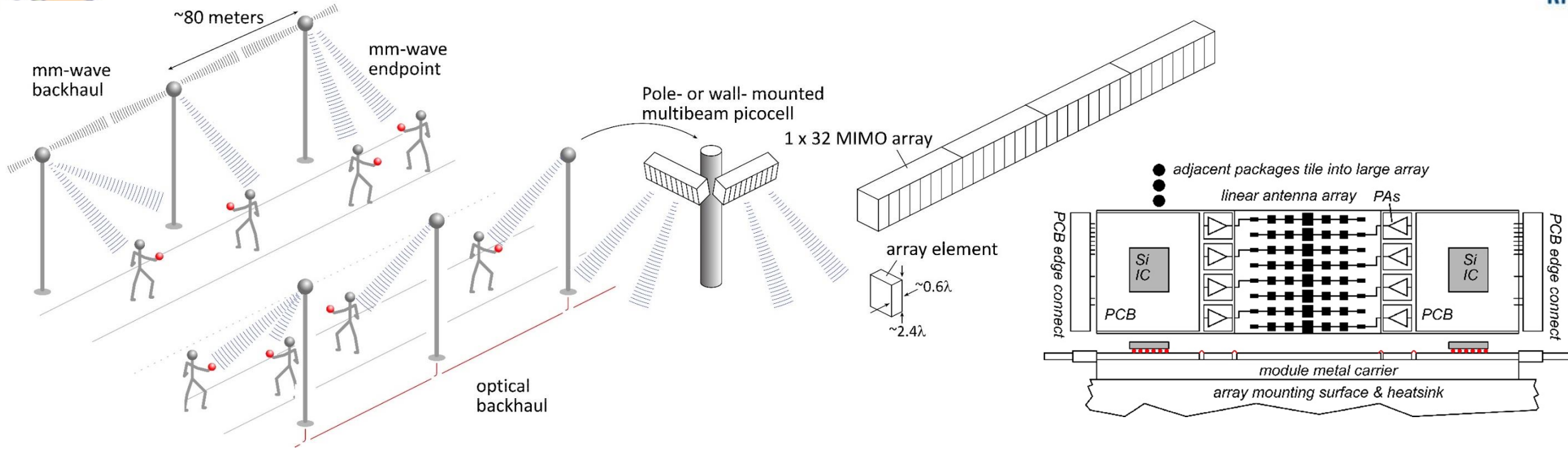
128 users/face, 4 faces.  $P_{1dB} = 21 \text{ dB}_m$  PAs,  $F = 8 \text{ dB}$  LNAs

512 total users @ 1 user/beam, 1, 10 Gb/s/beam;

230, 100 m range in 50mm/hr rain with 17dB total margins



Handset:  
8 × 8 array  
(9×9mm)

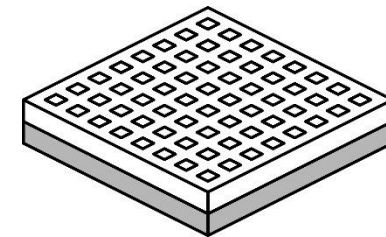


**If demo uses 32-element array (four 1×8 modules):**

16 users/array.  $P_{1dB} = 21 \text{ dB}_m$  PAs,  $F = 8 \text{ dB}$  LNAs

1,10 Gb/s/beam → 16, 160 Gb/s total capacity

70, 40 m range in 50mm/hr rain with 17dB total margins



Handset:  
8 × 8 array  
(9×9mm)

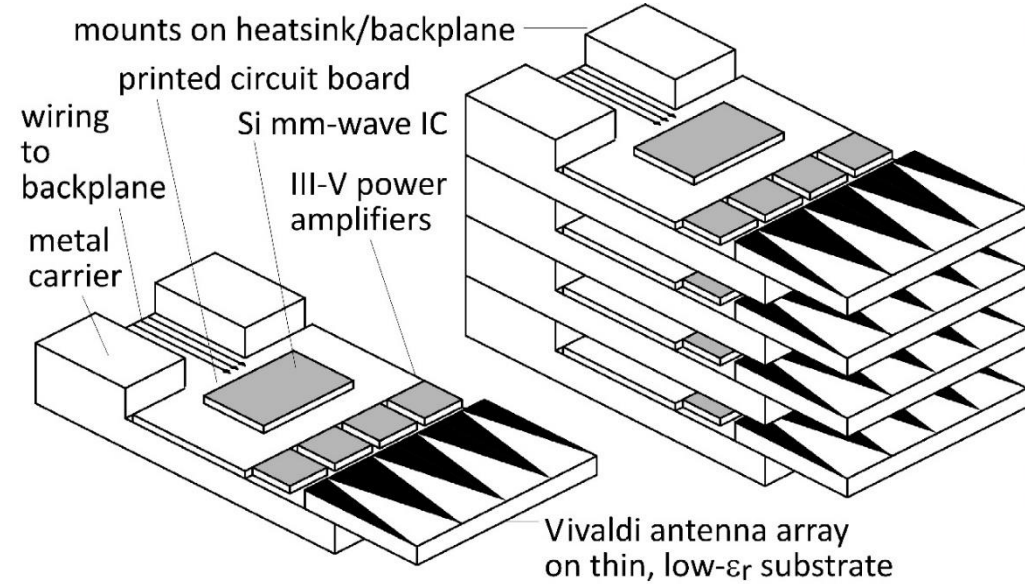
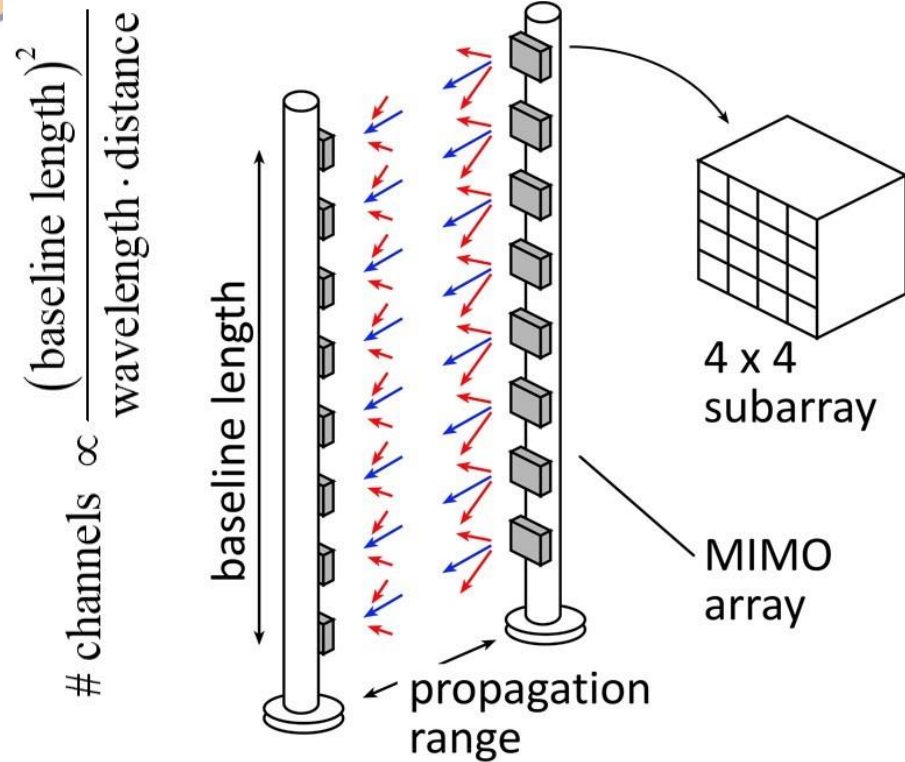
Range varies as  $(\# \text{ hub elements})^{0.5} \rightarrow (\text{Service area/element})$  is constant

If we use instead a 70GHz carrier,  
the range increases to **168 meters** (vs. **100 meters**)  
but the handset becomes 16mm×16mm (vs. 8mm×8mm),  
and the hub array becomes 20mm×524mm (vs. 10mm×262mm)

Or, use a 4×4 (8mm×8mm) handset array,  
and the range becomes **..100 meters**.

Same handset area (more handset elements)→ same link budget  
Easier to obtain license for 140±2.5GHz than 75±2.5GHz

# 220 GHz, 640 Gb/s MIMO Backhaul



## 8-element MIMO array

3.1 m baseline.

80Gb/s/subarray → 640Gb/s total

4 x 4 sub-arrays → 8 degree beamsteering

## Key link parameters

500 meters range in 50 mm/hr rain; 23 dB/km

24 dB total margins:

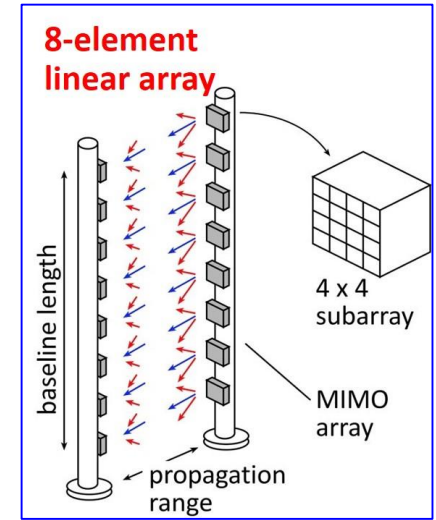
packaging loss, obstruction, operating, design, aging

PAs: 24mW  $P_{out}$  (per element)

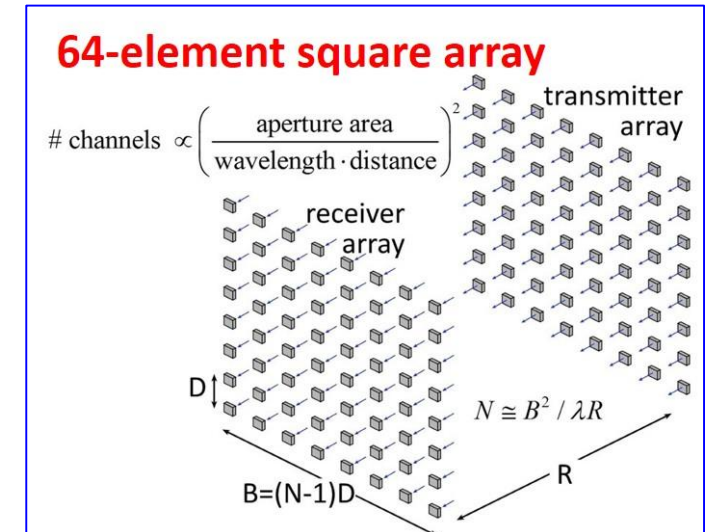
LNAs: 6dB noise figure

500m range in 50mm/hr. rain.

**8-element 640Gb/s linear array:**  
 requires  $14\text{dB}_m$  transmit power/element ( $P_{\text{out}}$ )  
 .... $3.2\text{W}$  total output power  
 requires  $2.1\text{m}$  linear array



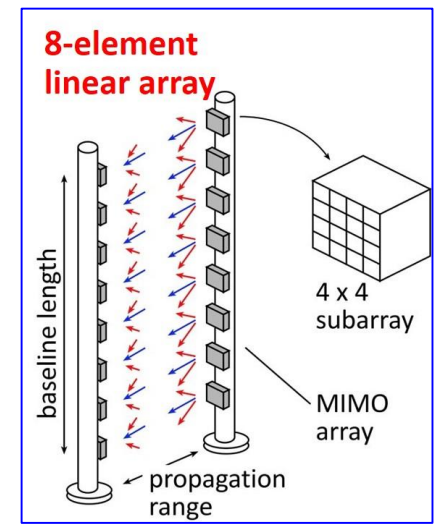
**64-element 5Tb/s square array:**  
 same link assumptions  
 requires  $5\text{dB}_m$  transmit power/element ( $P_{\text{out}}$ )  
 .... $3.2\text{W}$  total output power  
 requires  $2.1\text{m}$  square array



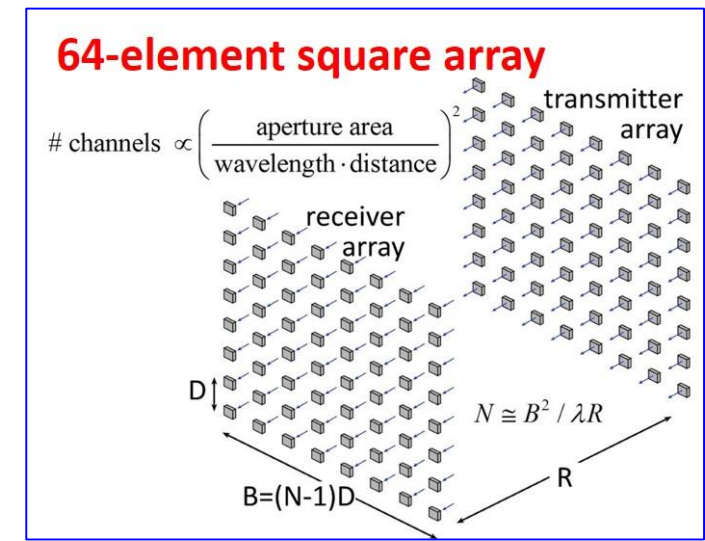
**Complex system: can we make it cheaply ?**

Why not use a lower-frequency carrier, e.g. 70 GHz ?

