



# 200 GHz Low Noise Amplifiers in 250 nm InP HBT Technology

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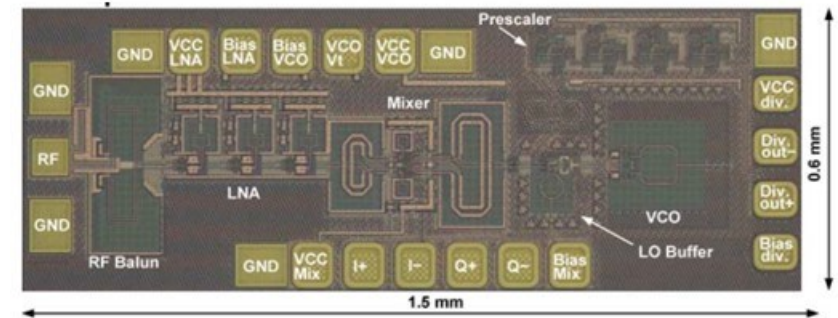
# Outline

- Motivation
- Amplifier design
  - Noise measure technique
  - Low-loss input matching
- Measurement results
- Application of the amplifier
- Summary

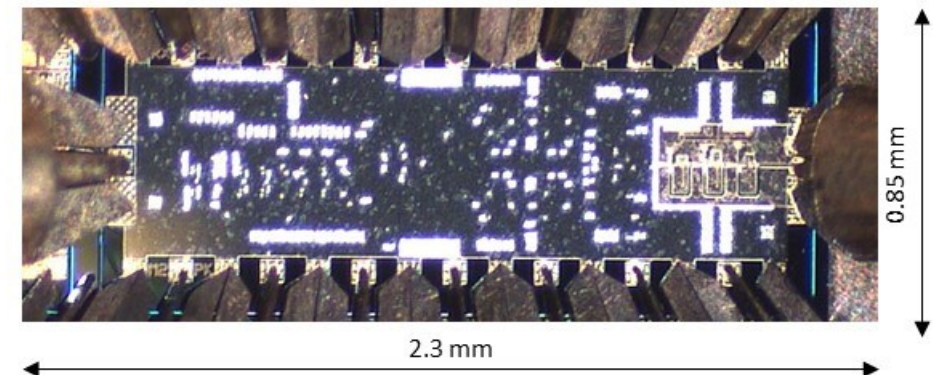
# Motivation

- Increasing interest in the 200-300 GHz:  
Large available BW → High data rate
- Greater integration scales in SiGe/InP HBT receivers than in III-V HEMT technologies
- More competitive 200-300 GHz all-HBT Rx with reduced HBT LNA noise figure
- Ideal Case:  
Hybrid Rx with HEMT LNA + HBT post-LNA

160 GHz Rx in SiGe HBT [2]

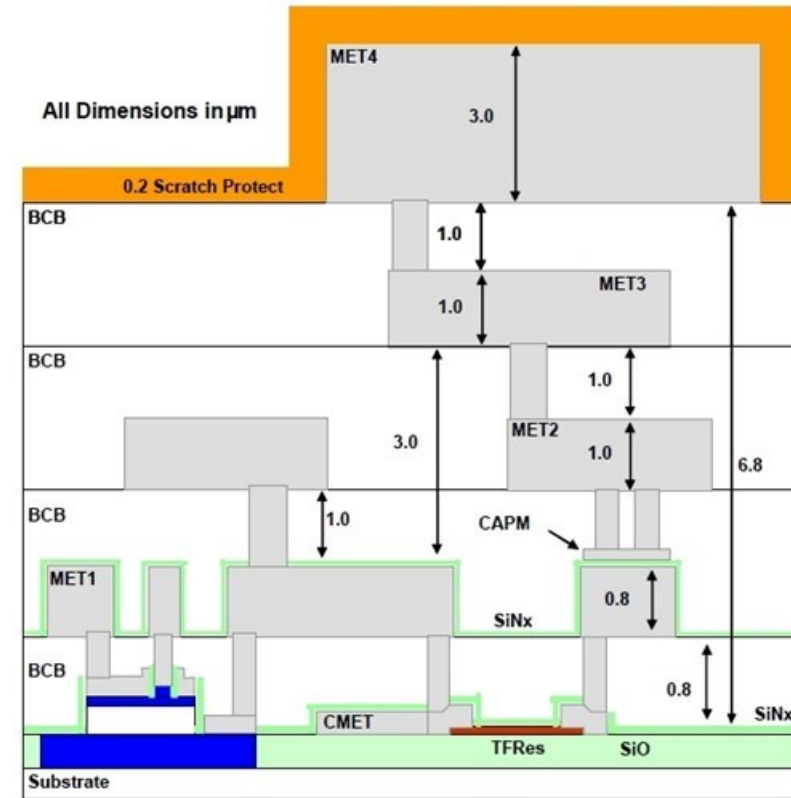


200 GHz Rx in InP HBT [14]



# 250 nm InP HBT Process (Teledyne [1])

- Four Au interconnect.
- TFR (50Ω/square).
- MIM cap (0.3fF/μm<sup>2</sup>).
- $f_{\max} = 650\text{GHz}$ .
- $J_{\max} = 3\text{mA}/\mu\text{m}$ .
- $BV_{\text{CEo}} = 4.5\text{V}$ .
- MET4 = Signal, MET1 = GND
- Loss of MSL = 1.1 dB/mm @ 200 GHz
- Loss of inv-MSL = 2.5 dB/mm @ 200 GHz



Representative cross-section of TSC250 IC technology. Drawing is not to scale.

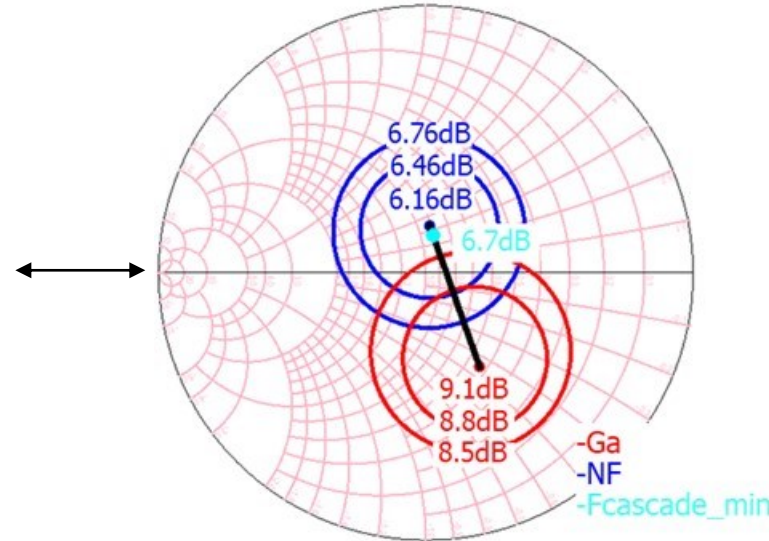
Cross Section of TSC250 IC

# Noise Measure

$$F_{\text{cascade}} = M + 1 = F + \frac{F - 1}{G} + \frac{F - 1}{G^2} + \dots = \frac{F - G^{-1}}{1 - G^{-1}} \rightarrow M = \frac{F - 1}{1 - G^{-1}}$$

```

F_cascade_min1=min(F_cascade1)
F_cascade_db_min1=min(F_cascade_db1)
gamma_S_opt=di_python("gamma_S_opt.py",gamma_S1,F_cascade1,F_cascade_min1,sweep_size(DC.le)(3))
ga_opt=real(di_python("gamma_S_opt.py",ga_value1,F_cascade1,F_cascade_min1,sweep_size(DC.le)(3)))
nf_opt=real(di_python("gamma_S_opt.py",n_value1,F_cascade1,F_cascade_min1,sweep_size(DC.le)(3)))
ga_opt_db=10*log10(ga_opt)
nf_opt_db=10*log10(nf_opt)
F_cascade_min_overall=min(F_cascade_min1)
F_cascade_min_index_overall=find(F_cascade_min1==F_cascade_min_overall)
    
```



Minimum Noise Measure Impedance:

1. Draw a line between centers of NF and Ga Circles.
2. Calculate M for each point on this line.
3. Determine the point on the line having the *smallest* M

Data is for an (0.25 x 5 μm<sup>2</sup>) HBT in CB configuration with 200 fF base capacitance biased at VCB=0.4 V and JE=0.5 mA/μm

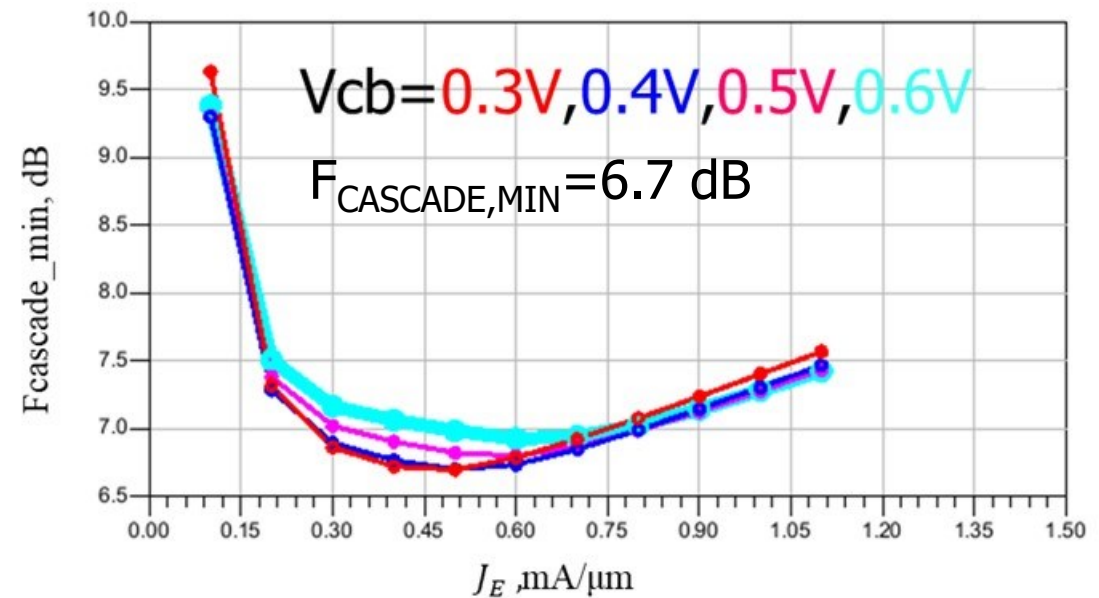


[14] H. Fukui, "Available Power Gain, Noise Figure, and Noise Measure of Two-Ports and Their Graphical Representations," in IEEE Transactions on Circuit Theory, vol. 13, no. 2, pp. 137-142, June 1966.