**8-element 640Gb/s linear array:**  
 requires  $11\text{dB}_m$  transmit power/element ( $P_{\text{out}}$ )  
 .... $1.7\text{W}$  total output power  
 requires  $5.5\text{m}$  linear array

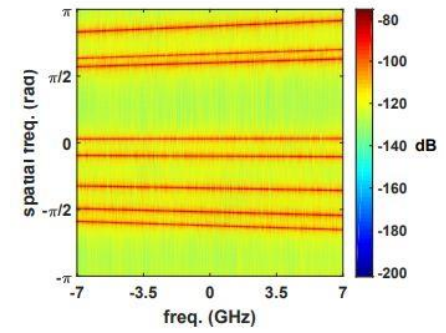
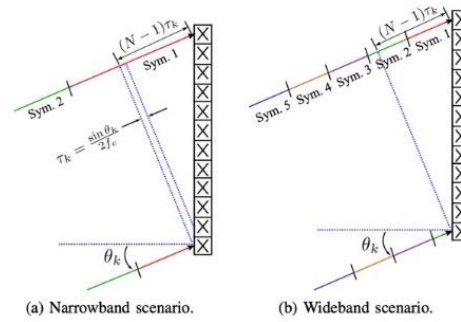
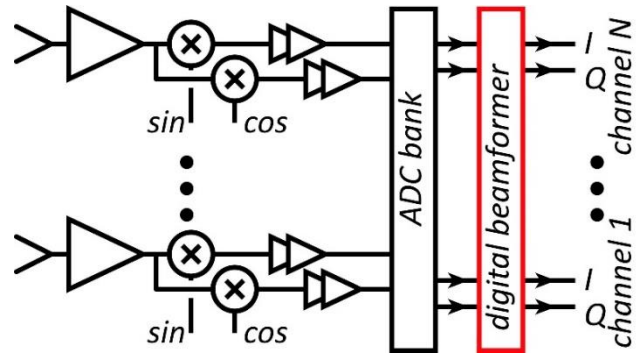


**64-element 5Tb/s square array:**  
 same link assumptions  
 requires  $2\text{dB}_m$  transmit power/element ( $P_{\text{out}}$ )  
 .... $1.7\text{W}$  total output power  
 requires  $5.5\text{m}$  square array



Similar RF power output, physically larger

# Systems



**ADCs/DACs:** only 3-4 bit ADC/DACs required (Madhow, Studer, Rodwell)

**Linearity:** Amplifier  $P_{1dB}$  need be only 3dB above average power (Madhow).

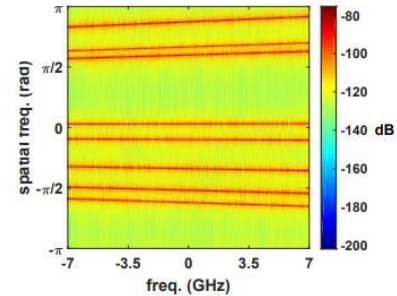
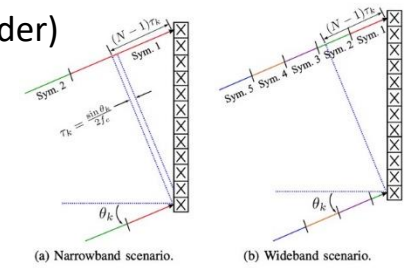
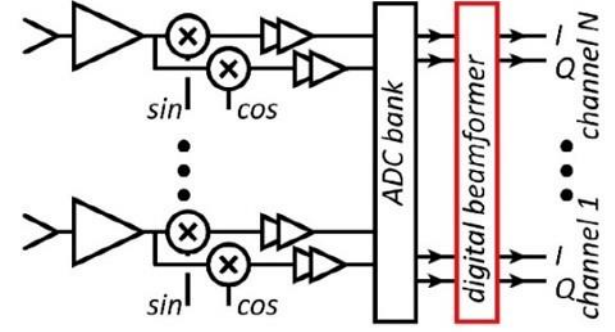
**Phase noise:** Requirements same as for SISO (Alon, Madhow, Niknejad, Rodwell)

**Efficient digital beamforming:** beamspace algorithm=complexity  $\sim N \times \log(N)$  (Madhow, Studer)

**Efficient digital beamforming:** low-resolution matrix (Studer)

**Efficient channel estimation :** fast beamspace algorithm (Studer)

**Efficiently addressing true-time-delay problem:** "rainbow" FFT algorithm (Madhow, Cabric, Studer)





## ADC resolution:

$N$  ADC bits,  $M$  antennas,  $K$  signals:  $SNR = 6N + 1.76 + 10 \cdot \log_{10}(M/K)$   
 3 bits,  $(M/K) = 2 \rightarrow SNR = 23$  dB. QPSK needs 9.8 dB.

## Jammer tolerance:

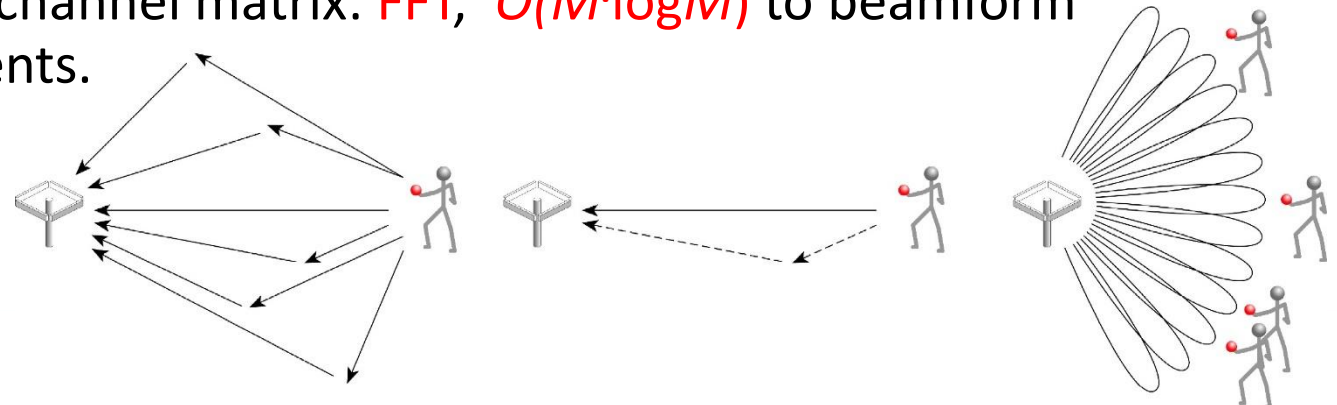
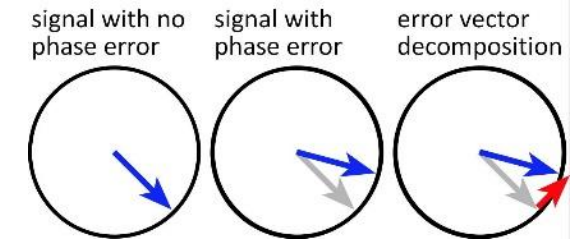
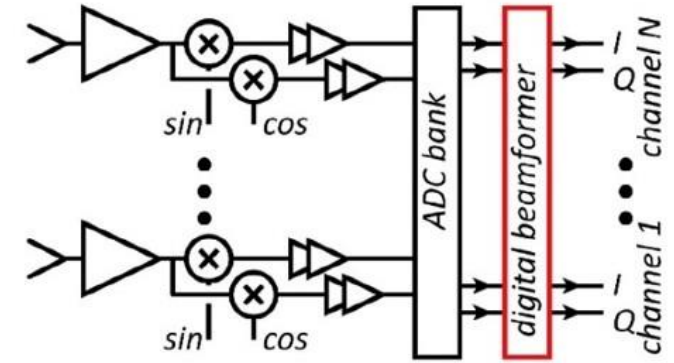
Increase ADC resolution by 1 bit  $\rightarrow P_{jammer,max} = K \cdot P_{signal}$   
 Maximum jammer power = sum of all user's power.

## Phase noise:

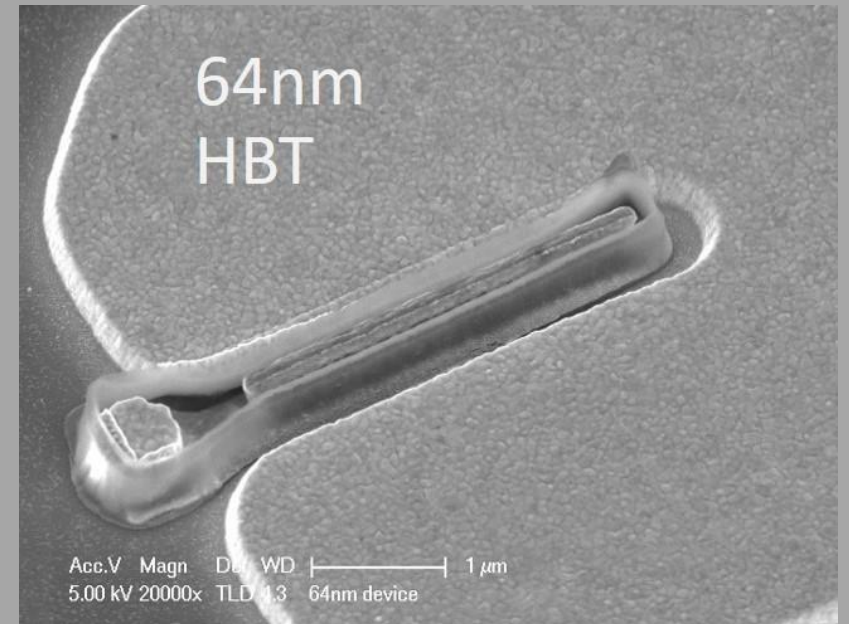
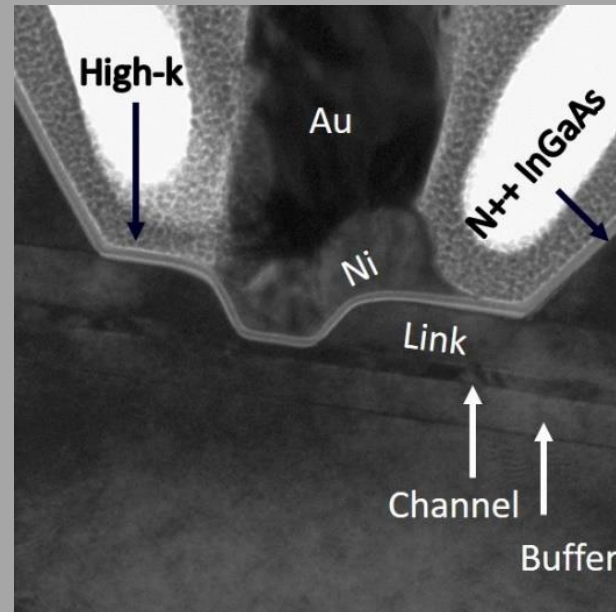
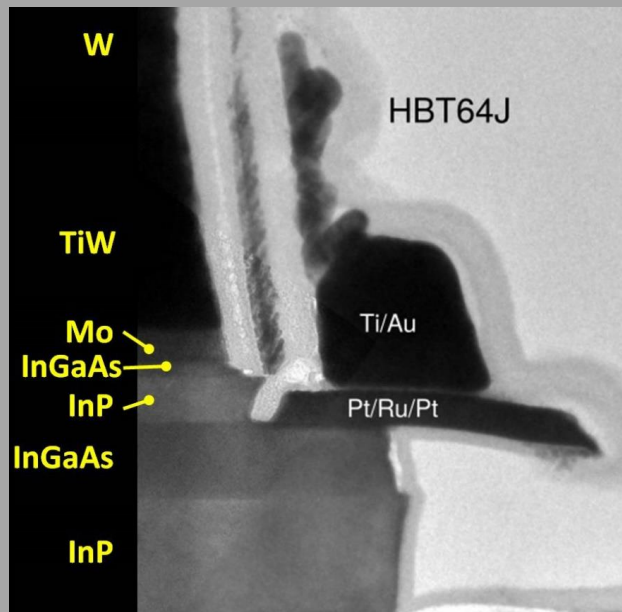
Phase error  $\sigma_\phi$ :  $SNR = -20 \cdot \log_{10}(\sigma_\phi) + 10 \cdot \log_{10}(M/K)$ , where  $\sigma_\phi^2 = \int_{f_{low}}^{f_{symbol}/2} L(f) df$ .  
 MIMO and SISO require similar  $L(f)$ .

## Beamspace:

lower frequencies, many NLOS paths, complicated channel matrix:  $O(M^3)$  to beamform  
 higher frequencies, few NLOS paths, simpler channel matrix: FFT,  $O(M \cdot \log M)$  to beamform  
 fewer bits in signal; fewer bits in FFT coefficients.



# Transistors



	$f_{\max}$ GHz	Good ICs to (GHz)	complexity	LNAs	PAS	increased bandwidth ?
CMOS	350	150/200	transceivers	good	weak: 10-30 mW	not easy
Production SiGe	300	200/250	transceivers	ok	OK: 20-100 mW	depends on \$\$
R&D SiGe	700	300/500	transceivers	good	OK: 20-100 mW	2-3THz
R&D InP HBT	1150	400/650	PA, converters	ok*	good: 100-200 mW	2-3THz
R&D InP HEMT	1500	500/1000	LNA	great	weak: 20-50 mW	2-3THz
R&D GaN	400	120/140	PAs	good	excellent: 0.1-1W	600GHz

ICs with useful performance, hero experiments

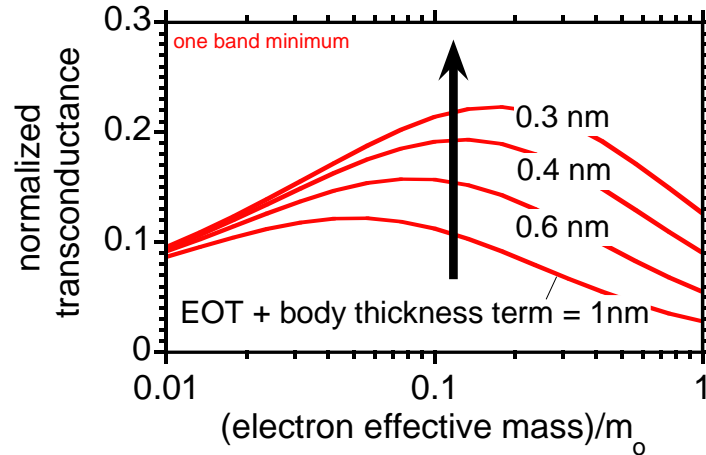
\*can be addressed

There are **THz transistors today**; their bandwidth will **increase**

**Challenge: reducing costs, increasing market size**

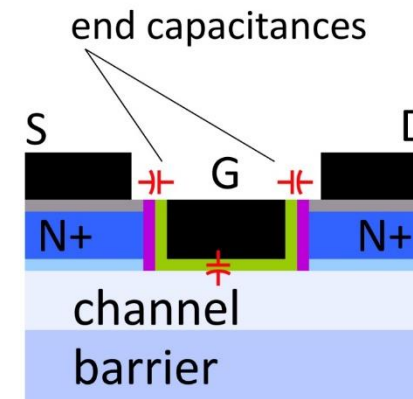
## Gate dielectric can't be thinned

→ on-current,  $g_m$  can't increase



## Shorter gates give no less capacitance

dominated by ends;  $\sim 1\text{fF}/\mu\text{m}$  total

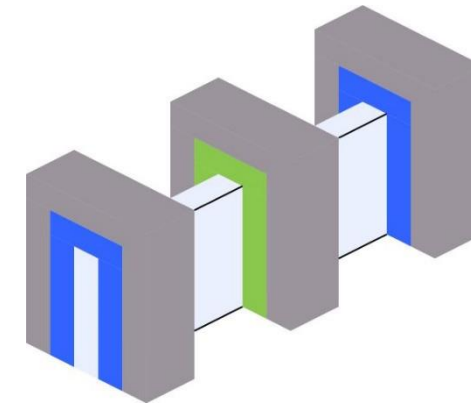


Maximum  $g_m$ , minimum  $C \rightarrow$  upper limit on  $f_T$   
about 350-400 GHz.

Tungsten via resistances reduce the gain

Inac et al, CSICS 2011

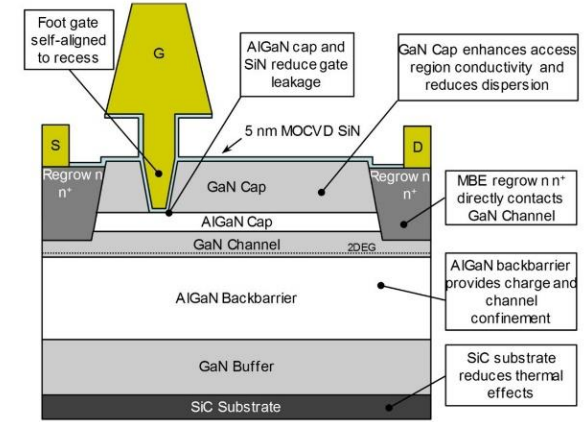
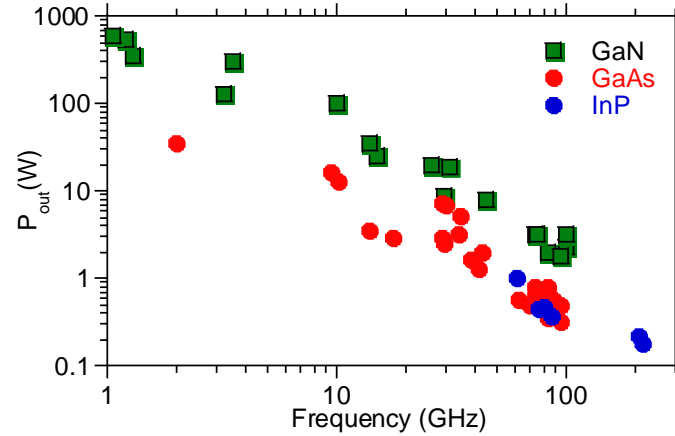
Present finFETs have yet larger end capacitances



## InGaN and GaN HEMTs:

High power from 100-340GHz

GaN: superior power density at all frequencies



N-polar GaN: Mishra, UCSB

## THz InP HBTs:

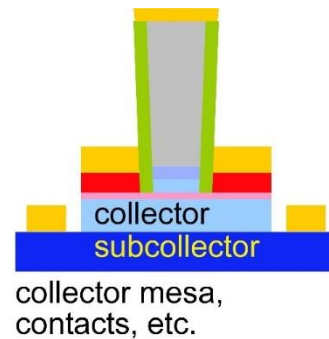
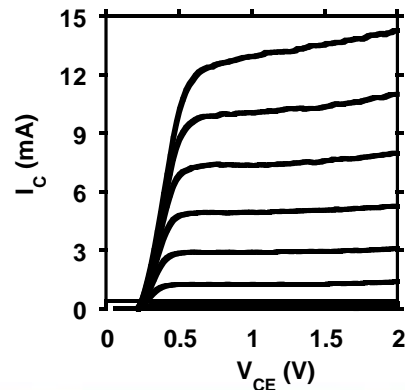
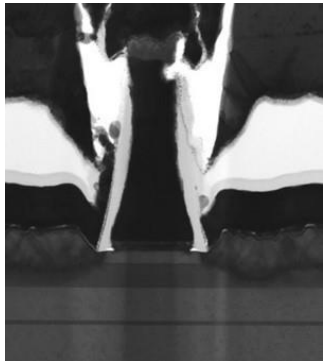
State-of-art: 1.1THz  $f_{max}$  @ 130nm node

Efficient 100-650GHz power

more  $f_{max}$ : more efficient, higher frequencies

base regrowth: better contacts → higher  $f_{max}$

status: working DC devices; moving to THz



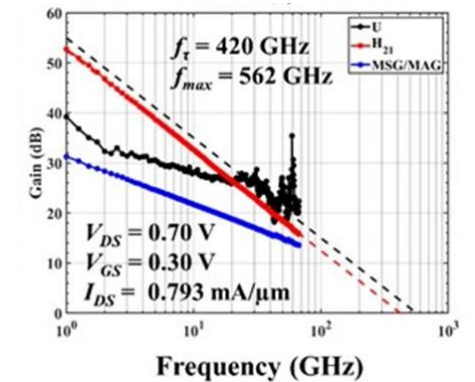
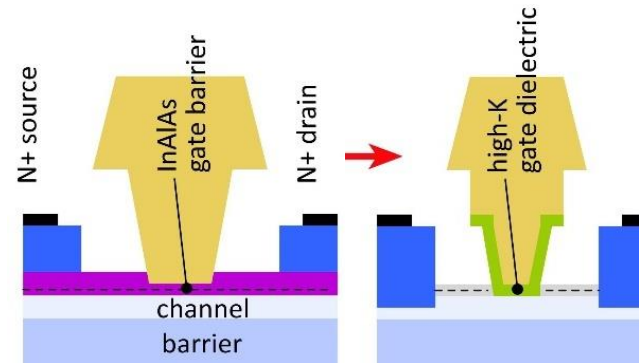
## THz InP HEMTs:

State-of-art: 1.5THz  $f_{max}$  @ 32nm node

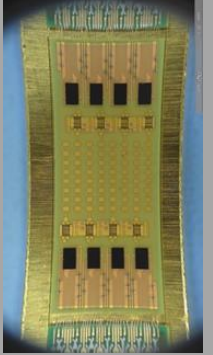
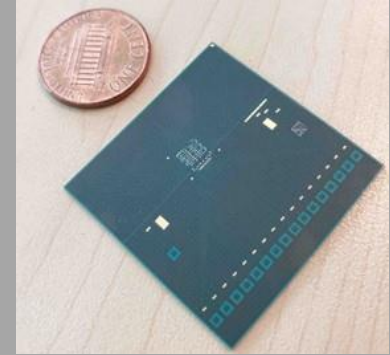
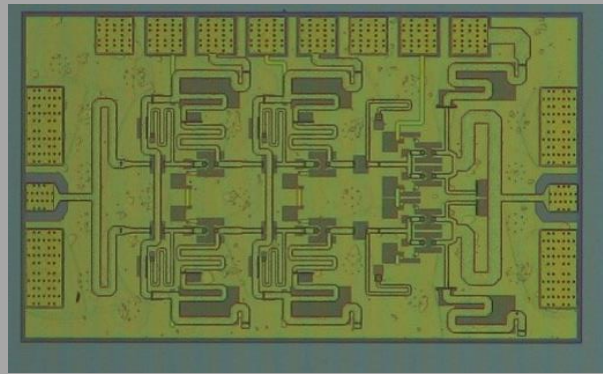
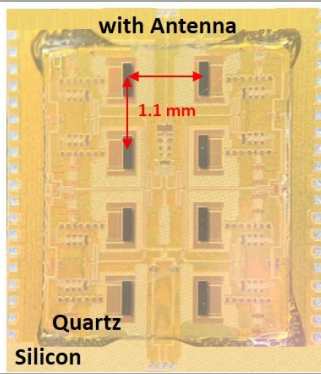
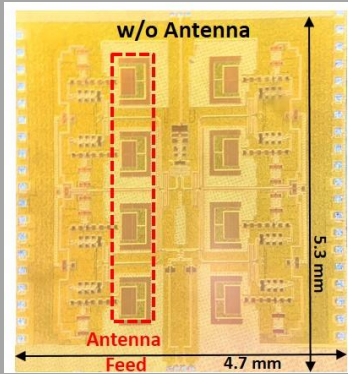
Sensitive 100-650GHz low-noise amplifiers

more  $f_{\tau}$ : lower noise, higher frequencies

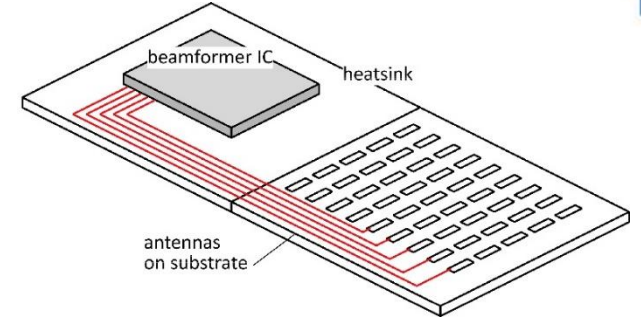
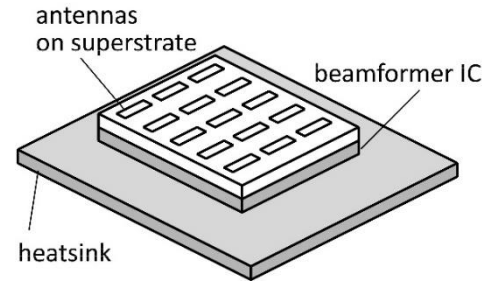
high-K gate dielectric → higher  $f_{\tau}$ ,  $f_{max}$