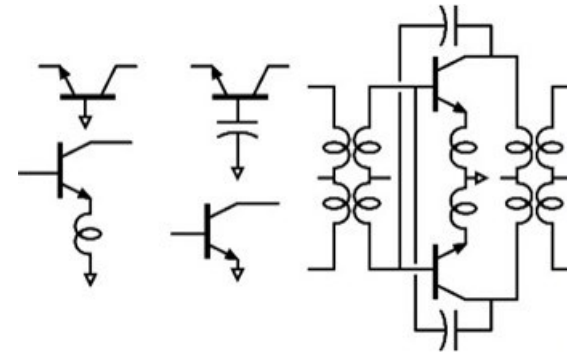
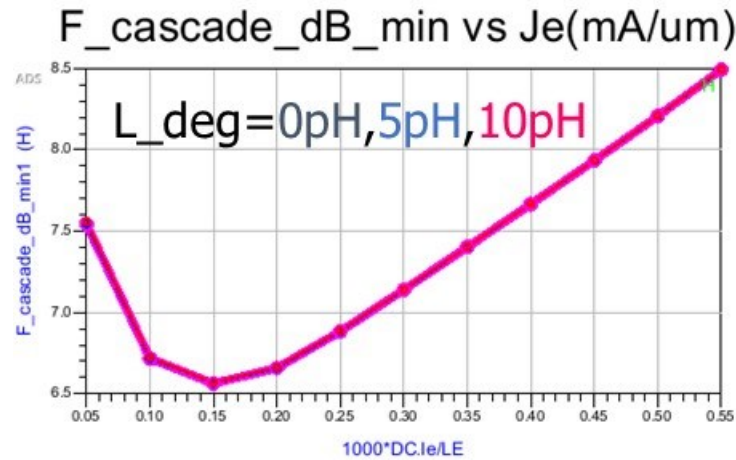
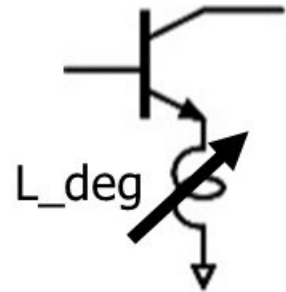
# Determining Bias Condition



$F_{\text{CASCADE,MIN}}$  as a function of emitter current density ( $J_E$ ) and collector-base voltage ( $V_{CB}$ ) for a  $0.25 \mu\text{m} \times 5 \mu\text{m}$  HBT.



# Noise Measure as a 2-port Invariant



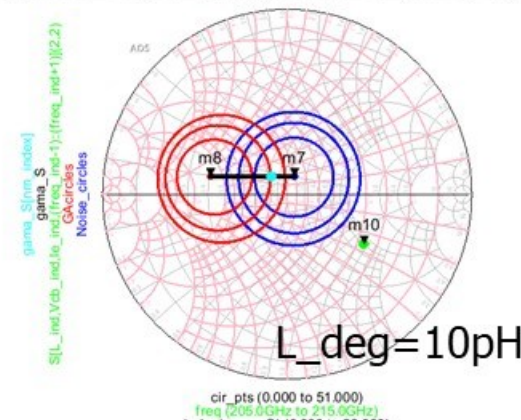
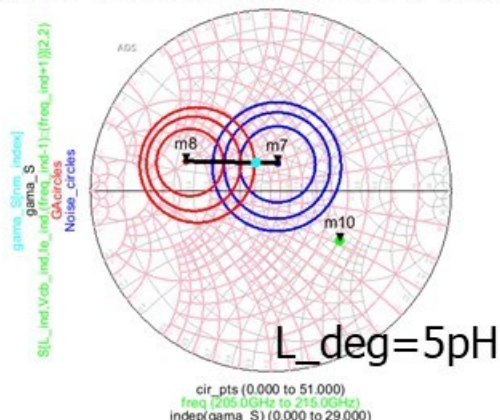
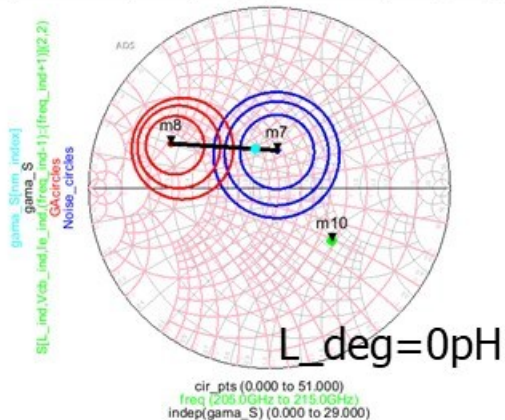
Base Capacitance in CB  
↕  
Emitter Inductance in CE

[10] H. A. Haus and R. B. Adler, "Optimum Noise Performance of Linear Amplifiers," in Proceedings of the IRE, vol. 46, no. 8, pp. 1517-1533, Aug. 1958.

m8 indep(m8)=18 GAcircles=0.597 / 155.759 gain=6.226 impedance = Z0 * (0.263 + j0.201)	m7 indep(m7)=51 Noise_circles=0.212 / 78.090 ns figure=6.859 impedance = Z0 * (0.997 + j0.434)
--	--

m8 indep(m8)=18 GAcircles=0.497 / 160.929 gain=5.473 impedance = Z0 * (0.345 + j0.148)	m7 indep(m7)=51 Noise_circles=0.156 / 75.834 ns figure=6.729 impedance = Z0 * (1.029 + j0.320)
--	--

m8 indep(m8)=18 GAcircles=0.426 / 166.833 gain=4.954 impedance = Z0 * (0.407 + j0.096)	m7 indep(m7)=51 Noise_circles=0.104 / 67.735 ns figure=6.603 impedance = Z0 * (1.061 + j0.206)
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Minimum M is independent of circuit configuration;

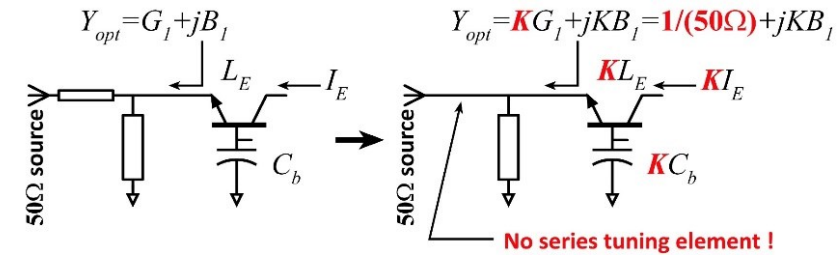
Pick for high bandwidth or high gain/stage (=low P<sub>DC</sub>)

Data is for an (0.25 x 3 μm<sup>2</sup>) HBT in CE configuration biased at VCB=0.4 V and JE=0.5 mA/μm

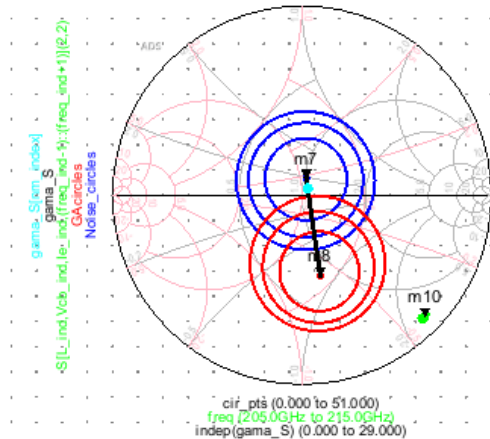
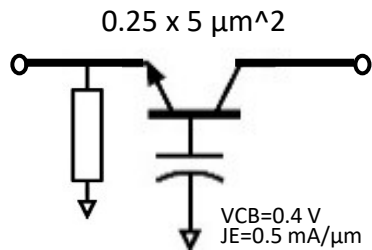
# Input/Output Matching Matching

$$F_{\text{cascade}} = F + \frac{F-1}{G} + \frac{F-1}{G^2} + \dots = \frac{F-G^{-1}}{1-G^{-1}}$$