# ICs and Packages: 140 GHz

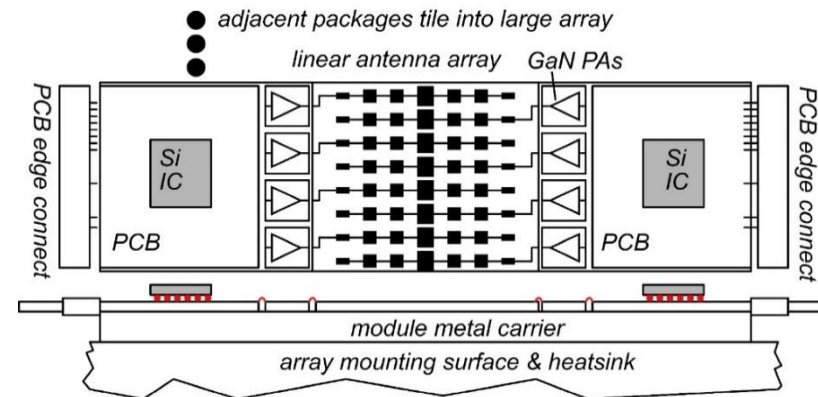
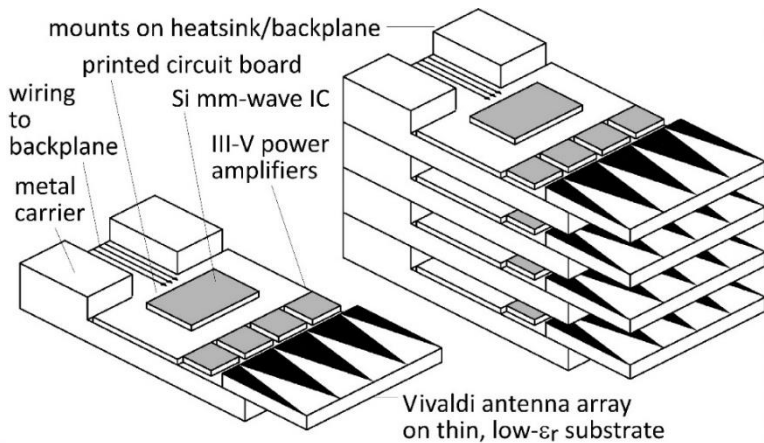
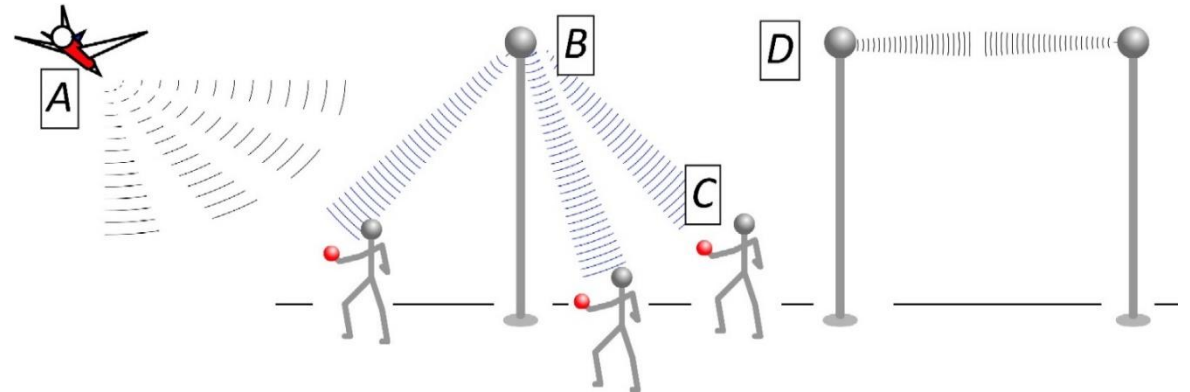


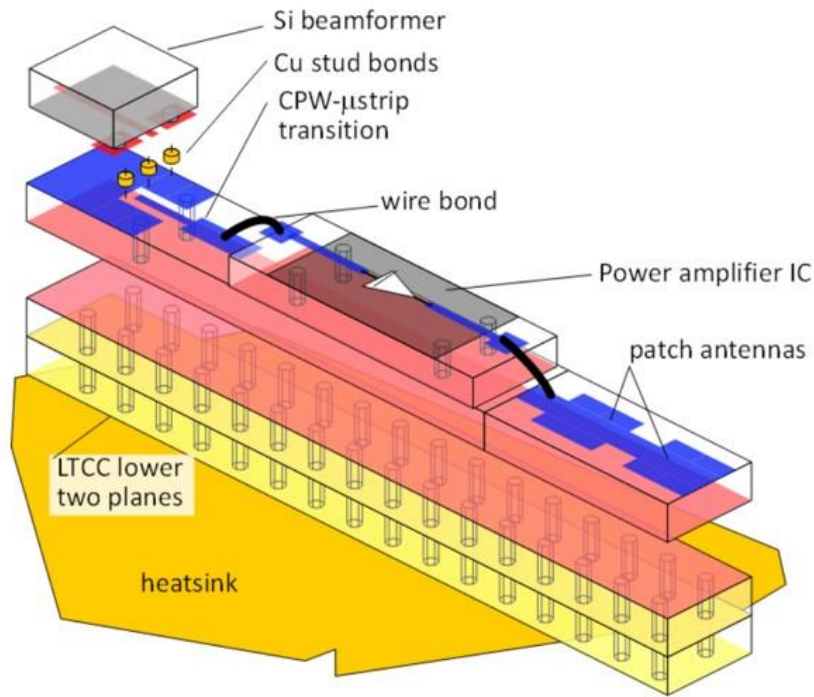
How to make the IC electronics fit ?  
 How to avoid catastrophic signal losses ?  
 How to remove the heat ?



Not all systems steer in two planes...  
 ...some steer in only one.

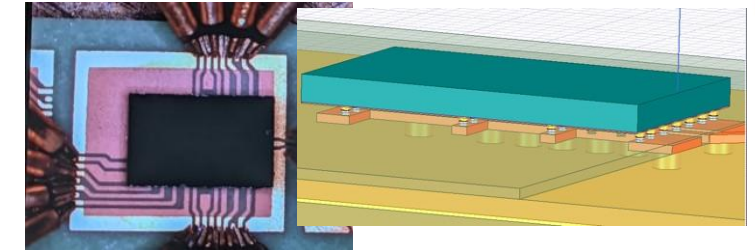
Not all systems steer over 180 degrees...  
 ...some steer a smaller angular range





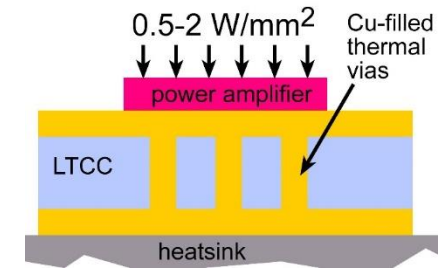
## IC-package interconnects

Difficult at > 100 GHz



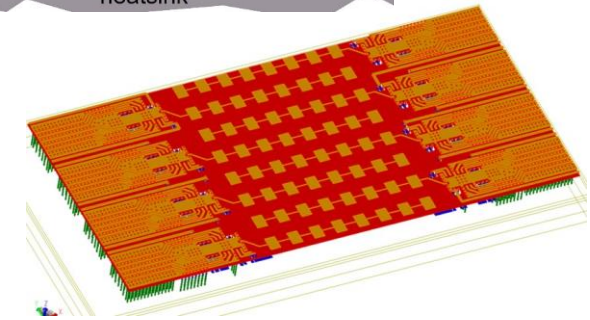
## Removing heat

Thermal vias are marginal



## Interconnect density

Dense wiring for DC, LO, IF, control.  
Hard to fit these all in.



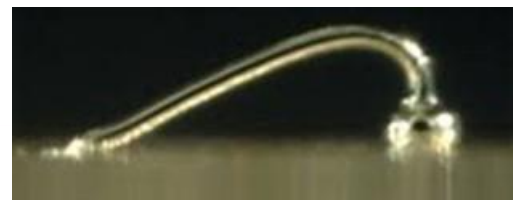
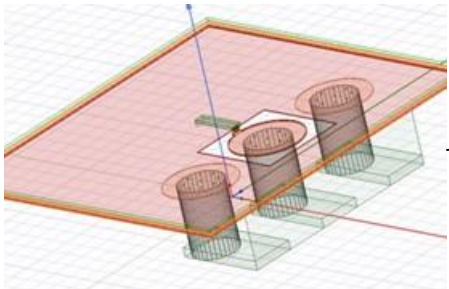
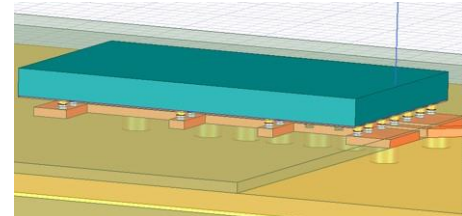
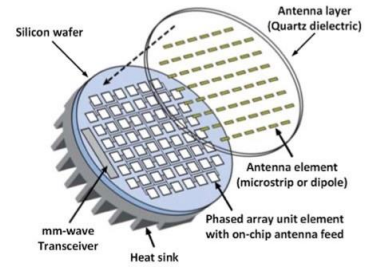
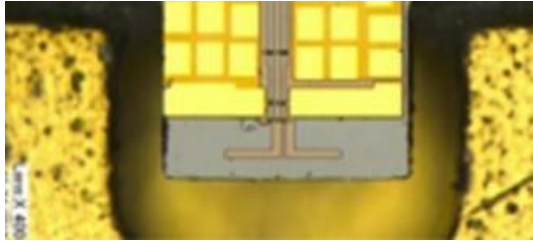
## Economies of scale

Advanced packaging standards require sophisticated tools  
High-volume orders only  
Hard for small-volume orders (research, universities)  
Packaging industry is moving offshore



# 100-300GHz IC-package connections

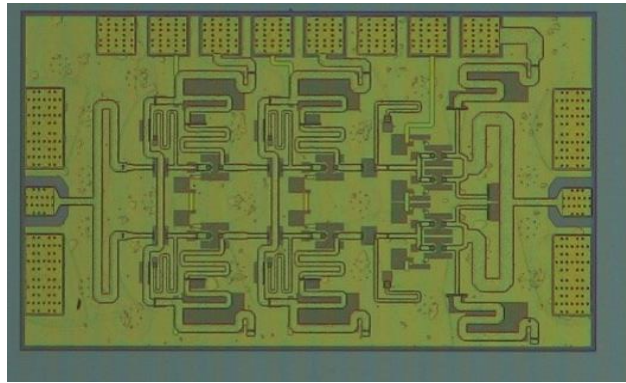
Deal, IEEE Trans THz, Sept 2011



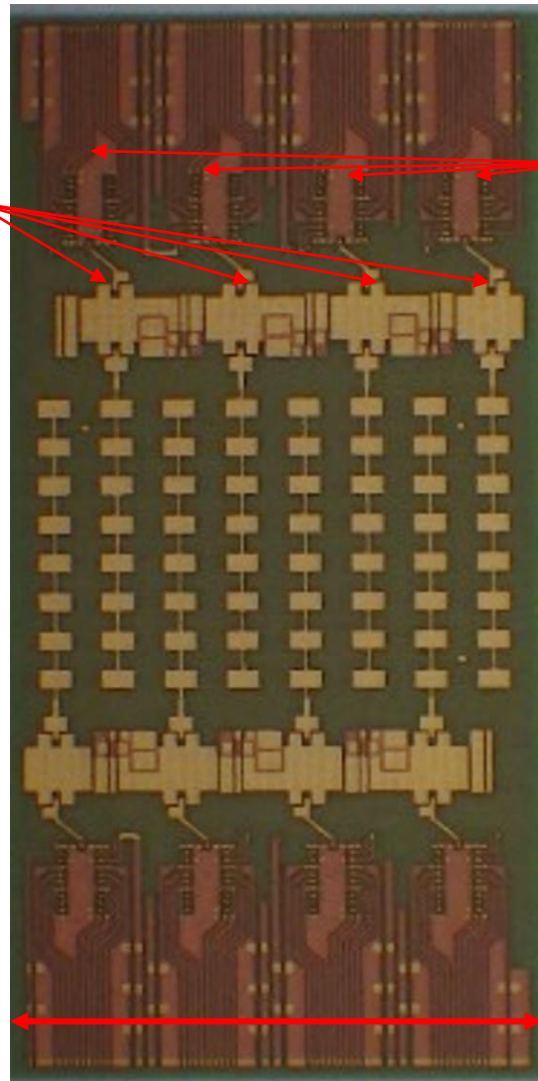
type	Frequency	technology	cost	heatsinking
micromachined waveguide interface	1000 GHz	Research. Cheap one day ?	high X	good
ribbon, mesh bond	200 GHz	Handcrafted.	high X	good
patch antennas on superstrate	1000 GHz	Straightforward	low	good
Cu stud flip-chip	>200 GHz	Industry standard	low	ok, marginal for PA X
hot vias	200 GHz	Development	low ?	good
(ball) wirebonds	100 GHz X	Industry standard	low	good

# 140GHz CMOS+InP MIMO hub array tile

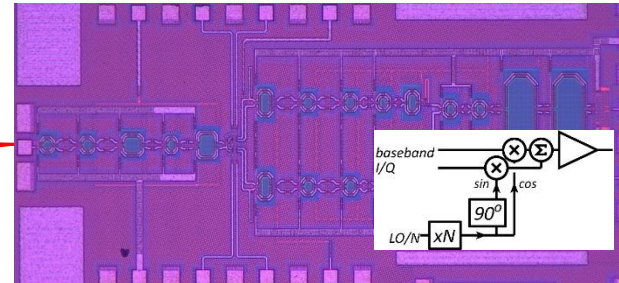
**110mW InP Power Amplifier**  
20.8% PAE



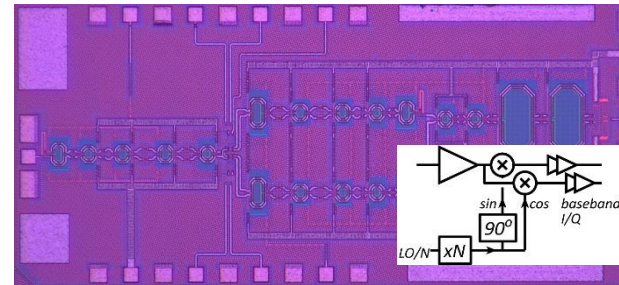
**LTCC Array module**



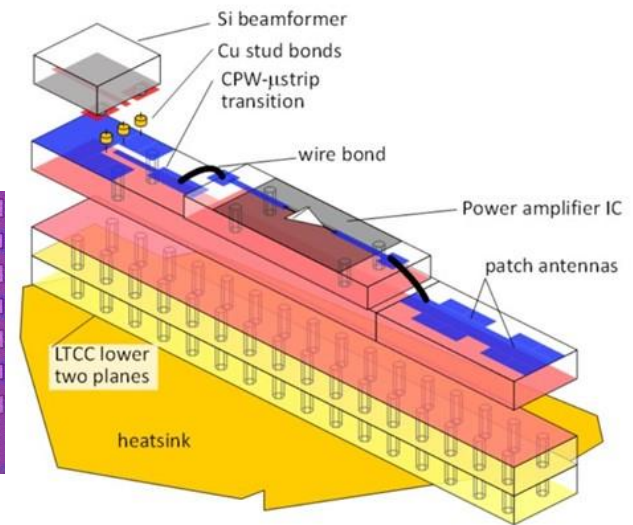
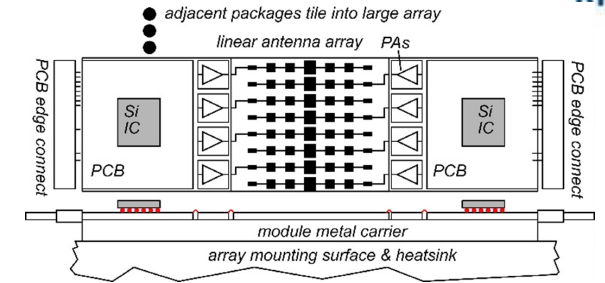
**CMOS Transmitter IC**  
22nm SOI CMOS.



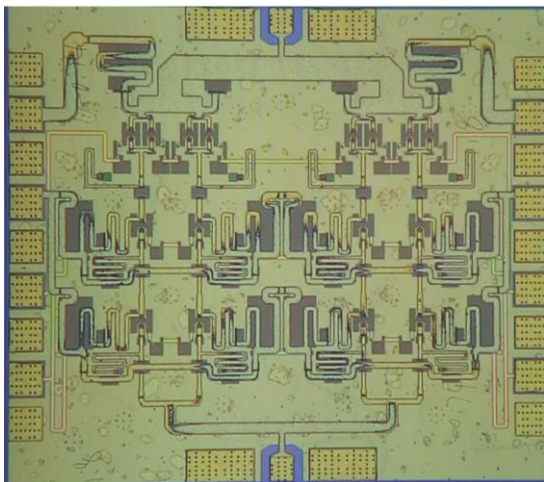
**Receiver IC**  
22nm SOI CMOS.



GlobalFoundries 22nm SOI CMOS



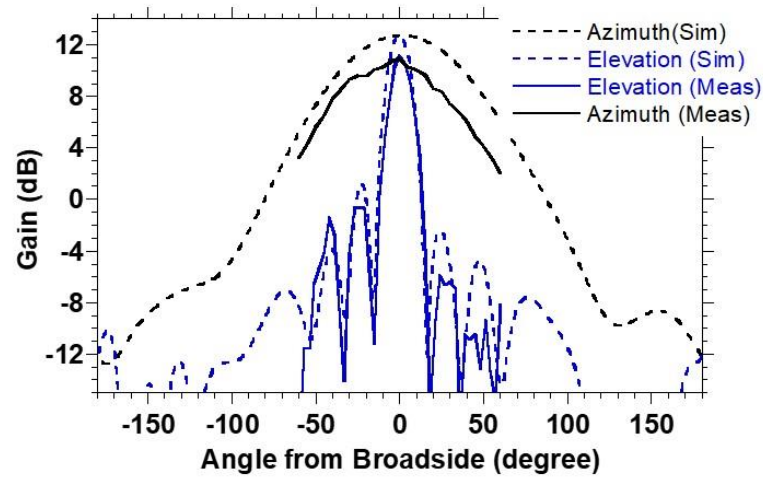
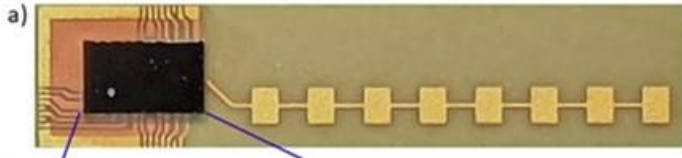
**190mW InP Power Amplifier**  
16.7% PAE



Teledyne InP HBT

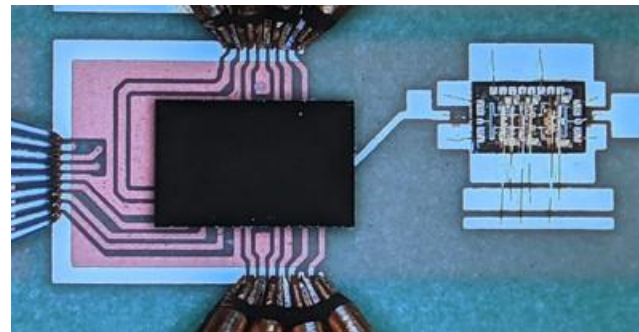
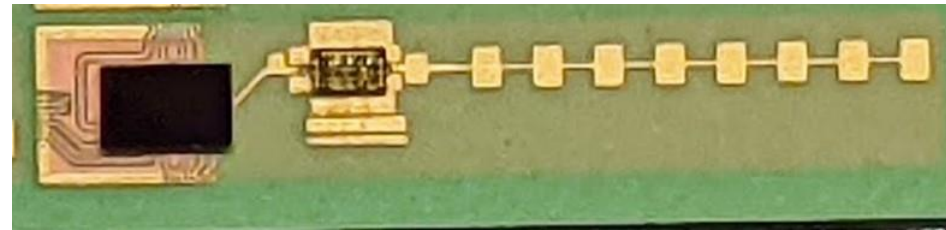
Kyocera

## CMOS-only TX channel

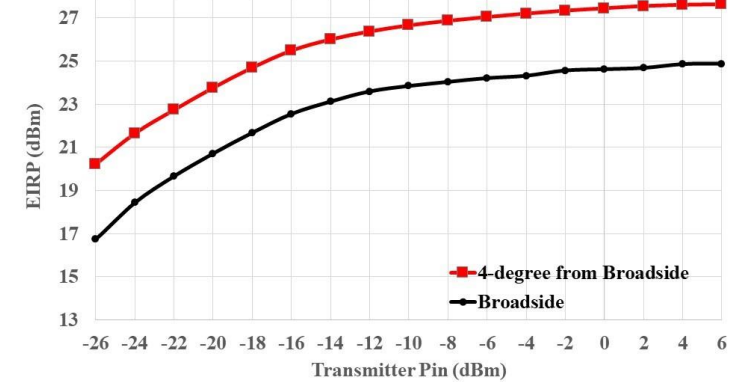


rate	4 Gbaud	4 Gbaud	2 Gbaud
power	3dB backoff	6dB backoff	8dB backoff
EVM (RMS)	7.9%	9.2%	7.4%