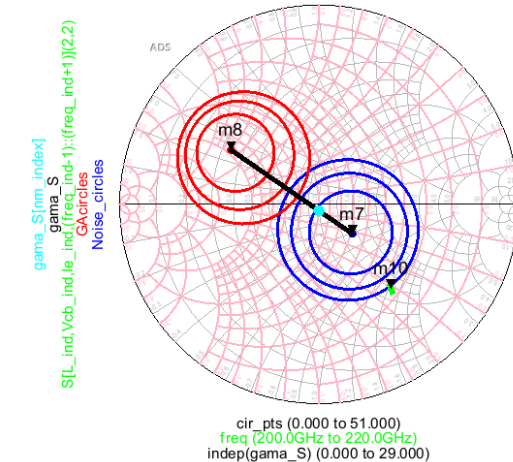
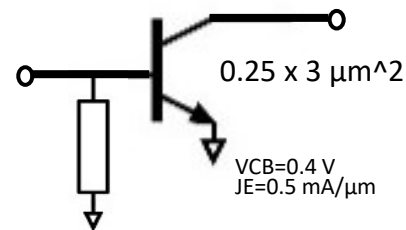
$$F_{\text{cascade}} = F_{\text{input\_match}} + F_t + \frac{F_{\text{output\_match}} - 1}{G_t} + \dots = \frac{F-G^{-1}}{1-G^{-1}}$$



m10 freq=210.0GHz S[L_ind,Vcb_ind,Ie_ind,(freq_ind-1):(freq_ind+1)](2,2)=0.913 / -44.626 impedance = Z0 * (0.313 - j2.402)		m7 indep(m7)=51 Noise_circles=0.091 / 75.744 ns figure=6.280 impedance = Z0 * (1.029 + j0.183)	
m8 indep(m8)=18 GAcircles=0.436 / -77.155 gain=8.978 impedance = Z0 * (0.813 - j0.854)		m7 indep(m7)=51 Noise_circles=0.223 / -41.455 ns figure=4.986 impedance = Z0 * (1.329 - j0.413)	

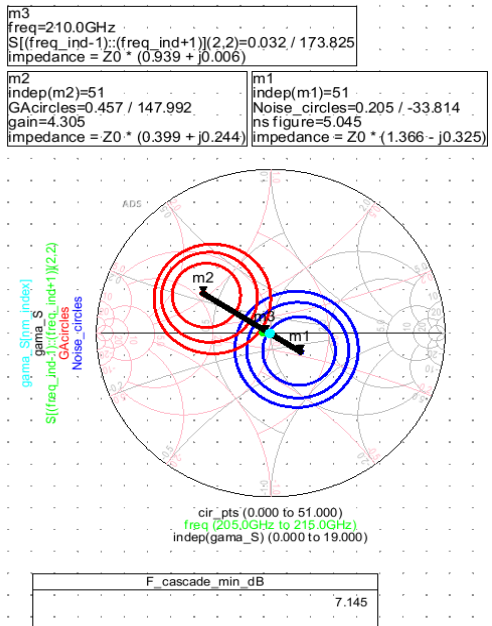


m10 freq=210.0GHz S[L_ind,Vcb_ind,Ie_ind,(freq_ind-1):(freq_ind+1)](2,2)=0.555 / -49.449 impedance = Z0 * (1.180 - j1.438)		m7 indep(m7)=51 Noise_circles=0.223 / -41.455 ns figure=4.986 impedance = Z0 * (1.329 - j0.413)	
m8 indep(m8)=18 GAcircles=0.518 / 148.302 gain=4.938 impedance = Z0 * (0.340 + j0.253)		m7 indep(m7)=51 Noise_circles=0.223 / -41.455 ns figure=4.986 impedance = Z0 * (1.329 - j0.413)	

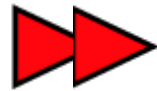
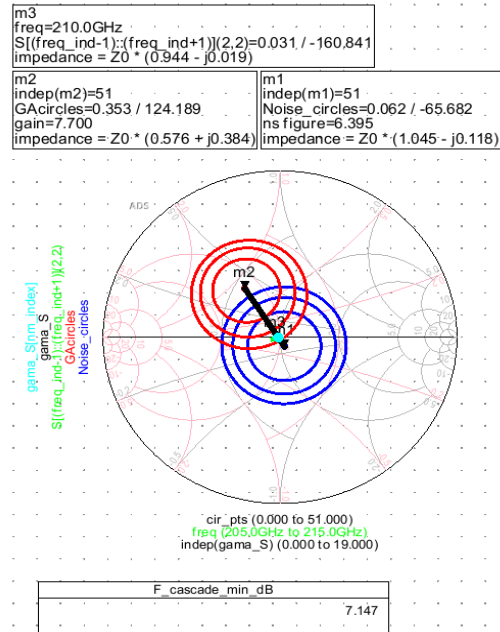




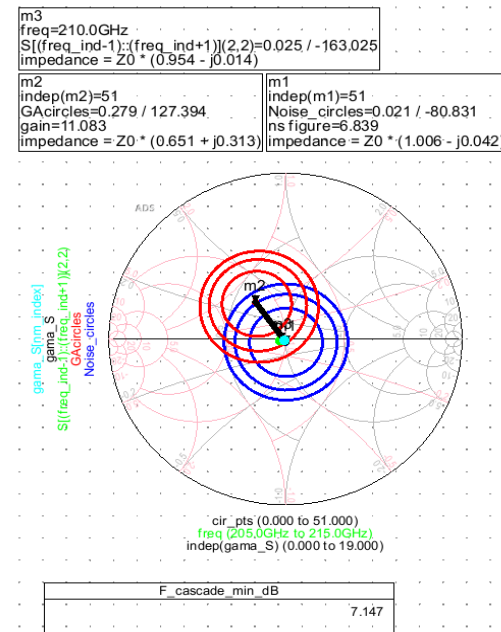
# Cascading



1 stage, CE



2 stage, CE

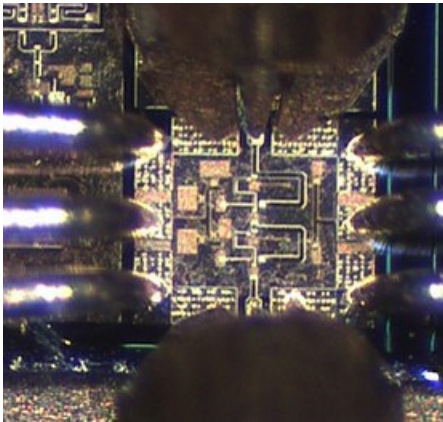


3 stage, CE

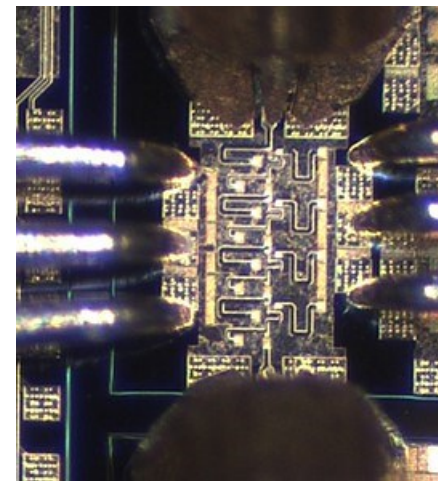
As we cascade more stages:

1.  $NF_{MIN}$  converges to  $F_{CASCADE}$
2. Minimum NF impedance converges to *Minimum NM impedance*

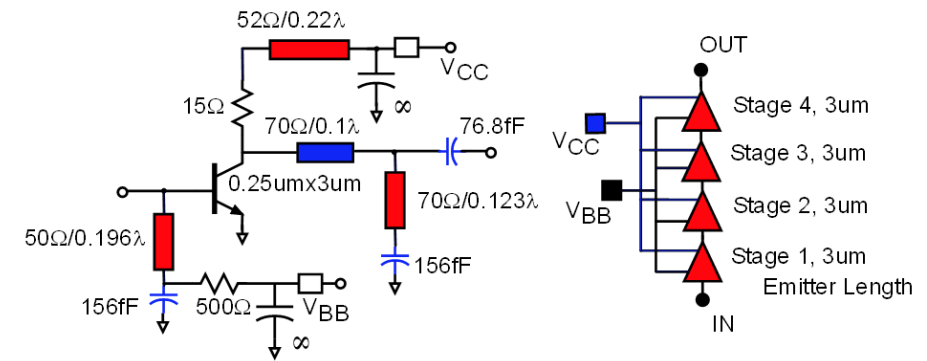
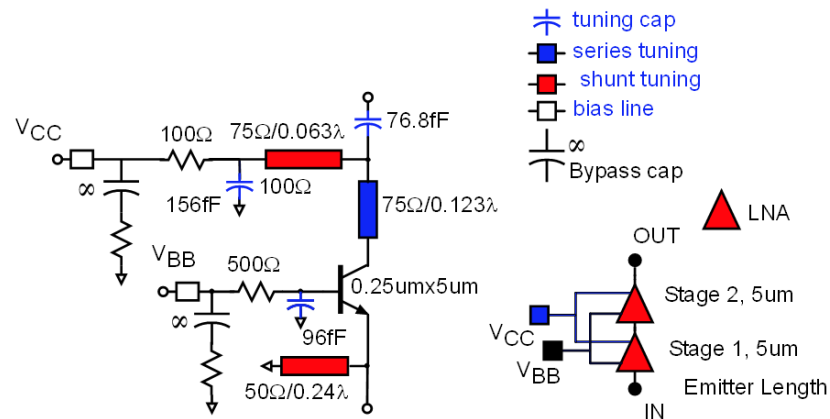
# CB and CE Low Noise Amplifiers



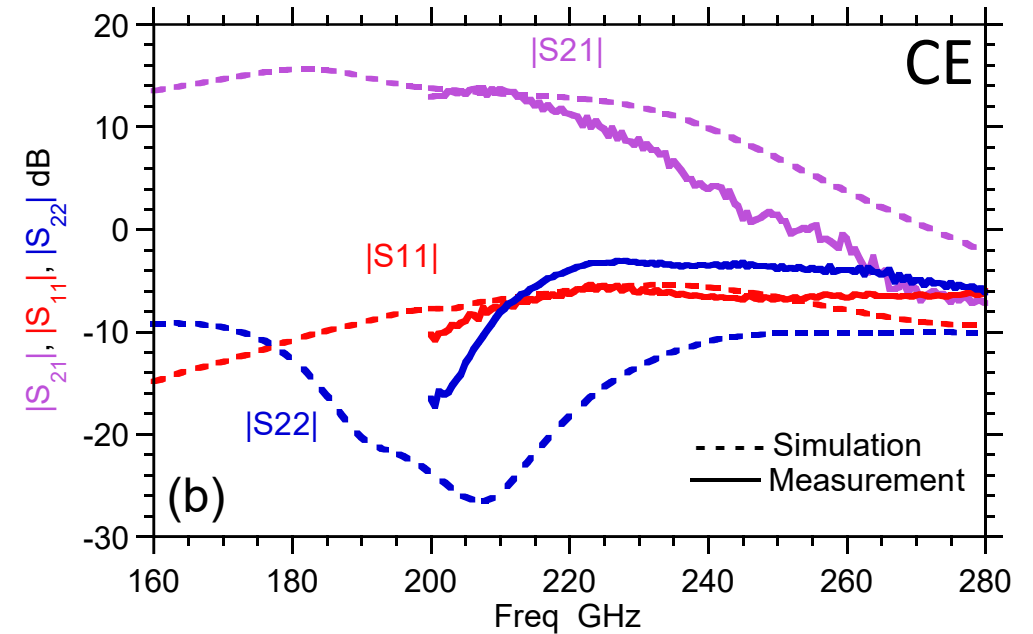
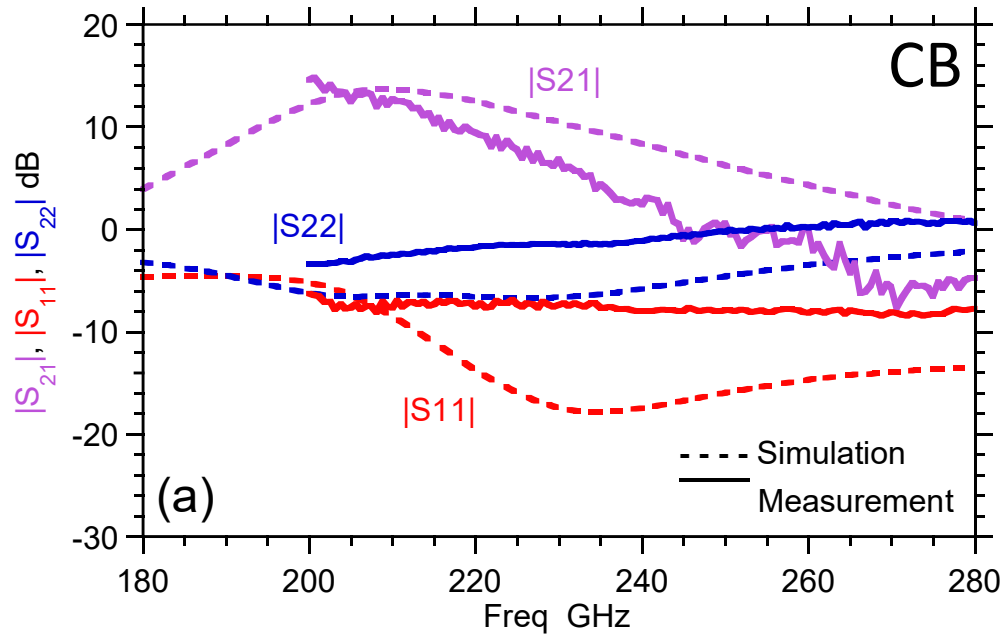
200GHz	CB, 5um, 2 stage
Gain	14.5 dB
BW	33 GHz
NF	7.4 dB
$P_{1dB,in}$	-21.1 dBm
$P_{DC}$	9.2 mW
Die Area	290umx245um
$J_{emitter}$ $V_{cb}$	0.6mA/um 0.4V



200GHz	CE, 5um, 4 stage
Gain	13 dB
BW	60 GHz
NF	7.2 dB
$P_{1dB,in}$	-18.2 dBm
$P_{DC}$	19.22 mW
Die Area	290umx465um
$J_{emitter}$ $V_{cb}$	1.0mA/um 0.56V

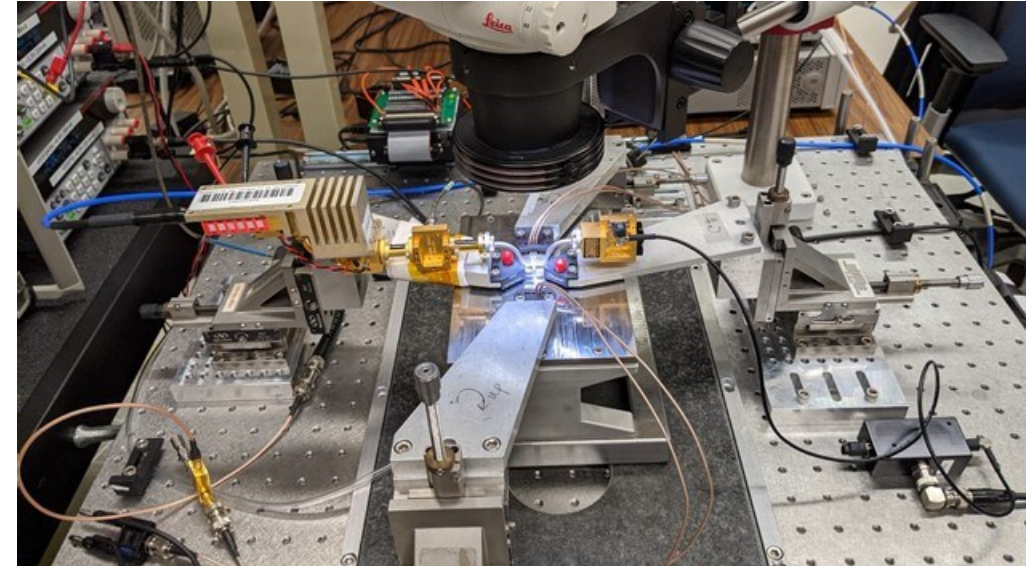
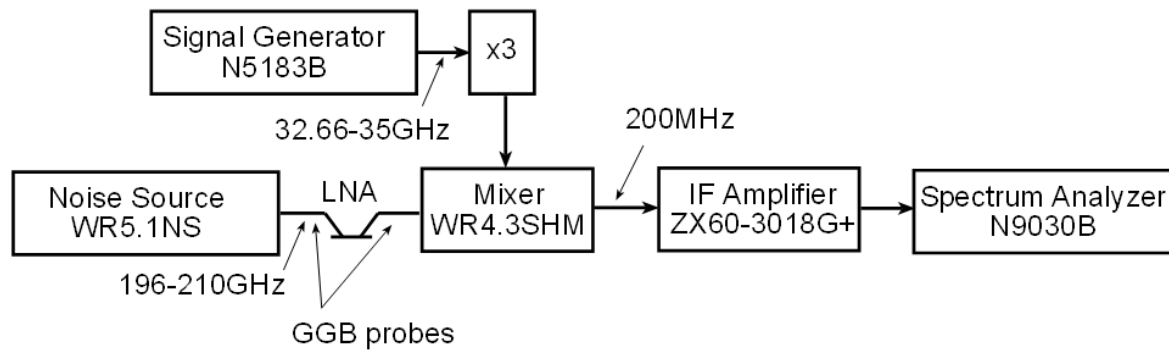


# Measurement Results: S-parameters



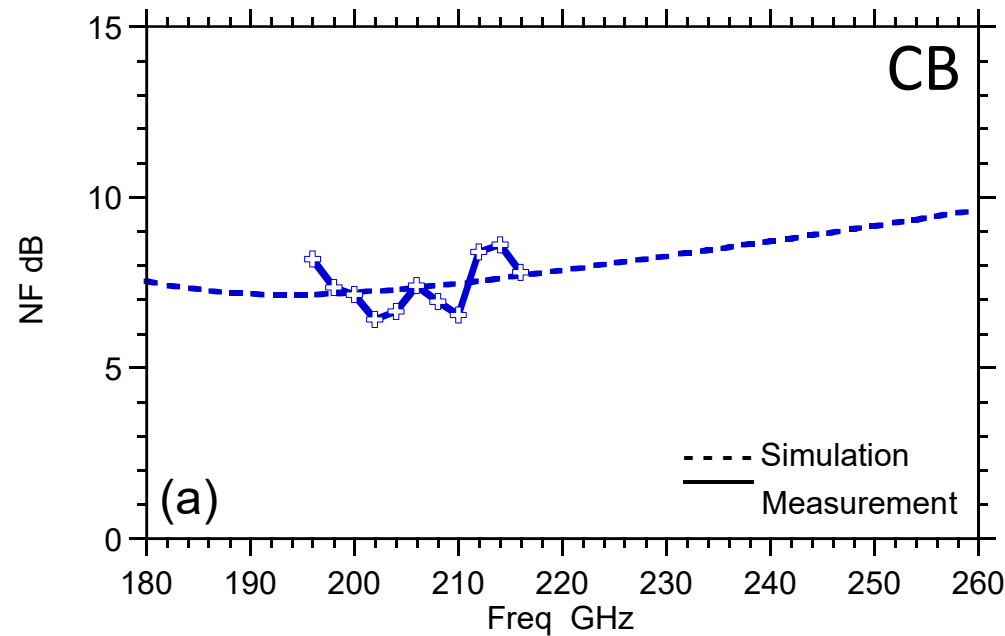
- WR-03 frequency extenders and probes, SOLT calibration
- 5 % frequency down-shift
- Good agreement between peak simulated and measured gain