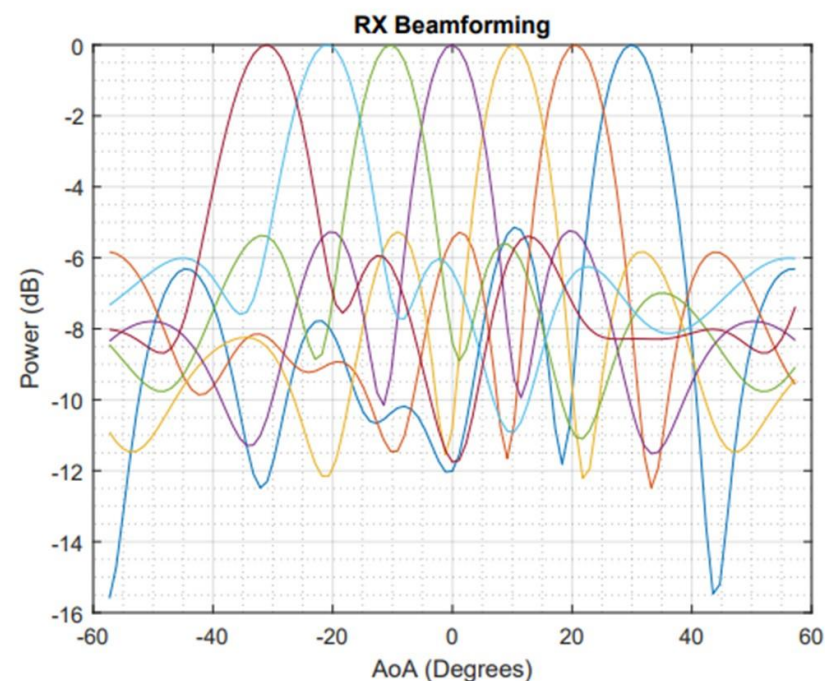
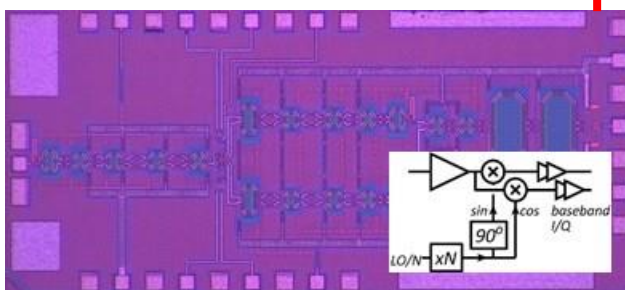
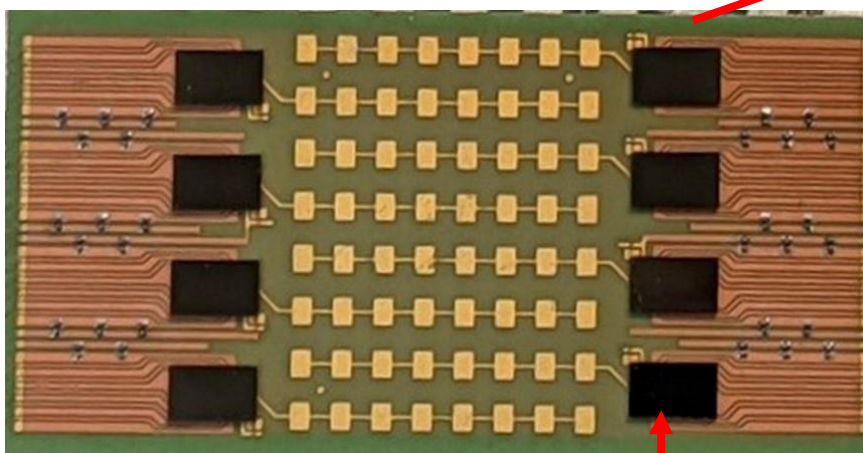
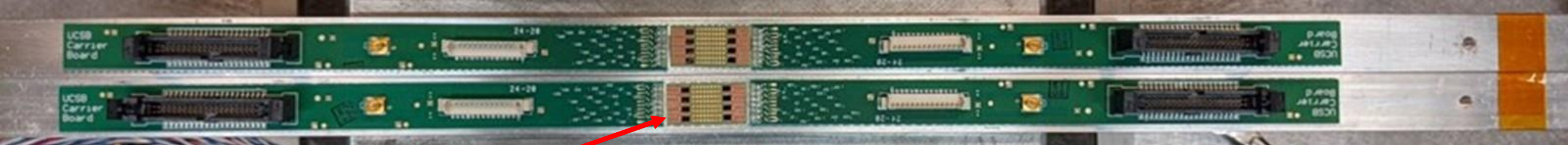
## CMOS+InP TX channel



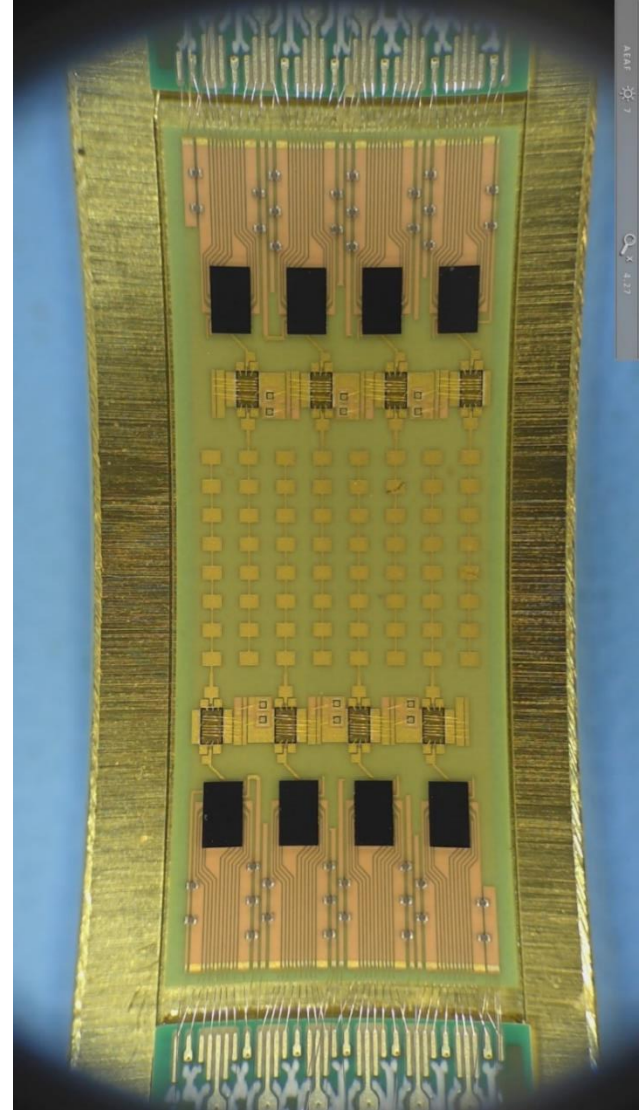
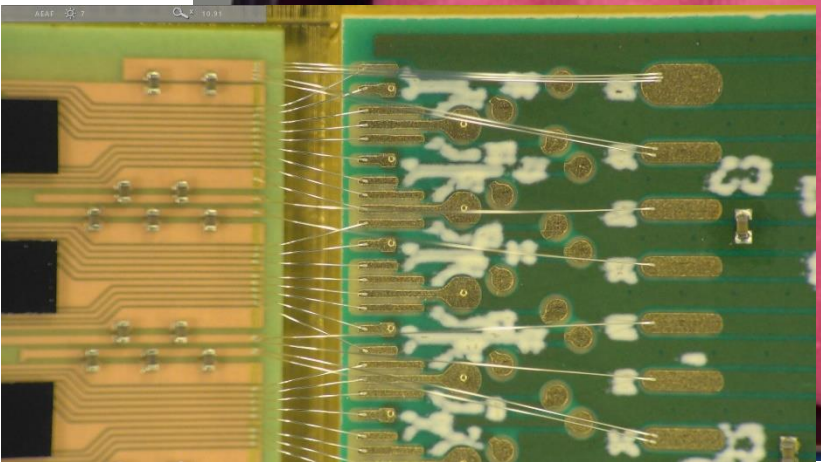
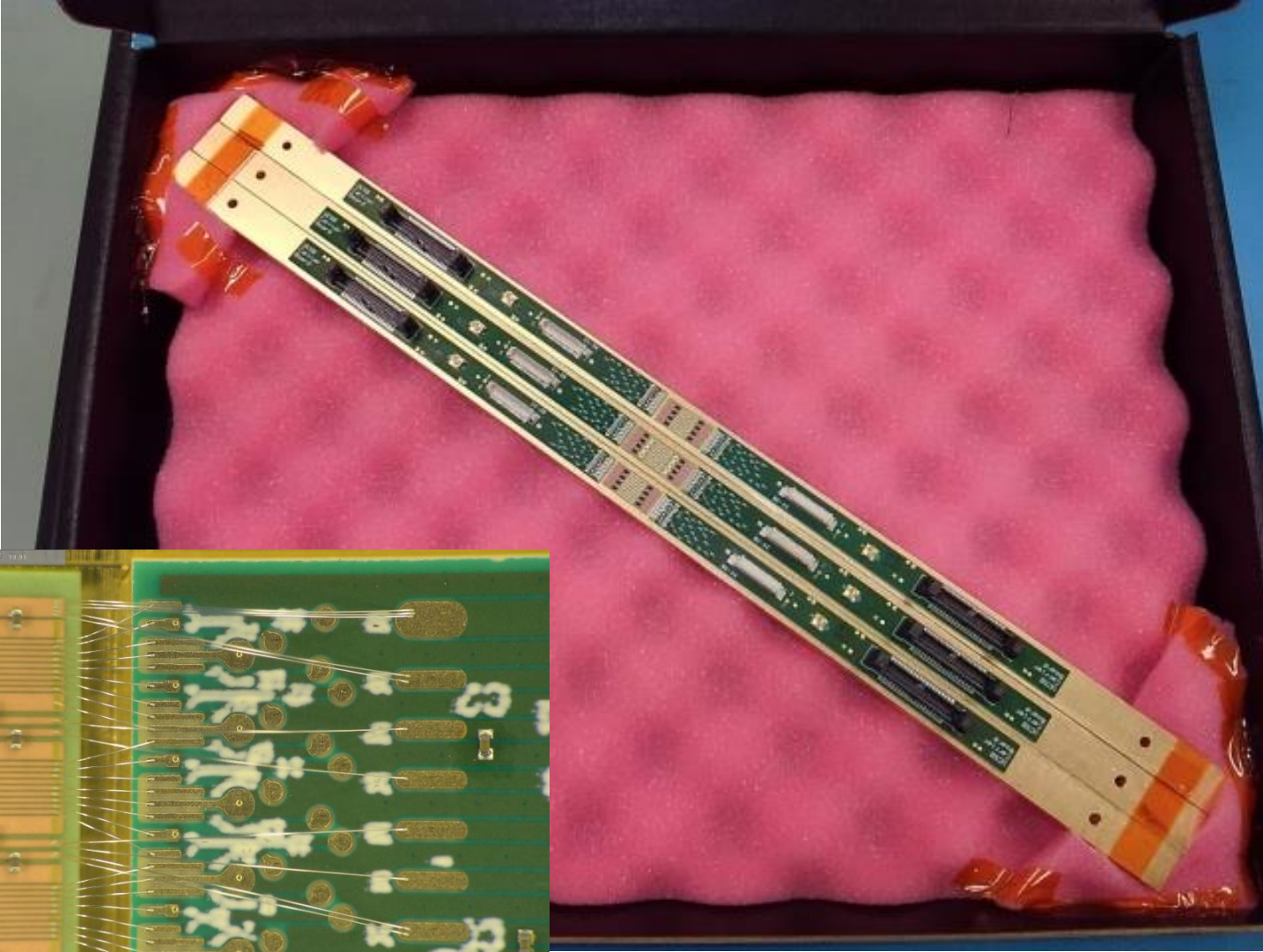
135GHz Transmitter's EIRP Vs Pin



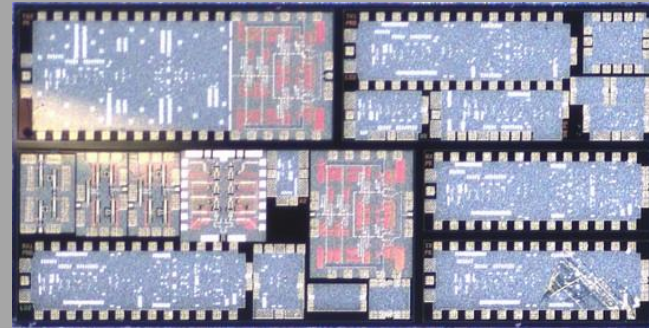
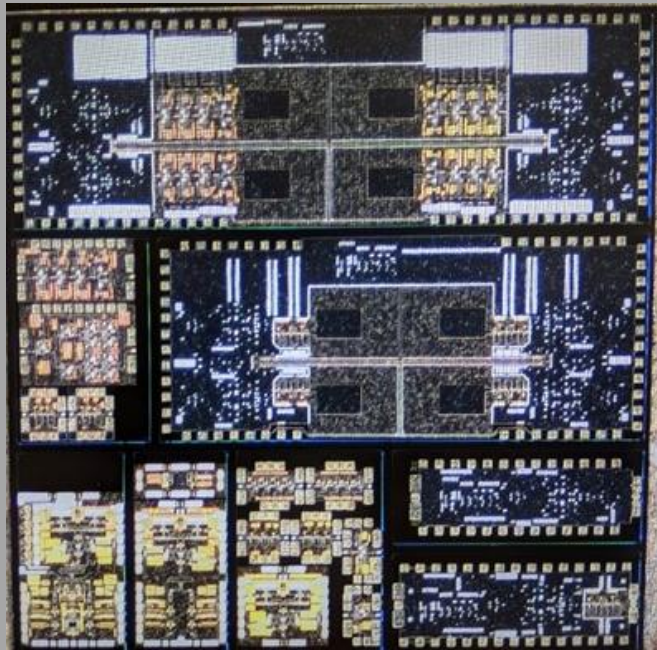
EIRP =19dBm/ 6dB-BO from Psat		
QPSK (5G Baud)	16QAM (5G Baud)	64QAM (5G Baud)
7.69% (RMS)	8.4% (RMS)	8.5% (RMS)