# Setup: Noise Measurement

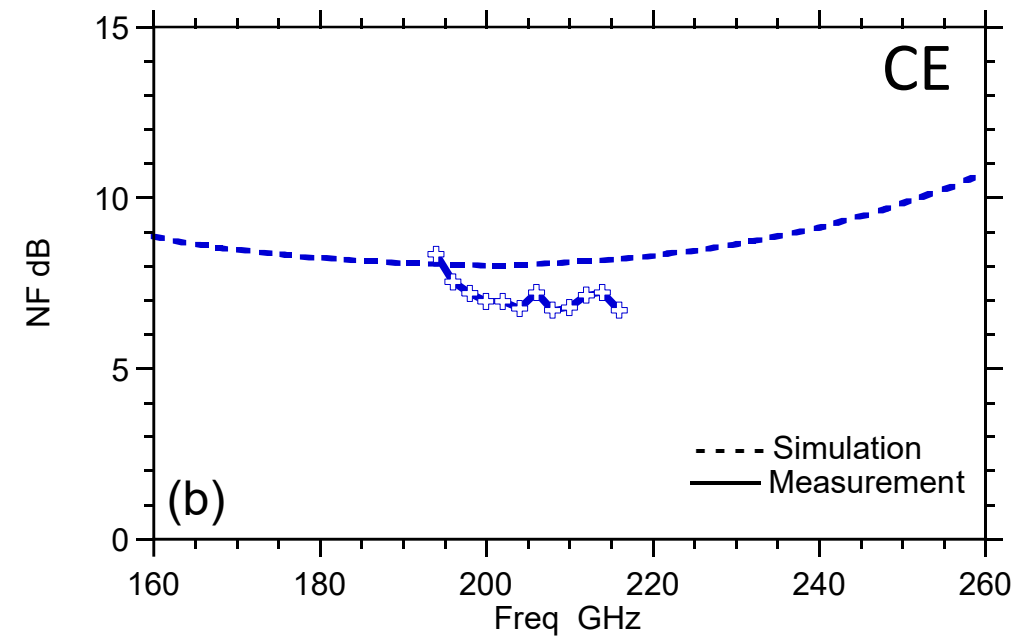


- Hot/Cold Y-parameter method (VDI-WR5.1NS noise source)
- The probe loss (2.0 dB @ 200 GHz) and is deembedded from NF
- ~20 dB BB LNA used to reduce noise contribution from SA

# Measurement Results: Noise Figure

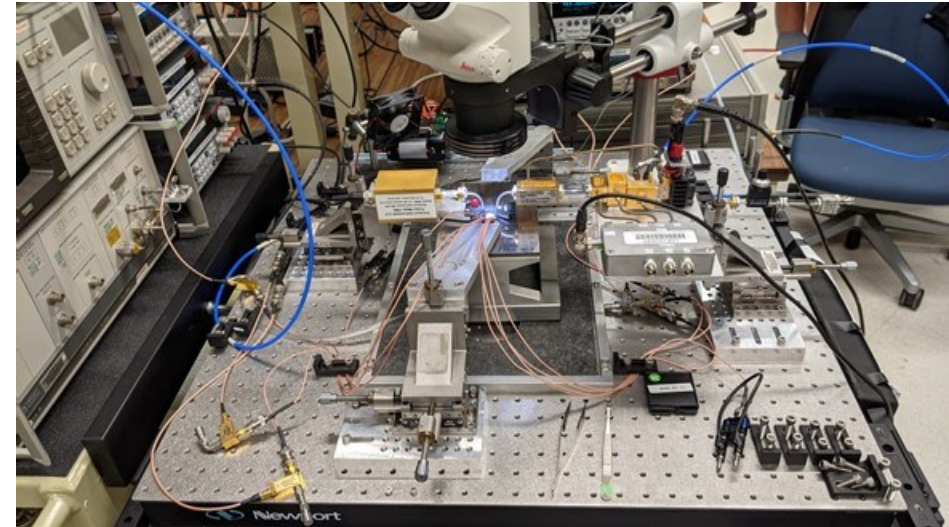
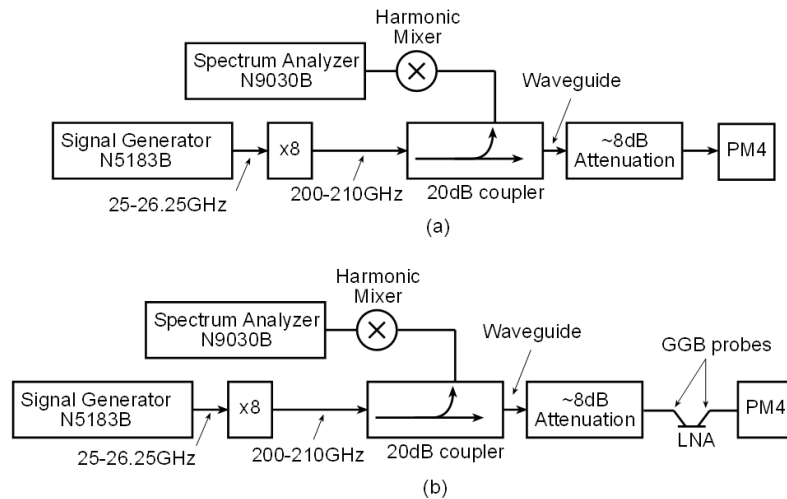


- CB:  $7.4 \pm 0.7$  dB noise figure over 196-216 GHz



- CE:  $7.2 \pm 0.4$  dB noise figure over 196-216 GHz

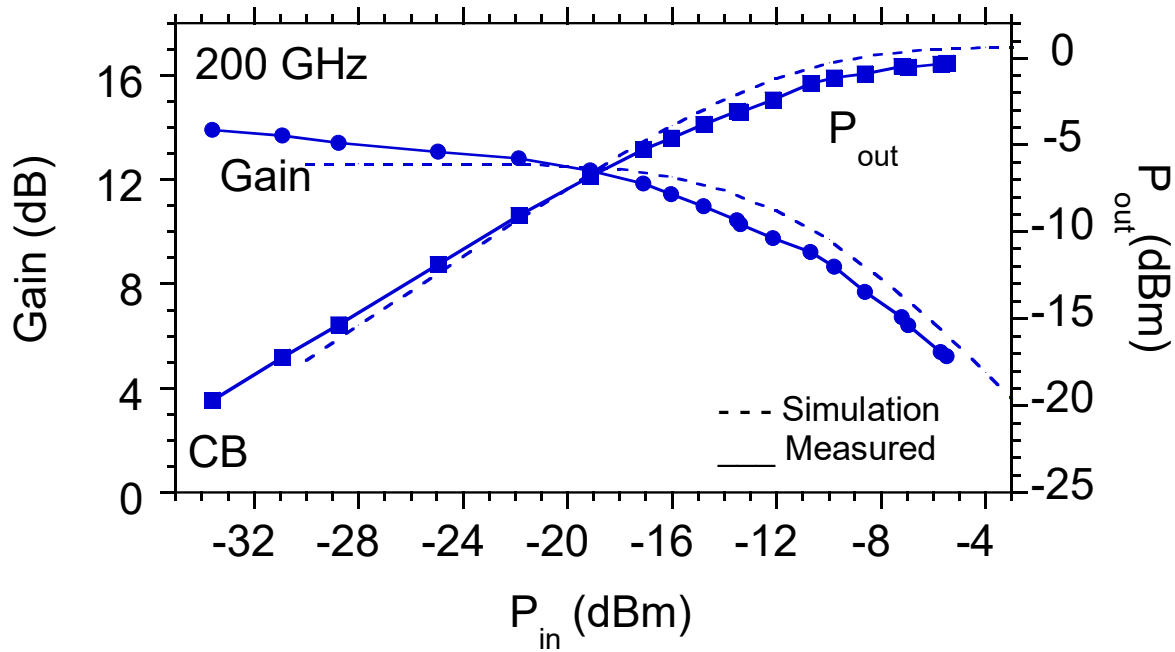
# Setup: Power Measurement



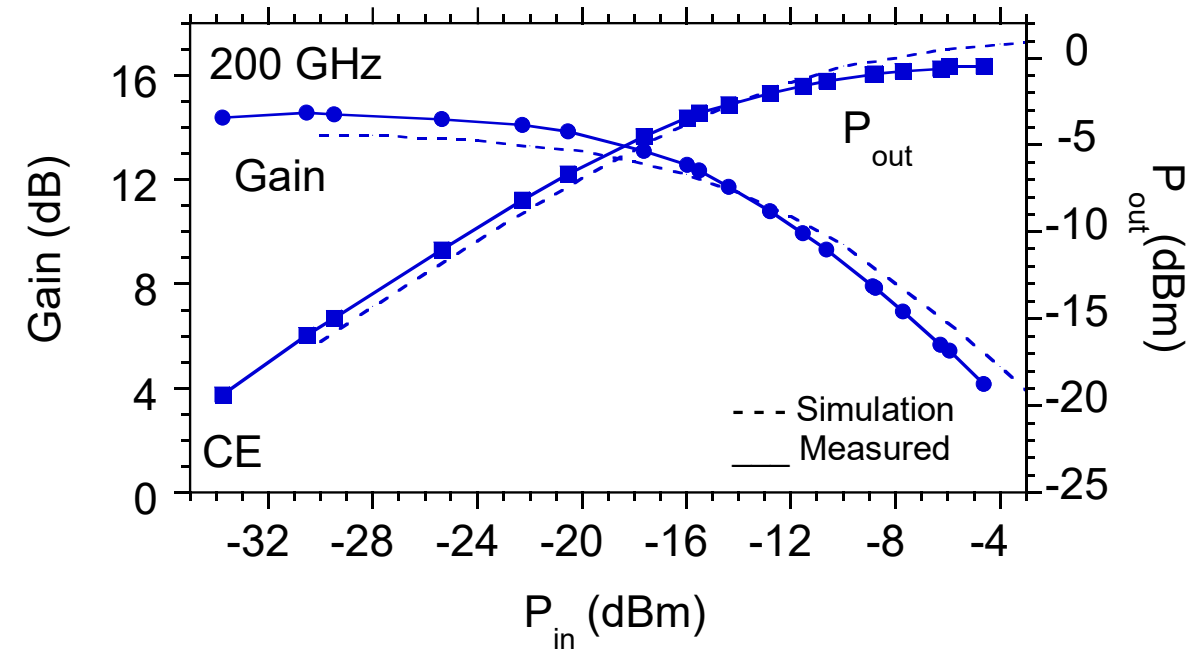
- Simultaneous input and output power measurement
- Accurate gain measurement even at low input power
- All measurements are done without lifting the probes

[13] A. S. H. Ahmed, U. Soyulu, M. Seo, M. Urteaga, J. F. Buckwalter and M. J. W. Rodwell., "A 190-210GHz Power Amplifier with 17.7-18.5dBm Output Power and 6.9-8.5% PAE.," in press, Proc. IMS2021.

# Measurement Results: Power



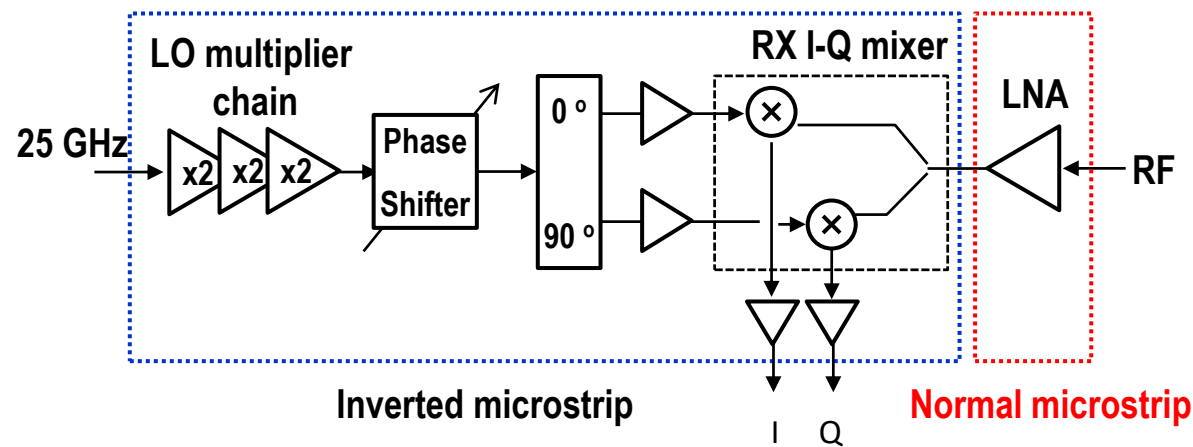
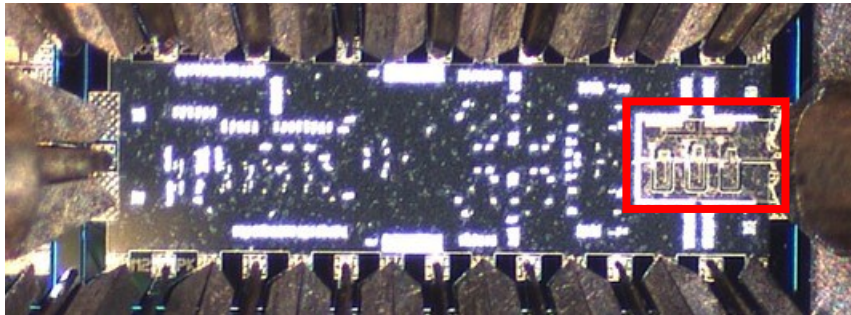
- CB: -21.1 dBm Pin1dB @ 200 GHz



- CE: -18.2 dBm Pin1dB @ 200 GHz

# Receiver with Staggered Tuned CB LNA

200 GHz direct-conversion RX



- LNA: Normal microstrip for lowest NF
- Mixer, LO multiplier, phase shifter: Inverted microstrip for low-inductance ground
- 3-stage staggered tuned CB LNA to increase 3 dB modulation bandwidth of the receiver

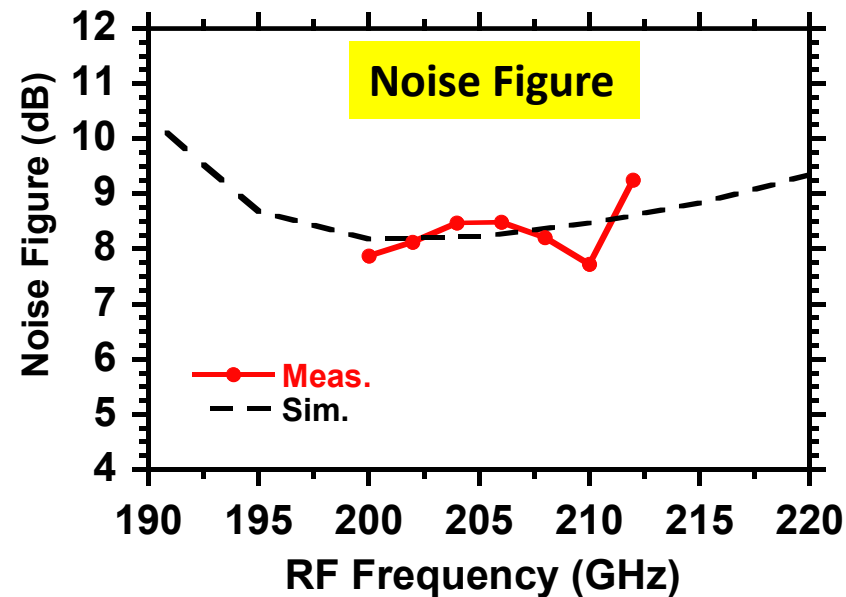
[14] M. Seo, A. S. H. Ahmed, U. Soyly, A. Farid, Y. Na and M. Rodwell, "A 200 GHz InP HBT Direct-Conversion LO-Phase-Shifted Transmitter/Receiver with 15 dBm Output Power," 2021 IEEE MTT-S International Microwave Symposium (IMS), 2021.



# Measured Rx NF and Conversion Gain

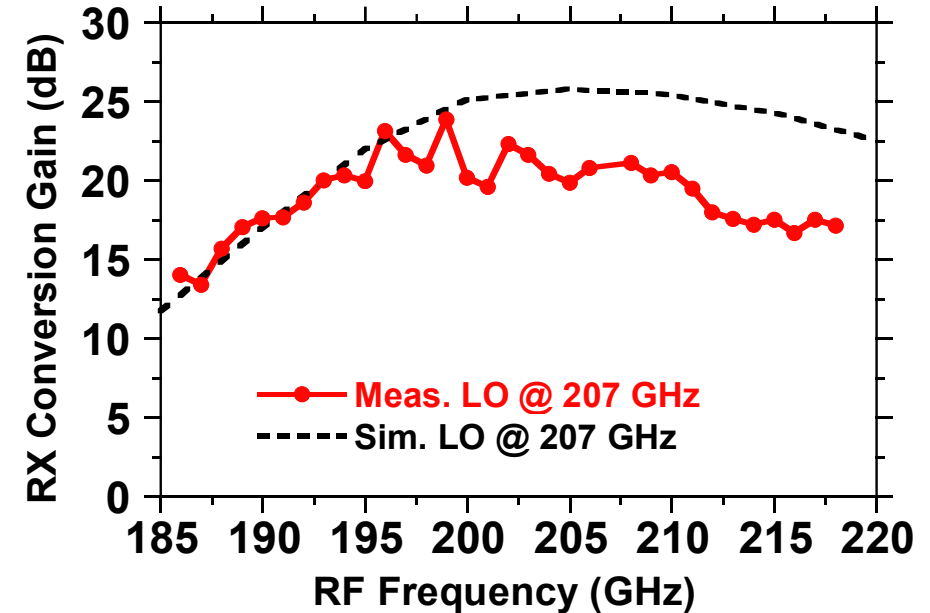


VDI-WR5.1-NS



$F_{\text{CASCADE}}$	
5 um device	6.7 dB
1-stage LNA	7.5 dB
NF of Rx	8.5-8.8 dB

**Conversion Gain @ fixed  $f_{\text{LO}}$**



- Peak conversion gain = 22 dB

# Performance Comparison

Table 1. Comparison of recently published >150 GHz low noise amplifiers

Ref.	Technology	Topology	Freq (GHz)	Gain (dB)	Gain/stage (dB)	NF (dB)	P <sub>DC</sub> (mW)
[2]	250 nm SiGe HBT	Cascode, diff. mode	156	26	8.7	8.5	-
[3]	50 nm mHEMT	CS	178-185	24.5	4.9	3.5	24
[4]	50 nm mHEMT	CS	206	16	4.0	4.8	-
[5]	32 nm CMOS	CS	200-220	10-18	1.4-2.6	11	44.5
[6]	130 nm Sige HBT	Cascode, diff. mode	220	18	6	16	151.2
[8]	250 nm InP HBT	CE	265	24	4.8	10	81.7
[9]	250 nm InP HBT	Cascode	288	8.4	8.4	11.2 at 300 GHz	-
<b>This work</b>	<b>250 nm InP HBT</b>	CE	200	13	3.25	7.2	19.22
		CB	200	14.5	7.25	7.4	9.2

- This work shows record **Noise Figure** in **HBT** technology.