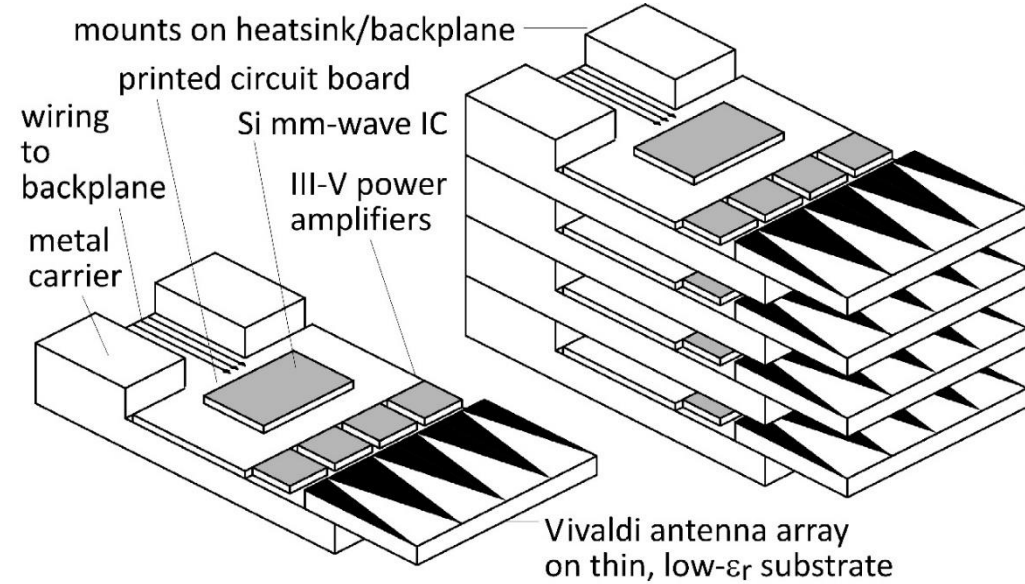
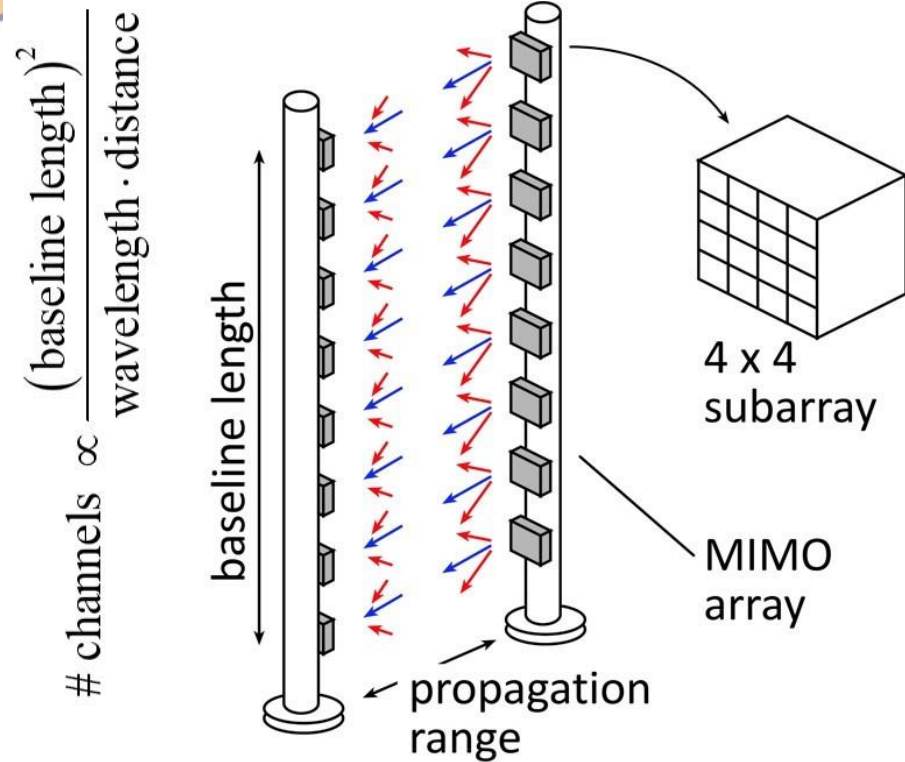


# 8-Channel 140GHz MIMO hub modules being tested.



# ICs and Packages: 210 & 280 GHz





## 8-element MIMO array

3.1 m baseline for 500m link.

80Gb/s/subarray → 640Gb/s total

4 × 4 sub-arrays → 8 degree beamsteering

## Key link parameters

500 meters range in 50 mm/hr rain; 23 dB/km

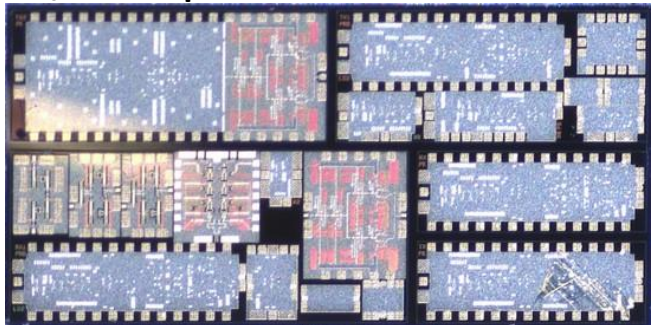
20 dB total margins:

packaging loss, obstruction, operating, design, aging

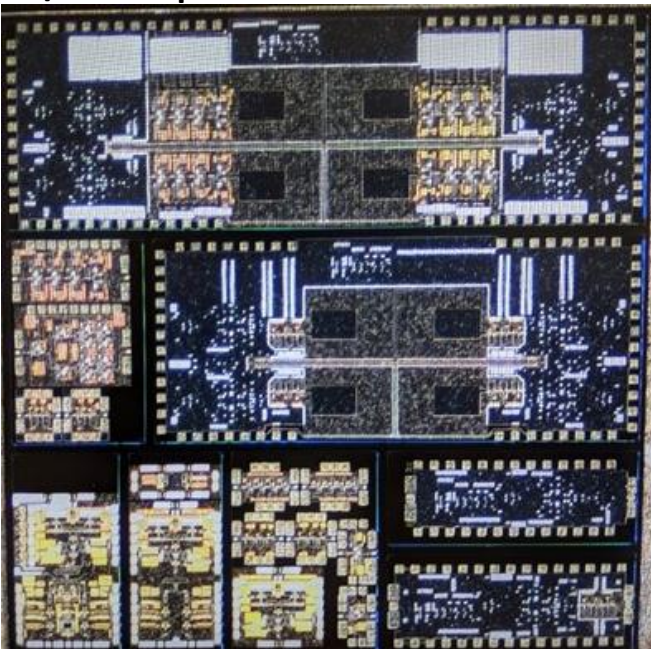
PAs: 63mW =  $P_{1dB}$  (per element)

LNAs: 6dB noise figure

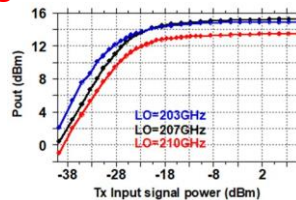
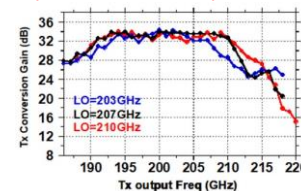
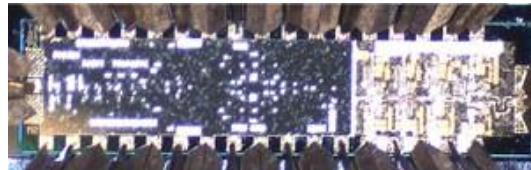
2/2020 tapeout



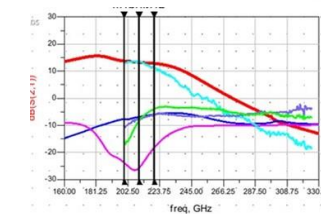
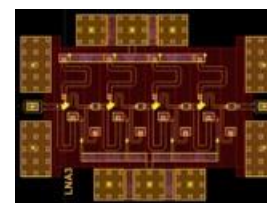
5/2020 tapeout



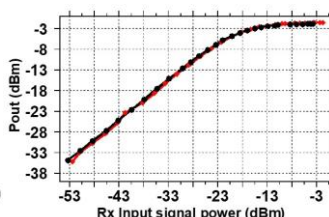
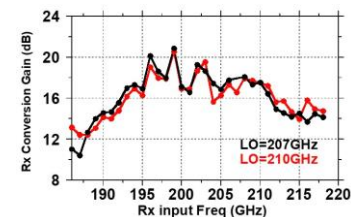
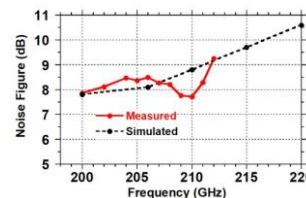
210 GHz TX: measured 15.8 dBm Psat, 20GHz BW, 33dB gain



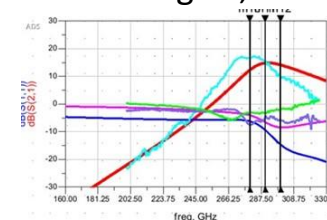
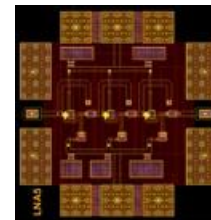
210 GHz LNA: measured 13dB gain, NF TBD.



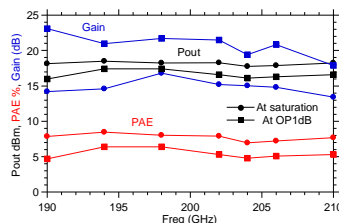
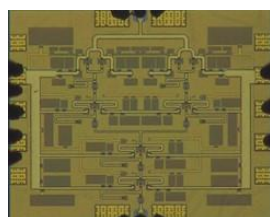
210 GHz RX: measured 27dB gain, 25GHz BW, 8-8.5dB NF.



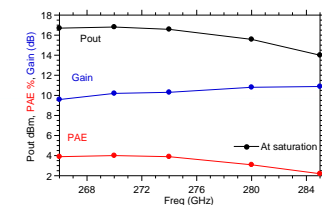
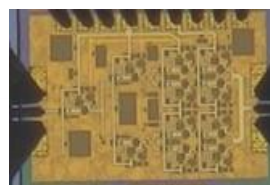
280 GHz LNA: measured 12dB gain, NF TBD.