# Acknowledgement

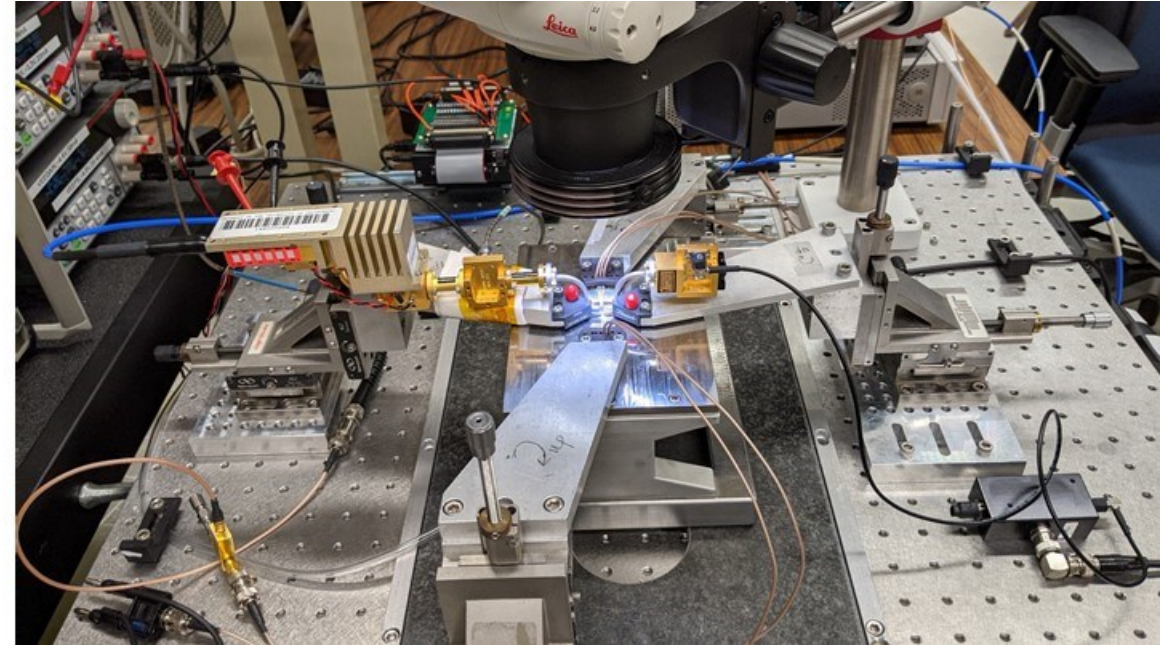
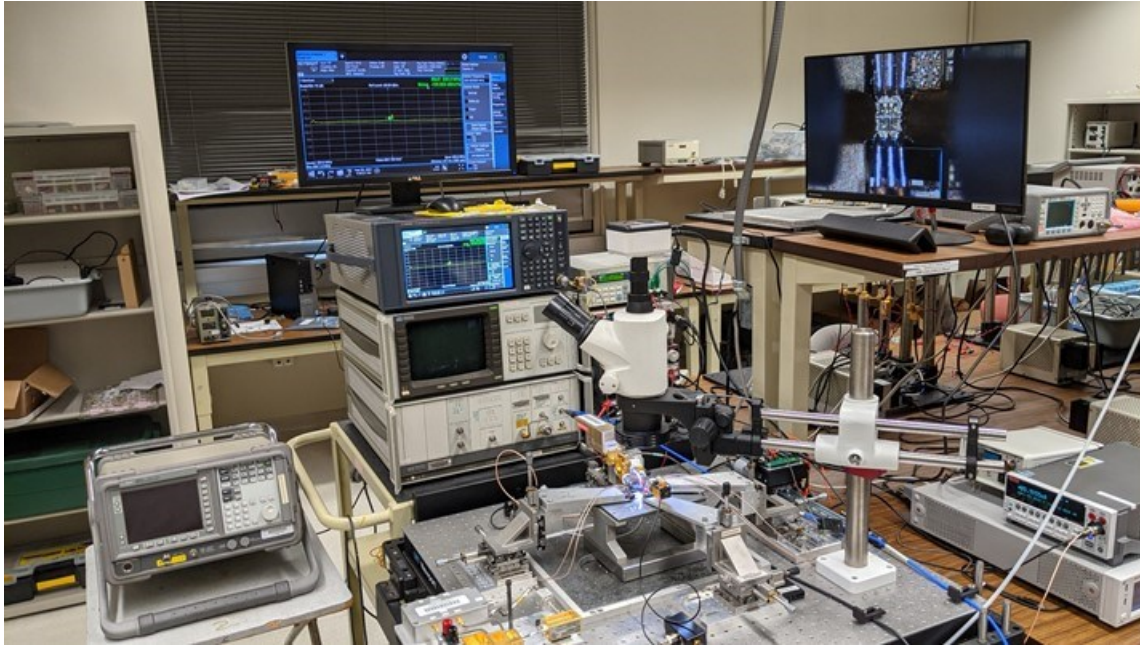
- This work was supported by ComSenTer, a JUMP program sponsored by the Semiconductor Research Corporation.
- The authors thank Teledyne Scientific & Imaging for the IC fabrication.

# Thank You!

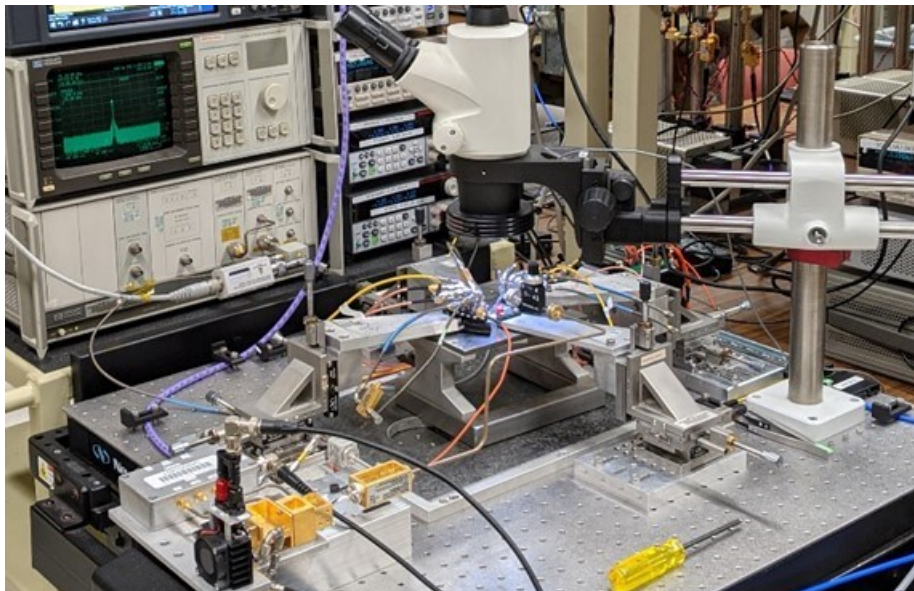
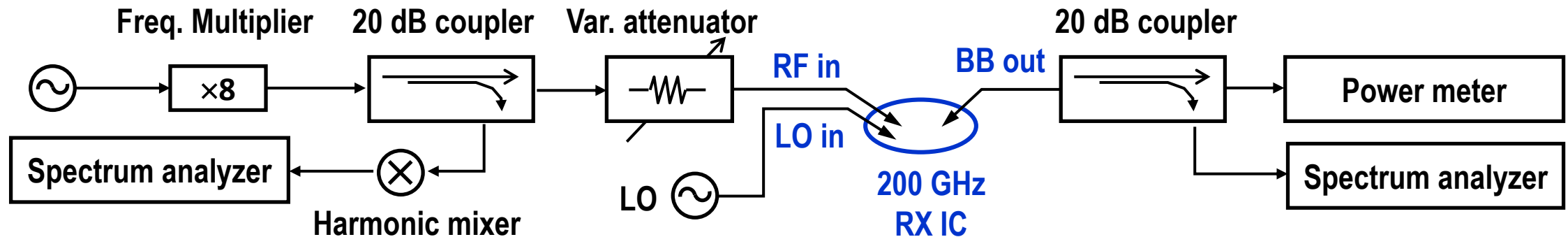
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- [2] Y. Zhao, E. Ojefors, K. Aufinger, T. F. Meister and U. R. Pfeiffer, "A 160-GHz Subharmonic Transmitter and Receiver Chipset in an SiGe HBT Technology," in IEEE Transactions on Microwave Theory and Techniques, vol. 60, no. 10, pp. 3286-3299, Oct. 2012.
- [3] G. Moschetti et al., "A 183 GHz Metamorphic HEMT Low-Noise Amplifier With 3.5 dB Noise Figure," in IEEE Microwave and Wireless Components Letters, vol. 25, no. 9, pp. 618-620, Sept. 2015.
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- [5] Z. Wang, P. Chiang, P. Nazari, C. Wang, Z. Chen and P. Heydari, "A CMOS 210-GHz Fundamental Transceiver With OOK Modulation," in IEEE Journal of Solid-State Circuits, vol. 49, no. 3, pp. 564-580, March 2014.
- [6] E. Ojefors, B. Heinemann and U. R. Pfeiffer, "Subharmonic 220- and 320-GHz SiGe HBT Receiver Front-Ends," in IEEE Transactions on Microwave Theory and Techniques, vol. 60, no. 5, pp. 1397-1404, May 2012.
- [7] K. Eriksson, S. E. Gunnarsson, V. Vassilev and H. Zirath, "Design and Characterization of HH-Band (220–325 GHz) Amplifiers in a 250-nm InP DHBT Technology," in IEEE Transactions on Terahertz Science and Technology, vol. 4, no. 1, pp. 56-64, Jan. 2014.
- [8] J. Hacker et al., "THz MMICs based on InP HBT Technology," 2010 IEEE MTT-S International Microwave Symposium, Anaheim, CA, USA, 2010, pp. 1126-1129.
- [9] M. Urteaga, Z. Griffith, M. Seo, J. Hacker, M. Rodwell, "InP HBT Technologies for THz Integrated Circuits", Proceedings of the IEEE, Vol. 105, No. 6, pp 1051-1067 June 2017.
- [10] H. A. Haus and R. B. Adler, "Optimum Noise Performance of Linear Amplifiers," in Proceedings of the IRE, vol. 46, no. 8, pp. 1517-1533, Aug. 1958.
- [11] A. Singhakowinta & A. R. Boothroyd, "Gain Capability of Two-port Amplifiers", International Journal of Electronics, Volume 21, Issue 6, 1966, pages 549-560.
- [12] H. Fukui, "Available Power Gain, Noise Figure, and Noise Measure of Two-Ports and Their Graphical Representations," in IEEE Transactions on Circuit Theory, vol. 13, no. 2, pp. 137-142, June 1966.
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# Setup: Noise Measurement (Pictures)