200 GHz PA: measured 17.7-18.5 dBm Psat, 6.9-8.5% PAE.



270 GHz PA: measured 16.7 dBm Psat, 4% PAE.



**Acknowledgements & Publications:**

- 200GHz PA: A. Ahmed et al, 2021 IMS
- 270GHz PA: A. Ahmed et al, 2021 RFIC symposium
- 210GHz TX, RX: M. Seo et al, 2021 IMS

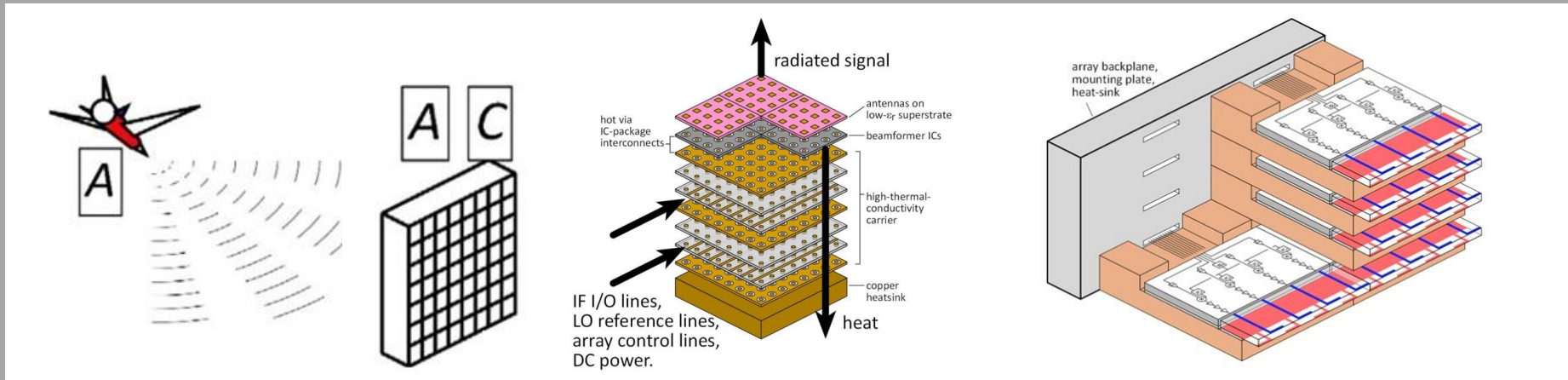
**Ongoing efforts:**

- 210GHz single-channel TX, RX modules
- 2x2 arrays (with antenna superstrate: Rebiez, UCSD)

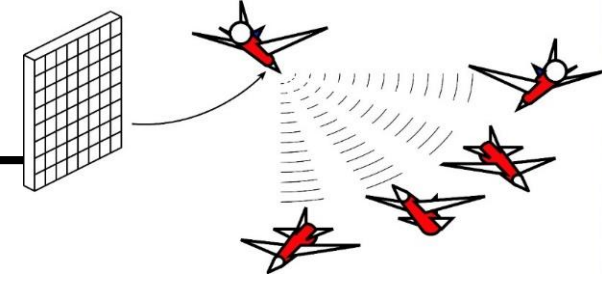
**Technology:**  
Teledyne 250nm InP HBT.



# 2D arrays



# The 100-300GHz 2D Array Challenge



## System architecture:

Single-beam: simpler RF front-end, simpler baseband

MIMO: complex digital baseband, flexible, many beams

Arrays can be made from either **tiles** or **trays**

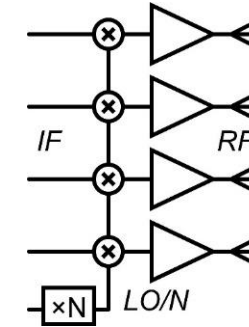
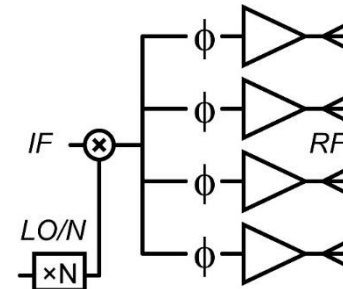
**Arrays must be vast:** 100-1,000-10,000 elements

**Arrays must be dense: packaging challenges**

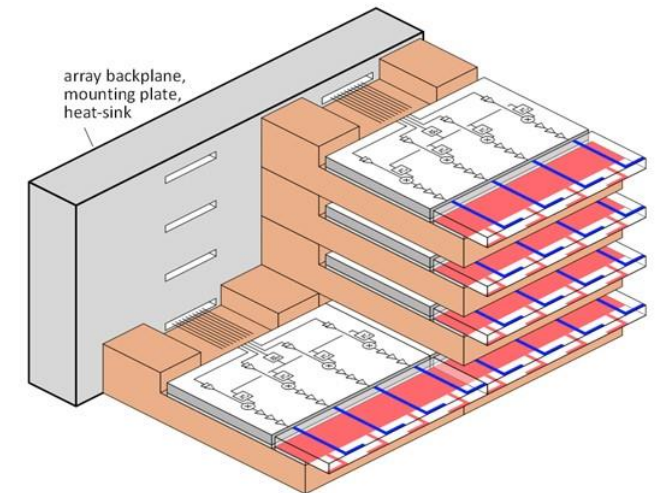
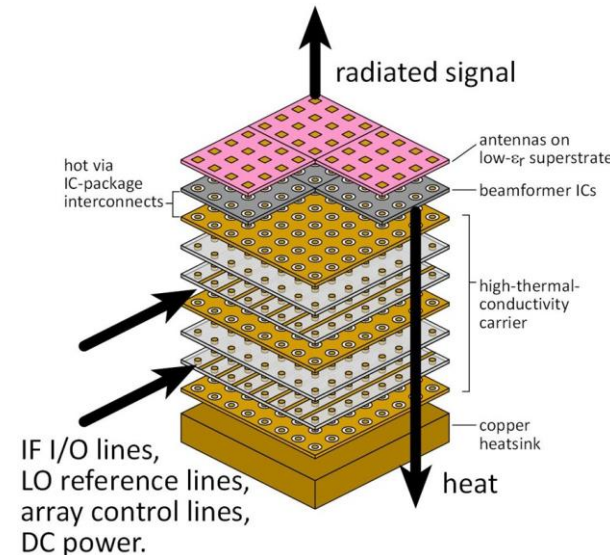
Many DC/IF/LO lines, plus antenna interface.

Fitting IC functions into available area.

Removing the heat.



f	100	150	200	250	300	GHz
$\lambda$	3	2	1.5	1.2	1	mm
$\lambda/2$	1.5	1	0.75	0.6	0.5	mm
$0.6\lambda$	1.8	1.2	0.9	0.72	0.6	mm

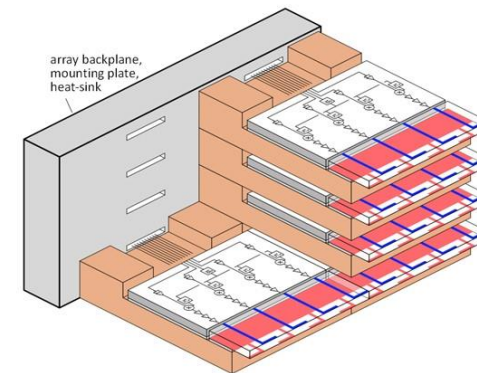
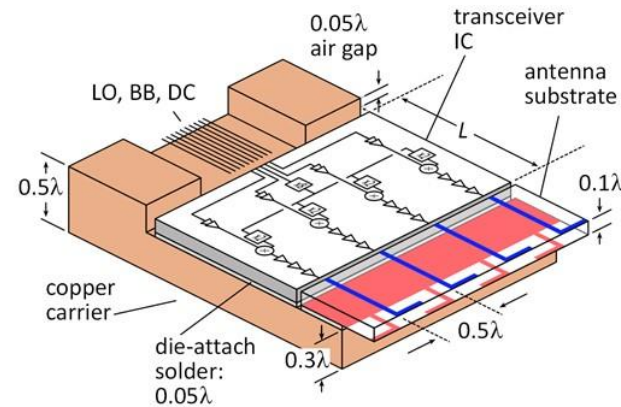
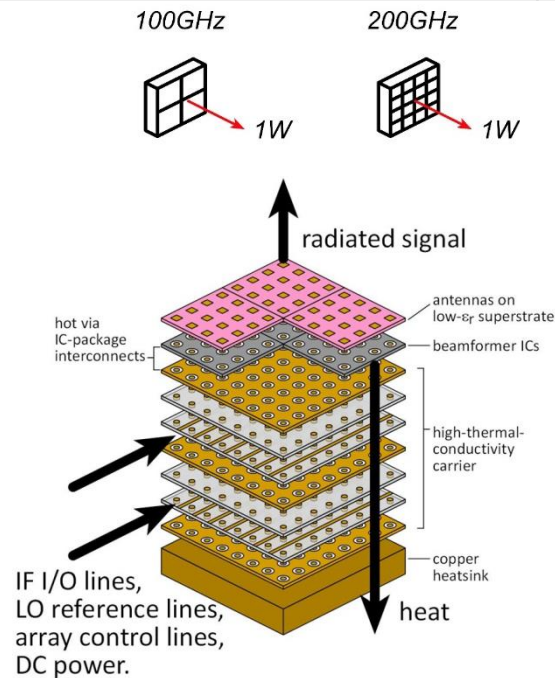


# 100-300GHz array frequency scaling

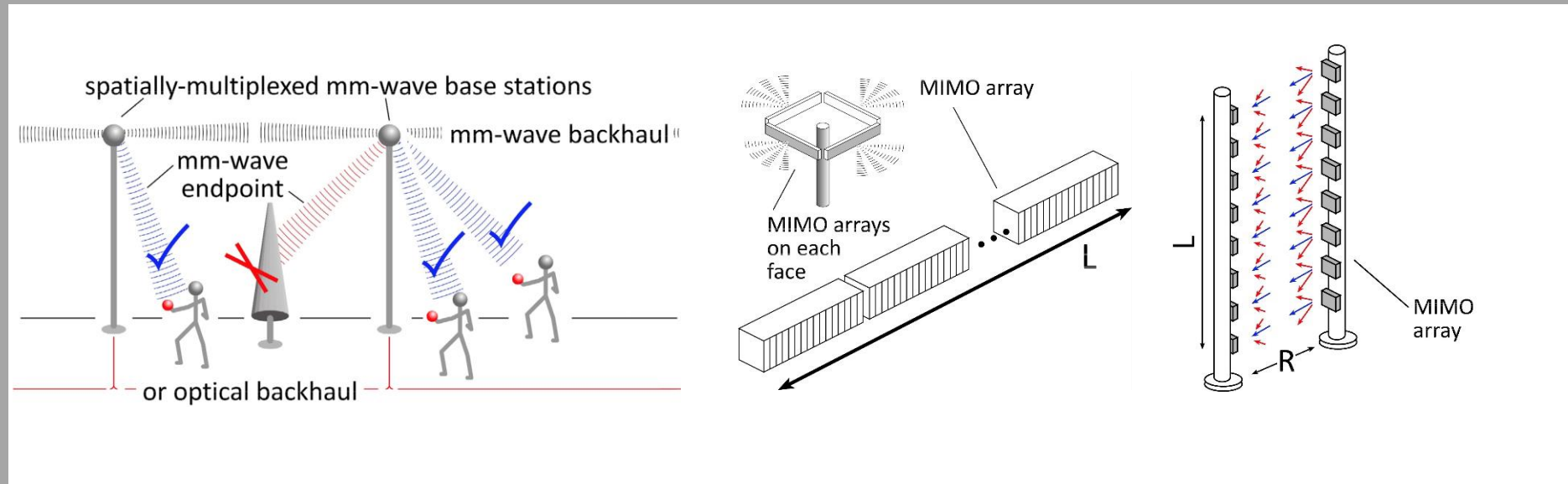
$$P_{received} = \frac{A_t A_r}{\lambda^2 R^2} e^{-\alpha R} \cdot P_{trans} \longrightarrow \# \text{beams} \cdot (\text{bit rate per beam}) \cdot kTF \cdot \text{SNR} = \frac{A_t A_r}{\lambda^2 R^2} e^{-\alpha R} \cdot P_{trans}$$

(Worst-case atmospheric loss: ~constant over 50-300GHz)

Proposed scaling law	change	Implication	change
carrier frequency	increase 2:1	capacity (# beams · bit rate per beam)	increases 4:1
aperture area	keep constant	number elements	increases 4:1
total transmit power	keep constant	RF power per cm <sup>2</sup> aperture area	stays constant
		RF power per element	decreases 4:1
		IC area/element (tiled array)	decreases 4:1
		IC area/element (trayed array)	decreases 2:1
		IC power/area (tiled array)	stays constant
		IC power/area (trayed array)	decreases 2:1



# 100-300GHz Wireless



## Massive capacities

large available bandwidths

massive spatial multiplexing in base stations and point-point links

## Very short range: few 100 meters

short wavelength, high atmospheric losses. Easily-blocked beams.

## IC Technology

All-silicon for short ranges below 200 GHz.

III-V LNAs and PAs for longer-range links. Just like cell phones today

III-V frequency extenders for 340GHz and beyond

## The challenges

computational complexity

packaging: fitting signal channels in very small areas

mesh networking to accommodate beam blockage

driving the technologies to low cost

(backup files follow)