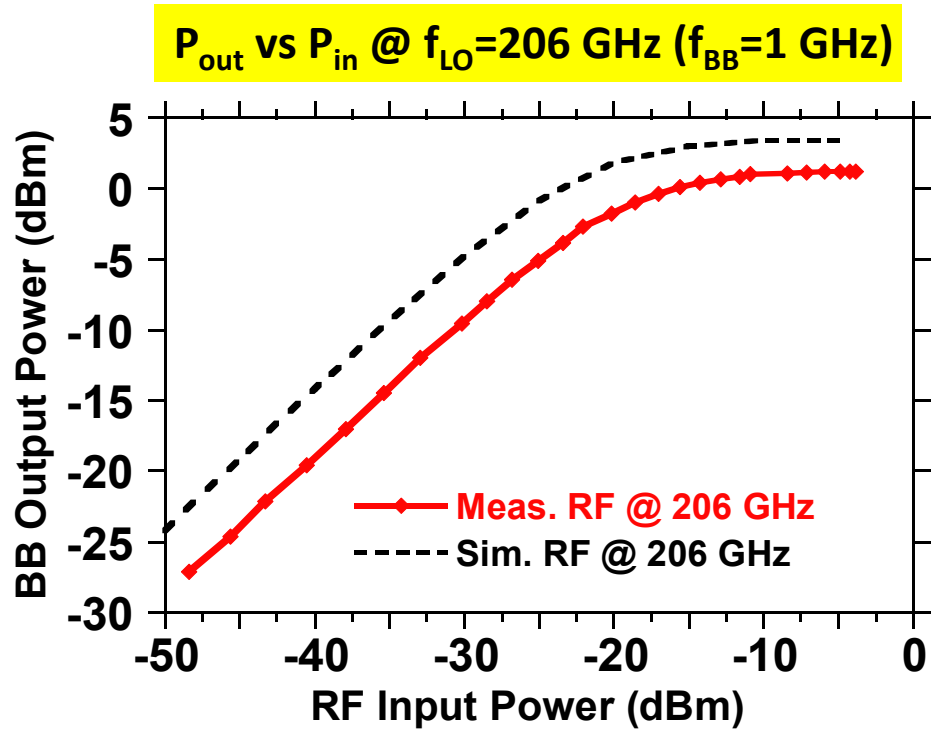


# Receiver Testing Setup

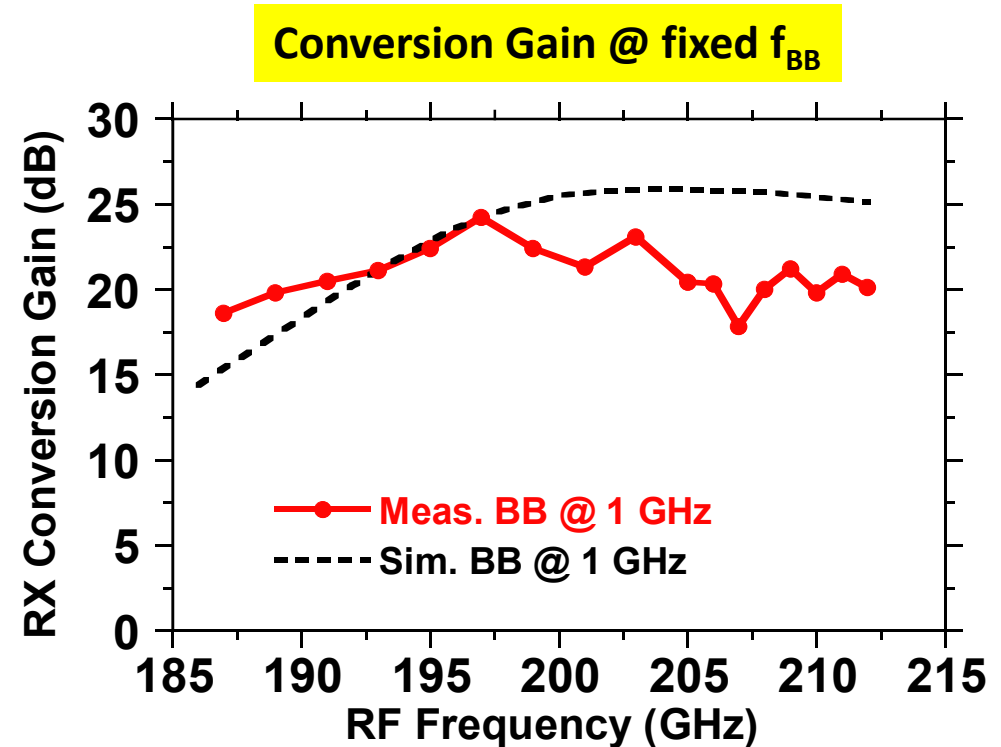


- 140-220 GHz (WR5) on-wafer testing
- Simultaneous freq. & power testing
- RX driven by multiplier & variable attenuator

# Measured Rx Power and Conversion Gain



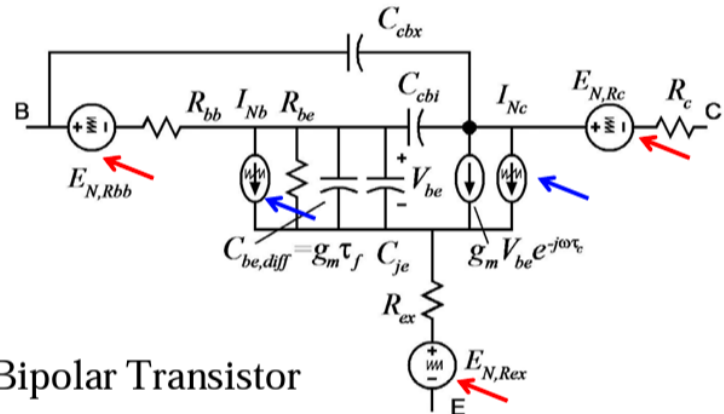
- Input  $P_{1dB} = -24$  dBm
- $P_{sat} = +1$  dBm
- $P_{DC} = 825$  mW



- Peak conversion gain = 22 dB
- LO multiplier tuning bandwidth > 25 GHz

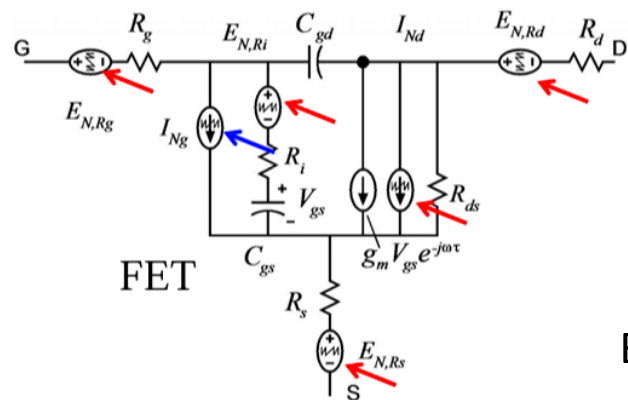


# Noise: HBT vs HEMT



Bipolar Transistor

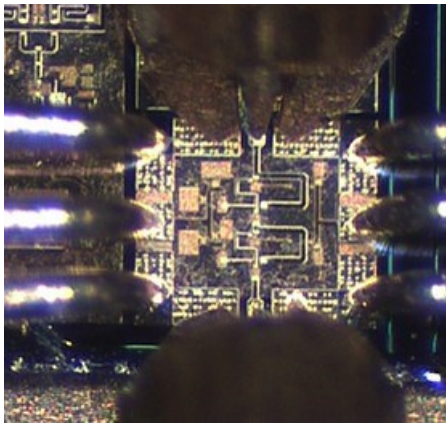
— thermal noise  
— shot noise



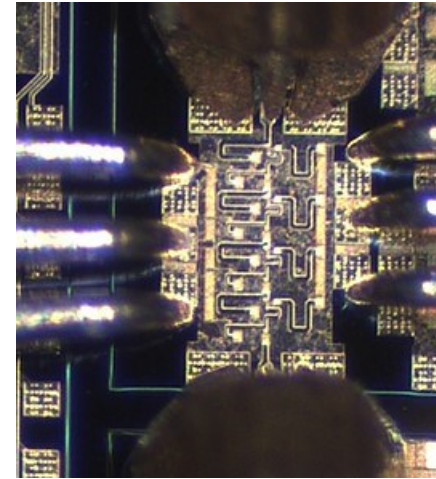
FET

By courtesy of Prof. Mark Rodwell

# CB and CE Low Noise Amplifiers



210GHz, Simulated	CB, 5um, 2 stage
Gain	13.6 dB
BW	33 GHz
NF	7.4 dB
$P_{1dB,in}$	-29.7 dBm
$P_{DC}$	9 mW
Die Area	290umx245um
$J_{emitter}$ $V_{cb}$	0.6mA/um 0.4V



200GHz, Simulated	CE, 5um, 4 stage
Gain	13.2 dB
BW	60 GHz
NF	8 dB
$P_{1dB,in}$	-16.8 dBm
$P_{DC}$	19.2 mW
Die Area	290umx465um
$J_{emitter}$ $V_{cb}$	1.0mA/um 0.56V

